Optimal Control Problems II

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1

1.1

Find and classify the critical points and the critical value of $L(u) = \frac{1}{2}u^{\top}Qu + S^{\top}u$ if

a.
$$Q = \begin{pmatrix} -1 & 1 \\ 1 & -2 \end{pmatrix}$$
, $S = \begin{pmatrix} 0 \\ 1 \end{pmatrix}$
b. $Q = \begin{pmatrix} -1 & 1 \\ 1 & 2 \end{pmatrix}$, $S = \begin{pmatrix} 0 \\ 1 \end{pmatrix}$

1.2

A meteor is in a hyperbolic orbit with respect to the earth, described by $\frac{x^2}{a^2} - \frac{y^2}{b^2} = 1$. Find the minimum distance to a satellite at a fixed position (x_1, y_1) .

1.3

a. Find the rectangle of maximum perimeter that can be inscribed inside an ellipse $\frac{x^2}{a^2} + \frac{y^2}{b^2} = 1$. b. Find the rectangle of maximum area that can be inscribed inside an ellipse $\frac{x^2}{a^2} + \frac{y^2}{b^2} = 1$.

$\mathbf{2}$

$\mathbf{2.1}$

For the bilinear system $\dot{x} = Ax + Bu + Dxu$ with a scalar input $u \in \mathbf{R}$, minimize the cost

$$J = \frac{1}{2}x^{\top}Sx|_{T} + \frac{1}{2}\int_{0}^{T}x^{\top}Qx + ru^{2} dt$$

Show that the optimal control involves a state-costate inner product. The optimal state-costate equations contain cubic terms and are very difficult to solve.

$\mathbf{2.2}$

Find the optimal control for the scalar plant $\dot{x} = u$, $x(t_0) = x_0$, with performance index

$$J(t_0) = \frac{1}{2}x^{\top}Sx|_T + \frac{1}{2}\int_{t_0}^T ru^2 dt$$

a. Solve the Riccati using separation of variables.

b. Suppose x(T) is fixed. Find the optimal control as a function of $x(t_0), x(T)$.

c. Use the results of Part b to develop a state-feedback control law. Solve for $x(t_0)$ and substitute to get an optimal input of the form u(t) = g(t)x(t) + h(t). Compare with the optimal control minimizing the cost to go J(t) in [t, t+T].

 $\mathbf{2.3}$

Let V,W be the $n \times n$ solutions to the Hamiltonian system

$$\begin{pmatrix} \dot{V} \\ \dot{W} \end{pmatrix} = \begin{pmatrix} A & -BR^{-1}B^{\top} \\ -Q & -A^{\top} \end{pmatrix} \begin{pmatrix} V \\ W \end{pmatrix}$$

with the boundary condition W(T) = S(T)V(T). Show that the solution to associated Riccati differential equation $-\dot{S} = A^{\top}S + SA - SBR^{-1}B^{\top}S + Q$ is given by $S(t) = W(t)V(t)^{-1}$.

$\mathbf{2.4}$

For the cart system $\dot{x}_1 = x_2$, $\dot{x}_2 = u$, minimize the cost

$$J = \frac{1}{2} \int_0^\infty x_1^2 + 2vx_1x_2 + qx_2^2 + u^2 dt$$

where $q - v^2 > 0$. Find the solution to the ARE, the optimal control and the optimal closed-loop system. Also, plot the loci of the closed-loop poles as q varies from 0 to ∞ .

3

3.1

Consider the harmonic oscillator

$$\dot{x} = \begin{pmatrix} 0 & 1 \\ -\omega_n^2 & 0 \end{pmatrix} x + \begin{pmatrix} 0 \\ 1 \end{pmatrix} u$$

- - -

Find the optimal control to drive any initial state to zero in minimum time, subject to $|u(t)| \leq 1, \forall t$.

a. Find and solve the costate equations.

b. Sketch the phase-plane trajectories for u = 1 and u = -1.

c. Find the switching curve and derive a minimum-time feedback control law.

3.2

Develop a minimum-fuel control law for Problem 3.1.

Ref: F. Lewis and V. Syrmos, Optimal Control. Wiley, New York, 1995.

Solutions to Optimal Control Problems II

1.1
$$L = \frac{1}{2} u^{T} Q u + S^{T} u$$
b) Critical pt:
$$L_{u} = 0 = Q u + S \Rightarrow u_{+} = -Q^{-1}S$$

$$= \begin{bmatrix} 1 \end{bmatrix}$$
Optimal Cast:
$$L(u^{+}) = \frac{1}{2}u^{T}$$
 Hessian
$$L_{uu} = Q$$
 is
negative definite $\Rightarrow u_{+}$ is a maximum
2) Critical point:
$$L_{u} = 0 \Rightarrow u_{+} = -Q^{-1}S = \begin{bmatrix} -\frac{1}{2}S \\ -\frac{1}{2}S \end{bmatrix}$$
Optimal cast is
$$L(u^{+}) = -\frac{1}{2}K + \frac{1}{2}K + \frac$$

Substituting the last two in the equation of the hyperbola,

$$\frac{a^{2} x_{1}^{2}}{(\lambda + a^{2})^{2}} - \frac{b^{2} y_{1}^{2}}{(\lambda - b^{2})^{2}} = 1.$$
This equation has two solutions for γ .
The left hand side as a function of γ looks like:
The roots can be found by means of numerical methods,
or as roots of polynomials (note that the conversion of
this eqn to a polynomial will introduce new roots that must be
discarded).
Finally, at the minimum, the curvature matrix
Lun = $\left[-\int_{T}^{T} \int_{T}^{T} \int_{T}^{T} \int_{T} \int_{$

1.3 (i) The optimization problem is
win
$$L = -4(x+y)$$

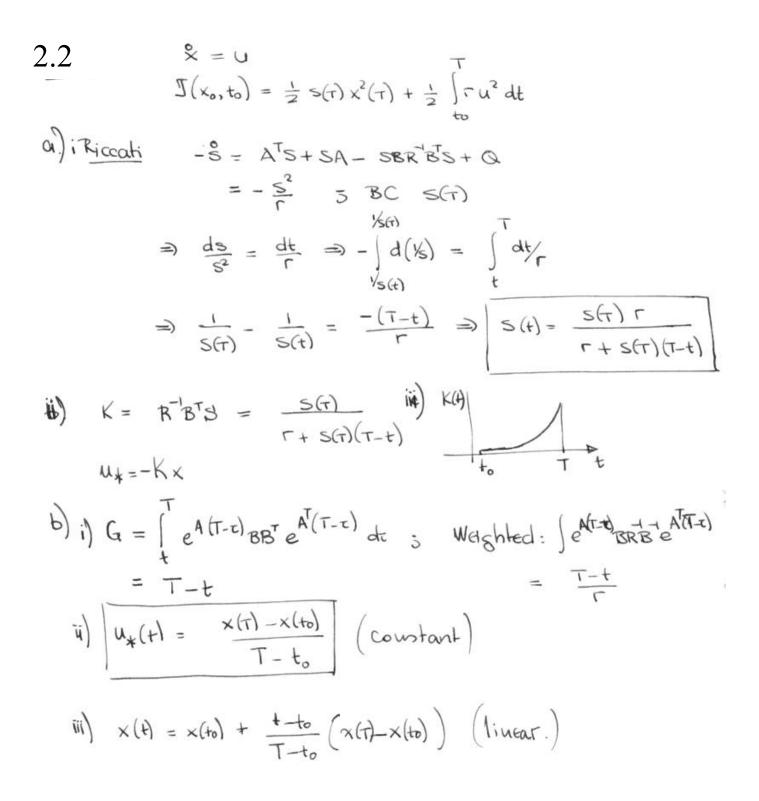
st. $f(x,y) = \frac{x^2}{a^2} + \frac{y^2}{b^2} - 1 = 0$
Fran this, $H = -4(x+y) + \lambda(\frac{x^2}{a^2} + \frac{y^2}{b^2} - 1)$
and $H_{\lambda} = 0 = \frac{x^2}{a^2} + \frac{y^2}{b^2} - 1$
 $H_{\lambda} = 0 = -4 + \frac{2}{\alpha^2} \implies x = \frac{2\alpha^2}{\lambda}$
 $H_{\gamma} = 0 = -4 + \frac{2}{\alpha^2} \implies x = \frac{2\alpha^2}{\lambda}$
 $H_{\gamma} = 0 = -4 + \frac{2}{\alpha^2} \implies y = \frac{2b^2}{\lambda}$
 $\Rightarrow x^* = \frac{a^2}{\sqrt{a^2+b^2}}, \quad y^* = \frac{b^2}{\sqrt{a^2+b^2}}$
 $\Rightarrow max$ Perimeter of a redaugle is $4(x^4y^*) = \frac{4\sqrt{a^2+b^2}}{\lambda}$
(ii) The optimization problem now is
min $-4xy$
 $= 5t. f(x,y) = 0$
 $H_{\lambda} = 0 = f(x,y)$
 $H_{\chi} = 0 = -4x + \frac{2}{\alpha^2} \implies x = \frac{2a^2y}{y}$ $\Rightarrow \frac{x^2}{a^2} = \frac{y^2}{b^2}$
 $\Rightarrow x^* = \frac{a}{\sqrt{2}}, \quad y^* = \frac{b}{\sqrt{2}}$ and $\lambda^* = 2ab$
Then the max area of a redungle is $4x^*y^* = 2ab$

2.1

$$\begin{aligned}
\hat{x} &= Ax + Dxu + bu \\
\hat{J} &= \frac{1}{2} x^{T}Sx \Big|_{T} + \frac{1}{2} \int_{T}^{T} x^{T}Qx + tu^{2} \\
H &= \frac{1}{2} x^{T}Qx + \frac{1}{2}ru^{2} + \sqrt{r} (Ax + Dxu + bu) \\
\text{Stationarity cond.} \quad \frac{\partial H}{\partial u} = 0 \Rightarrow ru + (Dx + b)^{T} \Re = 0 \\
\Rightarrow u_{*} &= -\frac{1}{r} (Dx + b)^{T} \Re \\
&= -\frac{b^{T} \Re}{r} - \frac{x^{T} (D + D^{T}) \Re}{2r}
\end{aligned}$$

: The optimal input contains a state-costate "inner product" (It is formally an inner product if $D + D^T > 0$). Substituting the optimal input into the state & costate equs.:

$$\begin{aligned} & & = A \times + (D \times + b) \left(\frac{-i}{r} \right) (D \times + b)^{T} \lambda \\ & = A \times - (D \times + b) (D \times + b)^{T} \frac{\eta}{r}, \qquad \longrightarrow \quad \text{addic in } \times, \eta \\ & & & & & \\ & & & & \\ & & & & \\ & & & & \\ & & & & \\ & & & & \\ & & & & \\ & & & & \\ & & & \\ & & & & \\ &$$



c)
$$x(t_0) = \frac{T-t_0}{T-t} x(t) - \frac{t-t_0}{T-t} x(T)$$

 $\Rightarrow u_*(t) = \frac{t}{T-t_0} x(T) - \frac{t}{T-t_0} \left[\frac{T-t_0}{T-t} x(t) - \frac{t-t_0}{T-t} x(T) \right]$
 $= \frac{1}{T-t} \left[x(T) - x(t) \right] = \left| h(t) = \frac{x(T)}{T-t} \right|$
(Comparing with (a).:
We want $x(T) \rightarrow 0$ with a high penalty to emulate
the fixed final state, so $z(T) \rightarrow \infty$.
Then $K \rightarrow \frac{1}{T-t}$, $u_*(t) \rightarrow -\frac{x(t)}{T-t}$
iii) $u_*(t) = \frac{t}{T-t} \left[x(T) - x(t) \right]$
win Cast - to-go $J(t_0)$ at time to: $u_*(t_0) = -k(t_0) x(t_0)$
 $= -\frac{s(T)}{r+s(T)(T-t_0)} x(t_0)$
They approach each other for $s(T) \rightarrow \infty$
(otherwise, T must be reformulated
in terms of a target state)

2.3

$$\begin{pmatrix} V \\ W \end{pmatrix} = \begin{pmatrix} A & -BR^{-}B^{-} \\ -Q & -A^{+} \end{pmatrix} \begin{pmatrix} V \\ W \end{pmatrix}$$

$$W(T) = S(T)V(T)$$
Show
$$S(t) = W(t)V^{-1}(t).$$
We verify the Riccodi: $-S = A^{T}S + SA - SBR^{-}D^{T}S + Q$

$$S = WV^{-} + WV^{-} = WV^{-} - WV^{-}VV^{-}$$

$$= (-QV - A^{T}W)V^{-} - WV^{-}(AV - BR^{-}B^{T}W)V^{-}$$

$$= -Q - A^{T}S - SA + SBR^{-}B^{T}S \quad (Verified)$$
Then, $Ct = T$, $S(T) = W(T)V^{-}(T) \Rightarrow S$ catisfies $ODS + BC$

$$\Rightarrow SO'n.$$

2.4 $\begin{aligned} x_{1} &= x_{2} \\ x_{2} &= u \end{aligned} \qquad \begin{aligned} y &= \int_{0}^{\infty} x^{T} \begin{pmatrix} 1 & v \\ v & q \end{pmatrix} x + u^{2} \\ &\leftarrow Q \rightarrow \\ A &= \begin{pmatrix} 0 & 1 \\ 0 & 0 \end{pmatrix}, B &= \begin{pmatrix} 0 \\ 1 \end{pmatrix}, BB^{T} &= \begin{pmatrix} 0 & 0 \\ 0 & 1 \end{pmatrix} \end{aligned}$

 $\begin{aligned} \text{Riccq}\Pi: \quad A^{T}S + SA - S^{T}BB^{T}S + Q &= 0 \quad (\text{Algebraic because} \\ \text{let } S = \begin{pmatrix} s_{1} & s_{2} \\ s_{2} & s_{3} \end{pmatrix}. \quad \text{Then,} \\ \begin{pmatrix} 0 & s_{1} \\ s_{1} & 2s_{2} \end{pmatrix} - \begin{pmatrix} s_{2}^{2} & s_{2}s_{3} \\ gb_{3} & s_{3}^{2} \end{pmatrix} + \begin{pmatrix} 1 & v \\ v & q \end{pmatrix} = 0 \\ \Rightarrow \begin{pmatrix} s_{2}^{2} = 1 \\ s_{3}^{2} = 1 \\ s_{3}^{2} = q + s_{2} \\ s_{1} = s_{2}s_{3} - v \\ &= \end{pmatrix} \quad \begin{aligned} S po \quad s_{1}s_{3} > s_{2}^{2}, \quad s_{1} > 0, \quad s_{3} > 0 \\ \Rightarrow \int s_{2}^{2} = q + s_{2} \\ &= \end{pmatrix} \quad s_{2}^{2} = +\sqrt{1} = 1, \quad s_{3}^{2} = \sqrt{q + 2}, \quad s_{1} = \sqrt{q + 2}, \quad s_{2} = \sqrt{q + 2}, \quad s_{1} = \sqrt{q + 2}, \quad s_{2} = \sqrt{q + 2}, \quad s_{1} = \sqrt{q + 2}, \quad s_{2} = \sqrt{q + 2}, \quad s_{1} = \sqrt{q + 2}, \quad s_{2} = \sqrt{q + 2}, \quad s_{1} = \sqrt{q + 2}, \quad s_{2} = \sqrt{q + 2}, \quad s_{3} = \sqrt{q + 2}, \quad s_{1} = \sqrt{q + 2}, \quad s_{2} = \sqrt{q + 2}, \quad s_{3} = \sqrt{q + 2}, \quad s_{3$

$$\Rightarrow S = \begin{pmatrix} Jq+2-v & 1 \\ 1 & Jq+2 \end{pmatrix} \quad \text{which is PD as long as } q > v^{2} \\ K = R^{T}B^{T}S = \begin{bmatrix} 1, Jq+2 \end{bmatrix} \\ \mathfrak{U}_{*}(t) = -K \propto (t) \\ \mathbb{O}phiwal \ closed \ loop: \qquad \overset{\circ}{x} = (A-BK) \times \frac{1}{2} A-BK = \begin{pmatrix} 0 & 1 \\ -1 & -\sqrt{q+2} \end{pmatrix} \\ \mathbb{P}ools \ of \ Char. \ Eqn: \qquad -Jq+2 \pm \sqrt{q-2} \qquad \begin{pmatrix} -\frac{1}{12} \pm 1\frac{1}{\sqrt{2}} \\ -1, -1 & 3 & q=2 \\ -\sqrt{q}, -\frac{1}{\sqrt{q}} \\ -1 & \sqrt{q} \end{pmatrix} \\ \mathbb{P}ools \ of \ Char. \ Eqn: \qquad -Jq+2 \pm \sqrt{q-2} \qquad \begin{pmatrix} -\frac{1}{12} \pm 1\frac{1}{\sqrt{2}} \\ -1, -1 & 3 & q=2 \\ -\sqrt{q}, -\frac{1}{\sqrt{q}} \\ -\frac{1}{\sqrt{2}} \\ -\frac{1}{\sqrt{2}} \\ \mathbb{P}ools \ of \ r q > 0. \qquad (guaranheed \ fraw Ler \ theory) \\ Also \ for \ q > -2 \end{cases}$$

3.1

e

$$\begin{aligned} \hat{\mathbf{x}} = A\mathbf{x} + B\mathbf{u} & A = \begin{pmatrix} 0 & 1 \\ -\omega^2 & 0 \end{pmatrix}, B = \begin{pmatrix} 0 \\ 1 \end{pmatrix} \\ B = \begin{pmatrix} 1 \\ 1 \end{pmatrix} \\ H = \begin{pmatrix} 1 + \lambda^T A \mathbf{x} + \lambda^T B \mathbf{u} \end{pmatrix} = A^T \mathbf{x} & A = \begin{pmatrix} 0 & 1 \\ -\omega^2 & 0 \end{pmatrix}, B = \begin{pmatrix} 0 \\ 1 \end{pmatrix} \\ B = \begin{pmatrix} 0 \\ 1$$

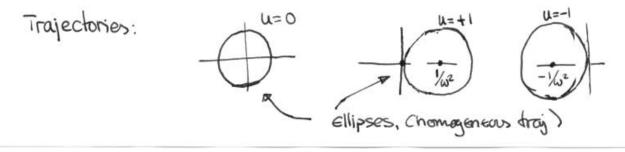
and
$$A_2(t) = A_{02} \cos \omega t - \frac{\alpha_{01}}{\omega} \sin \omega t$$
 $\left(= e^{-A^T t} A_0 \right)$
 A_{01}, A_{02} will be chosen to satisfy the BC, $x(0) = x_0, x(T) = 0$.
Thus the satisfy the BC, $x(0) = x_0, x(T) = 0$.

the optimal input has the form

$$u_{*}(t) = -\operatorname{sign} (\alpha \cos \omega t + \beta \sin \omega t)$$

Rewrite in
magnitude-phase = sign ($\rho \cos(\omega t + \phi)$)
= sign($\cos(\omega t + \phi)$)

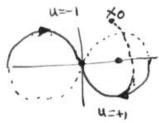
=) it is a square wave with period $T = \frac{2\pi}{w}$, same as the oscillator, and its only free parameter is the starting phase. Effectively it will look like a number of complete periods with beginning and ending segments of arbitrary duration $(< \underline{x})$



Notice that when
$$u = +1$$
, $\vec{x}_1 = \vec{x}_2$
 $\vec{x}_2 = -\vec{\omega}\vec{x}_1 + u = -\vec{\omega}\vec{x}_$

=> in the shifted coordinates the system is an unforced oscillate =) The trajectories will be ellipses centered at 0, -1/2, +1/22 for u= 0, -1, +1 respectively. (with a transformation $\overline{X}_1 = W X_1$, $\overline{X}_2 = X_2 \implies \overline{X}_1 = W \overline{X}_2$, $\overline{X}_2 = -W \overline{X}_1$, the ellipses become circles).

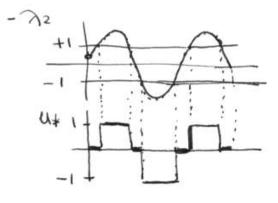
The last part of the switching curve is quite obvious. It is the ellipse (circle in normalized coordinates) that passes thru the origin (notice the orientation)



starting with an TC that does not belong to these two arcs, there will be at least one switching before the state becomes zero. (since A(A) ≠ real, He womber of So, starting from an IC Xo (see graph) the input would be initially -1, switching to +1 when x(f) hits the switching curve. The -1 part of the trajectory is a circle contered at - 1/22. To find the next switching point me observe that it must come from a square wave input that should suitch ma half-period. In the normalized coordinates, the entire circle is covered in one period and equal time segments correspond to equal arcs.

Hance the other switching point would be the arc that is - symmetric to the {u=+14-suitching - about - 1/w2 Honce, because of symmetry, the symmetry about a point can be riewed as a shifting, that would produce a much simpler expression for the switching curve. shifted half-circles Notes : It is instructive to look at the evolution of the system tackwards in the. Starting with $x_0 = 0$, solve (in Simulark) $\hat{X}_{=} - (A_X + B_U)$, $\hat{A} = A^T A$, $U = -sign A_Z$. Then, to arrive to any point in the state space, we simply need to follow the appropriate trajectory thruthe switching curve 1 in normalized coordinate 2). Elapsed time is proportional to the arc angle. To illustrate this point consider the following two. trajectories that have the same initial of final points (symmetric about x1-axis) The angle for u=+1 is clearly larger than the angle for u= a (similarly for u=-1) So, between xo and xy the clapsed time will be: $t_{u=1} > t_{u=0} > t_{u=-1}$

For the min fuel problem (fixed final time)



Small p(p < 1) implies that $u_{\pm} = 0 \implies x(0) = 0$ Large p(p >> 1) implies that $u_{\pm} = \pm 1$ except for very short time intervals \implies u_{\pm} approaches the time-optimal input. This situation occurs when $T \rightarrow T_{min}$

Clearly, if $T < T_{min}$ there is no solution. The win fuel trajectory is now a function of two parameters p and ϕ (or A_{01}, A_{02}). In the min-time problem we found that each value of ϕ would be associated with a "spiral" of initial conditions, and p was irrelevant. In the min-fiel problem the extra parameter p is an ociated with T-Tmin. A rough harmonic analysis shows that the control is applied $(u=\pm 1)$ when x_1 is small and the system "coasts" (u=0) when x_1 is large. But the thresholds are not simple functions. Again, solving the optimal equations u=1, u=-1, w_1 illustration of the optimal trajectories. u=+1, u=0 trajectory