

“High performance horn antenna design (II)”

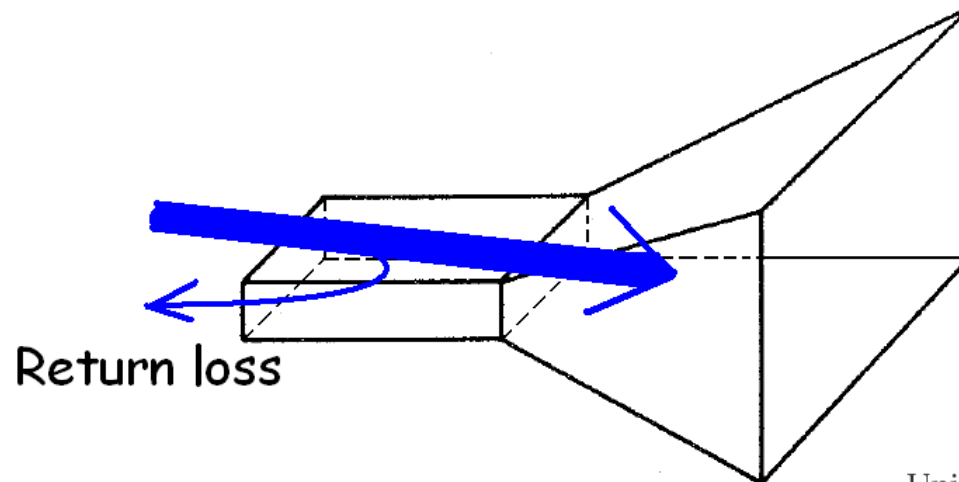
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Feed-horns

Feed-horns

- Feed-horns are widely used in microwave and millimeter wave bands because they provide moderate to high gain, low return loss, wide bandwidth and they are usually relatively easy to manufacture
- Feed-horn theoretical simulations are extremely similar to measurements, so feed-horn behavior can be easily predicted prior to be manufactured



Feed-horns

- Feed-horns can be classified:
 - Rectangular feed-horns:
 - Sectoral feed-horns (E-plane and H plane)
 - Pyramidal feed-horns
 - Conical feed-horns:
 - Smooth-walled conical feed-horns
 - Corrugated conical feed-horns
 - Profiled feed-horns
 - Multimode smooth-walled feed-horns
 - Corrugated profiled feed-horns

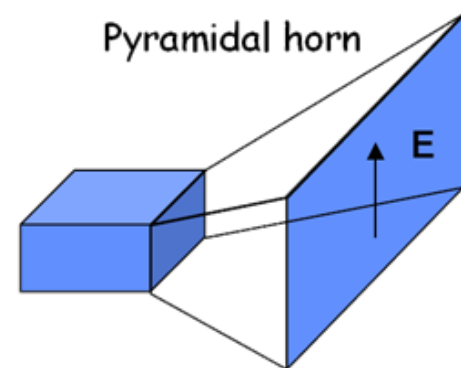
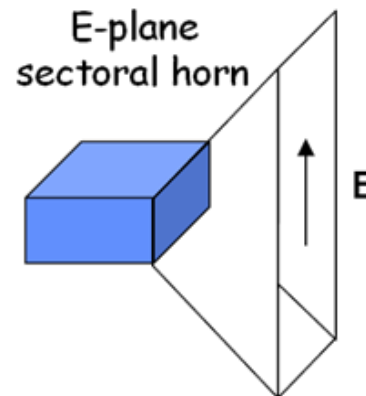
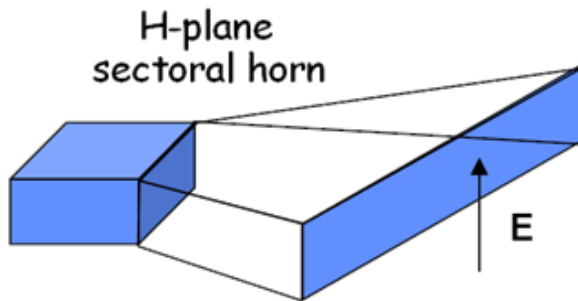
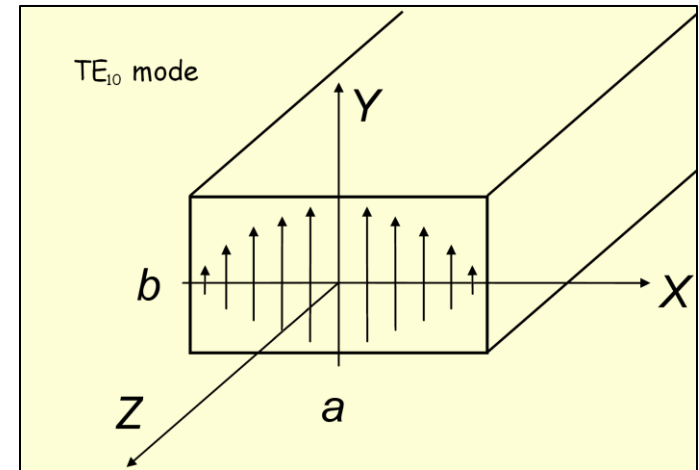


Rectangular feed-horns

Rectangular feed-horns

- Rectangular feed-horns are fed with the fundamental TE_{10} mode of a rectangular metallic waveguide:

$$\vec{E}(x, y, z) = E_0 \cdot \cos\left(\frac{\pi \cdot x}{a}\right) \cdot \vec{u}_y$$





H-plane sectoral feed-horns

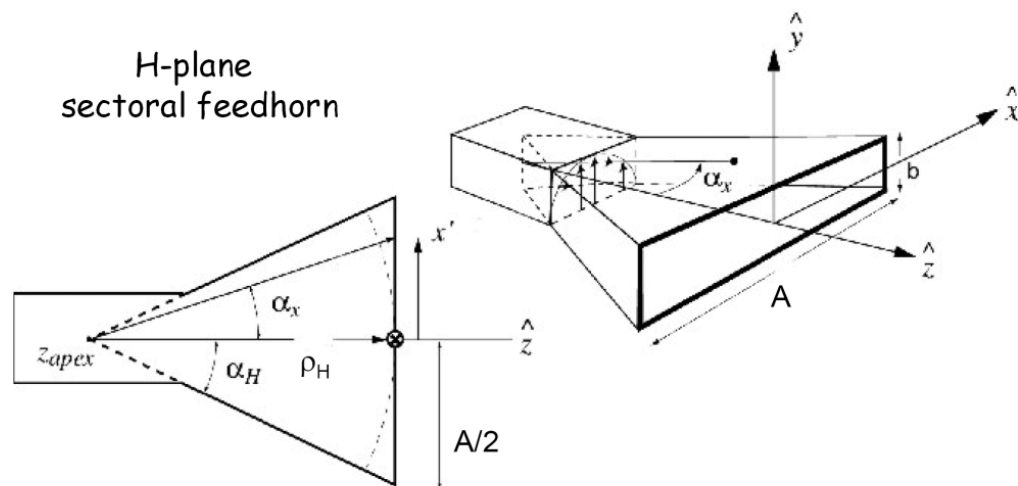
Rectangular feed-horns

- An H-plane sectoral feed-horn is the one where the wider dimension of the input waveguide is broadened remaining constant the other dimension
- The aperture field is affected by a phase error in the H-plane:

$$\vec{E}(x, y) = E_0 \cdot \cos\left(\frac{\pi \cdot x}{A}\right) \cdot e^{-j \cdot 8 \cdot \pi \cdot t \cdot \frac{x^2}{A^2}} \cdot \vec{u}_y$$

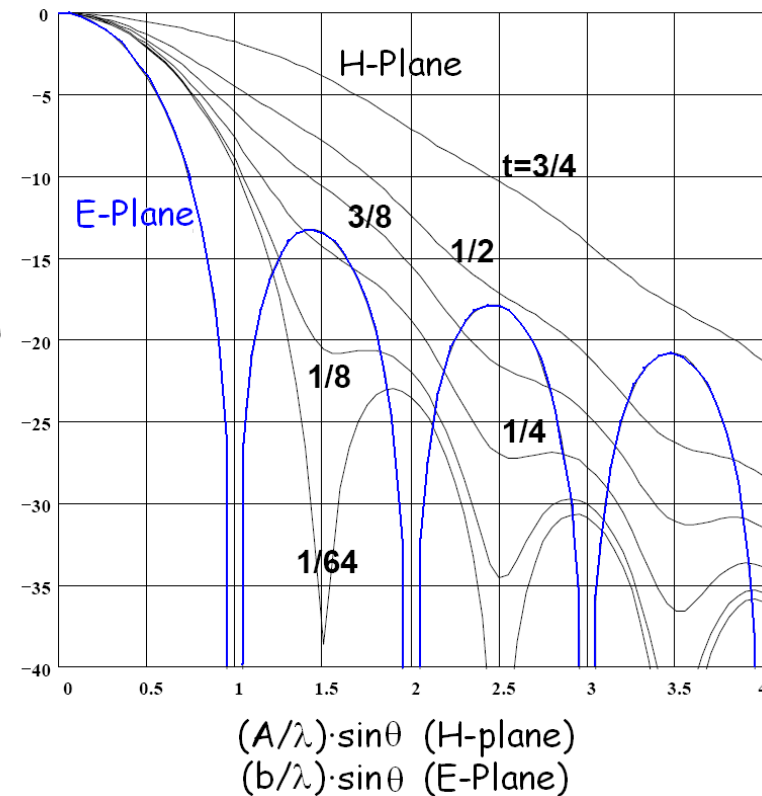
where t is the phase error at the aperture (in turns),

$$t = \frac{A^2}{8 \cdot \lambda \cdot \rho_H}$$



Rectangular feed-horns

- The universal radiation patterns for the H-plane sectoral feed-horn are represented with the phase error as parameter for H-plane
- E-plane radiation pattern is just a “Sinc” as a result of applying the Fourier transform to a pulse
- Phase error elevates the side-lobe level filling the nulls between adjacent side-lobes



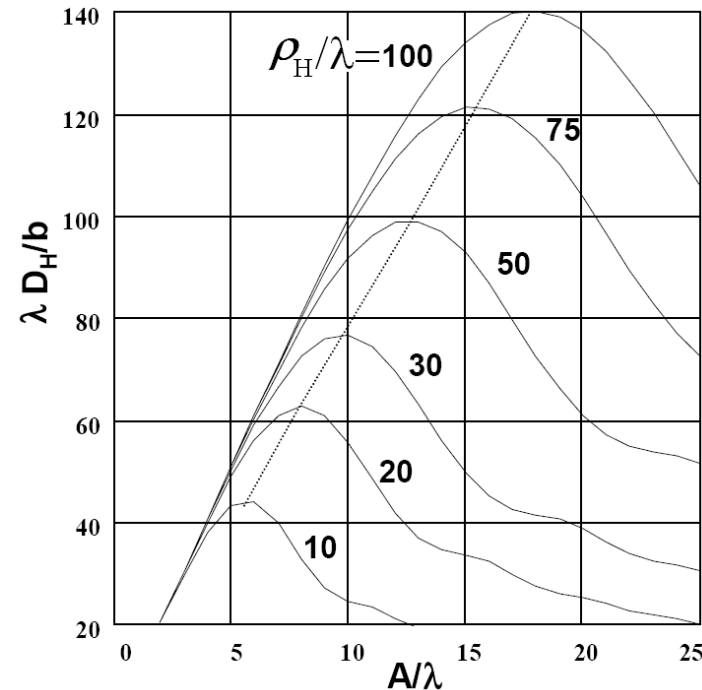
Rectangular feed-horns

- The directivity, (D_H), of the H-plane sectoral horn is obtained integrating the power at the aperture
- For every ρ_H value, there is an optimum aperture width (A) that gives maximum directivity
- Optimum aperture satisfy the equation:

$$A = \sqrt{3 \cdot \lambda \cdot \rho_H}$$

these optimum feed-horns
present a phase error of:

$$t_{opt} = \frac{A^2}{8 \cdot \lambda \cdot \rho_H} = \frac{3}{8}$$





E-plane sectoral feed-horns

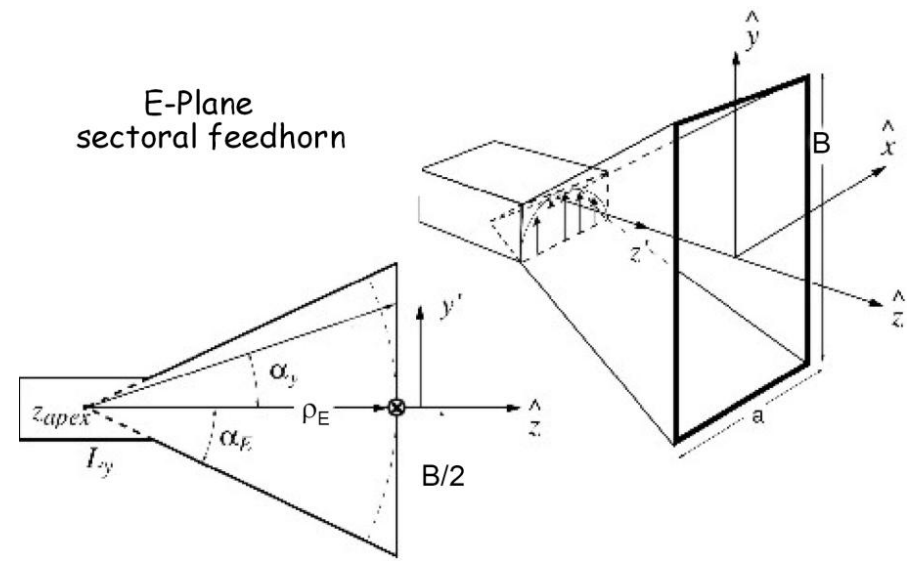
Rectangular feed-horns

- An E-plane sectoral feed-horn is the one where the narrower dimension of the input waveguide is broadened remaining constant the other dimension
- The aperture field is affected by a phase error in the E-plane

$$\vec{E}(x, y) = E_0 \cdot \cos\left(\frac{\pi \cdot x}{A}\right) \cdot e^{-j \cdot 8 \cdot \pi \cdot s \cdot \frac{y^2}{B^2}} \cdot \vec{u}_y$$

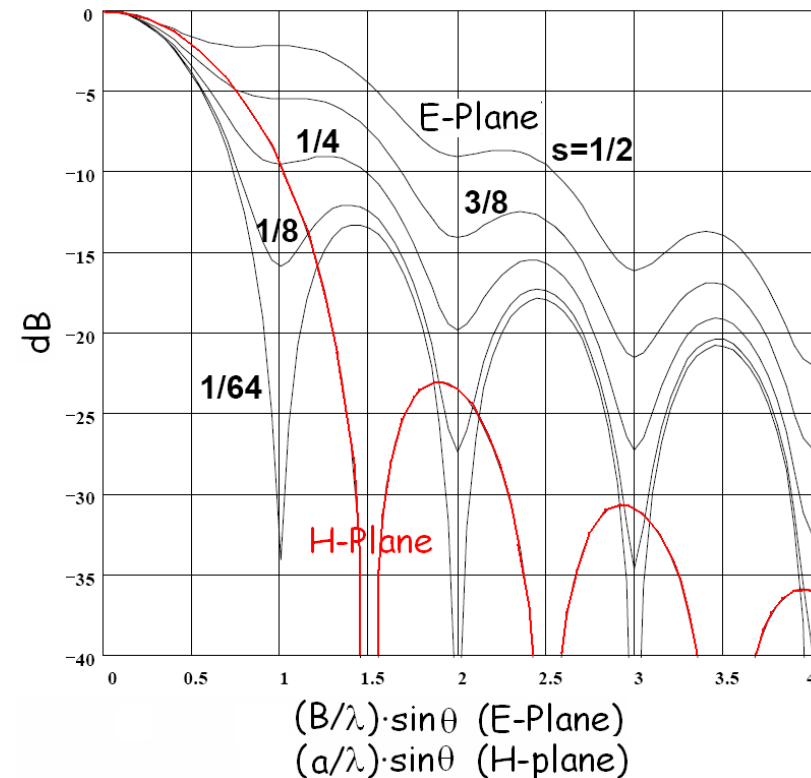
where s is the phase error expressed in turns:

$$s = \frac{B^2}{8 \cdot \lambda \cdot \rho_E}$$



Rectangular feed-horns

- The universal radiation patterns for the E-plane sectoral feed-horn are represented with the phase error as parameter for E-plane
- H-plane radiation pattern results by applying Fourier transform to a cosine windowed pulse (lower side-lobes than a Sinc function)
- Phase error elevates the side-lobe level filling the null between adjacent side-lobes



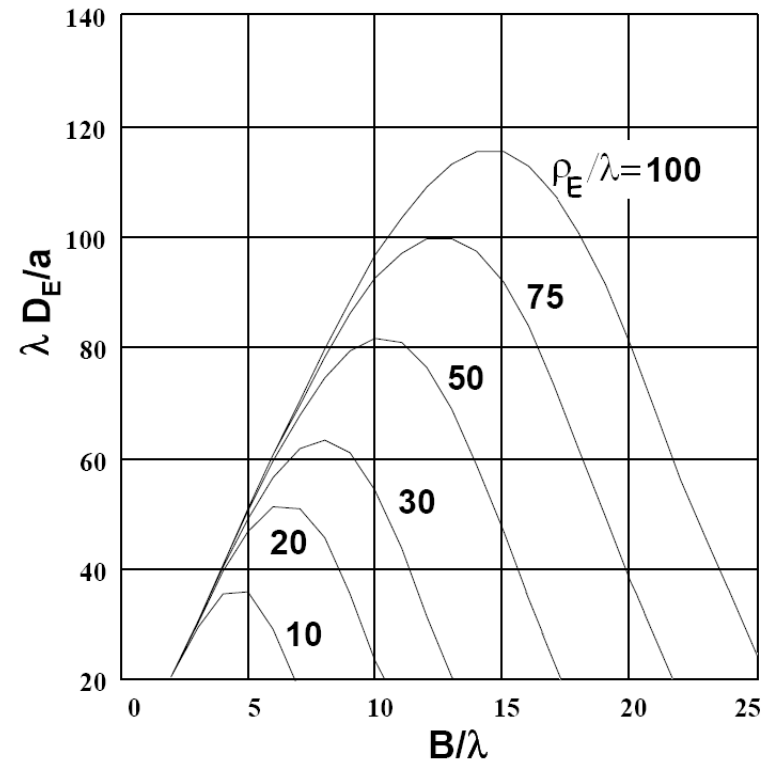
Rectangular feed-horns

- The directivity, (D_E), of the E-plane sectoral horn is obtained integrating the power at the aperture
- For every ρ_E value, there is an optimum aperture width (B) that gives maximum directivity
- Optimum aperture satisfy

the equation: $B = \sqrt{2 \cdot \lambda \cdot \rho_E}$

- These optimum feed-horns present a phase error of:

$$s_{opt} = \frac{B^2}{8 \cdot \lambda \cdot \rho_E} = \frac{1}{4}$$





Pyramidal feed-horns

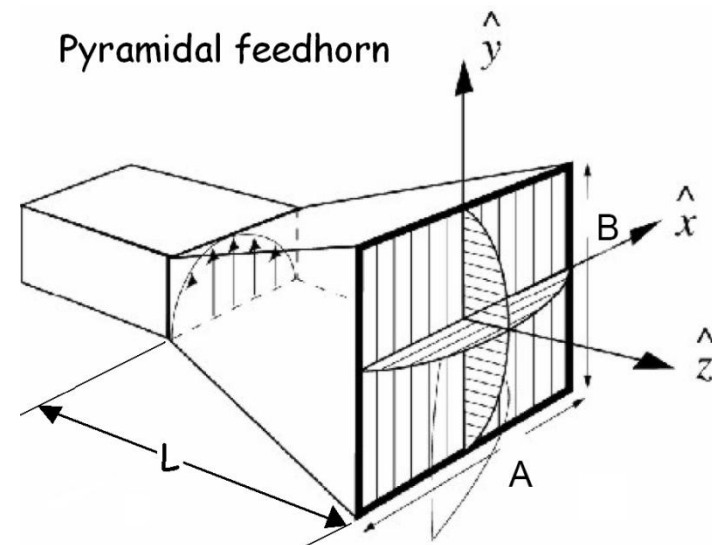
Rectangular feed-horns

- An pyramidal feed-horn is the one where both dimensions of the input waveguide are broadened
- The aperture field is affected by a phase error in both planes

$$\vec{E}(x, y) = E_0 \cdot \cos\left(\frac{\pi \cdot x}{A}\right) \cdot e^{-j \cdot 8 \cdot \pi \cdot t \cdot \frac{x^2}{A^2}} \cdot e^{-j \cdot 8 \cdot \pi \cdot s \cdot \frac{y^2}{B^2}} \cdot \vec{u}_y$$

- These feed-horns must satisfy geometrical condition:

$$\rho_H \cdot \left(1 - \frac{a}{A}\right) = \rho_E \cdot \left(1 - \frac{b}{B}\right) = L$$



Rectangular feed-horns

- The radiation pattern of a pyramidal horn can be directly obtained from the universal radiation patterns of sectoral E and H-plane feed-horns
- The directivity of a pyramidal horn can be expressed as a combination of the directivities of the sectoral feed-horns:

$$D_p = \frac{\pi}{32} \cdot \left(\frac{\lambda}{A} \cdot D_E \right) \cdot \left(\frac{\lambda}{B} \cdot D_H \right)$$

where the terms between parentheses can be obtained directly from the curves of directivity of sectoral feed-horns changing a with A and b with B

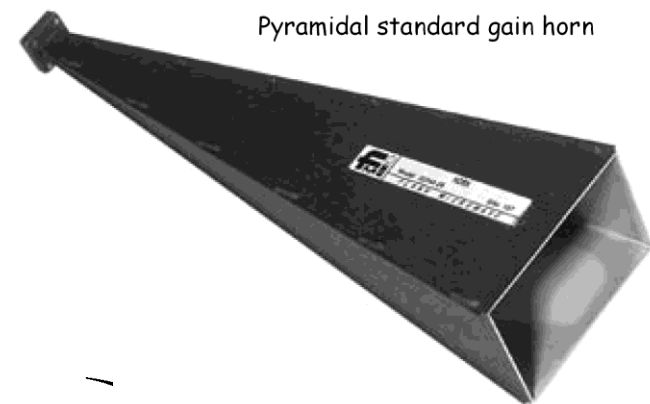
Rectangular feed-horns

- Pyramidal feed-horns are usually used as standard gain horns serving as reference for gain measurements
- Pyramidal standard gain horns are usually designed under the optimum condition:

$$A = \sqrt{3 \cdot \lambda \cdot \rho_H} \quad B = \sqrt{2 \cdot \lambda \cdot \rho_E}$$

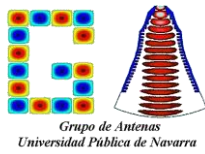
- The effective aperture of these pyramidal feed-horns is around 50% of its physical size, so:

$$G = \varepsilon_{ap} \cdot \frac{4 \cdot \pi}{\lambda^2} \cdot A_p = \frac{1}{2} \cdot \frac{4 \cdot \pi}{\lambda^2} \cdot A \cdot B$$



Pyramidal standard gain horn

Rectangular feed-horns



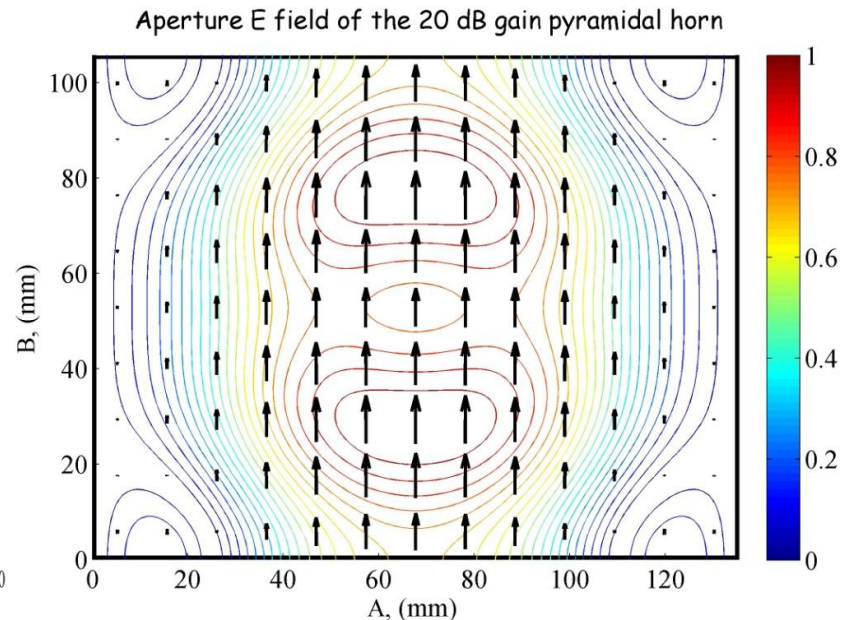
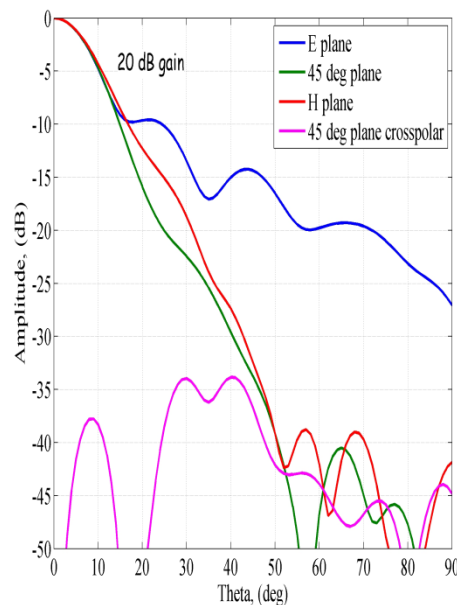
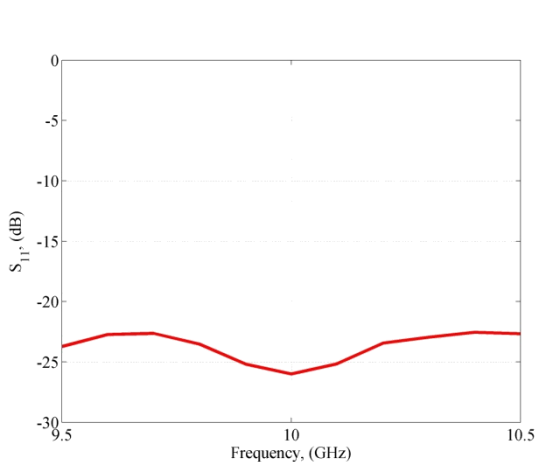
- Example: Design a 20 dB gain pyramidal horn at 10 GHz under optimum condition with equal E and H planes beam-width and with WR-90 standard waveguide as feeding waveguide
 - First we calculate the $A \cdot B$ area of the aperture by using approximate gain formula that relates gain and aperture surface:

$$A \cdot B = 14323.94 \text{ mm}^2$$

- For a metallic horn antenna with no dielectric material inside, gain and directivity can be considered to be the same
- WR-90 standard waveguide has $a=23$ mm and $b=10$ mm as internal dimensions

Rectangular feed-horns

- If we calculate for optimum condition the dimensions of the pyramidal horn satisfying its geometrical condition we obtain the following dimensions:
- $A = 135.4$ mm, $B = 105.8$ mm, $\rho_H = 203.6$ mm, $\rho_E = 186.6$ mm
- Resulting in a total length of $L = 169$ mm

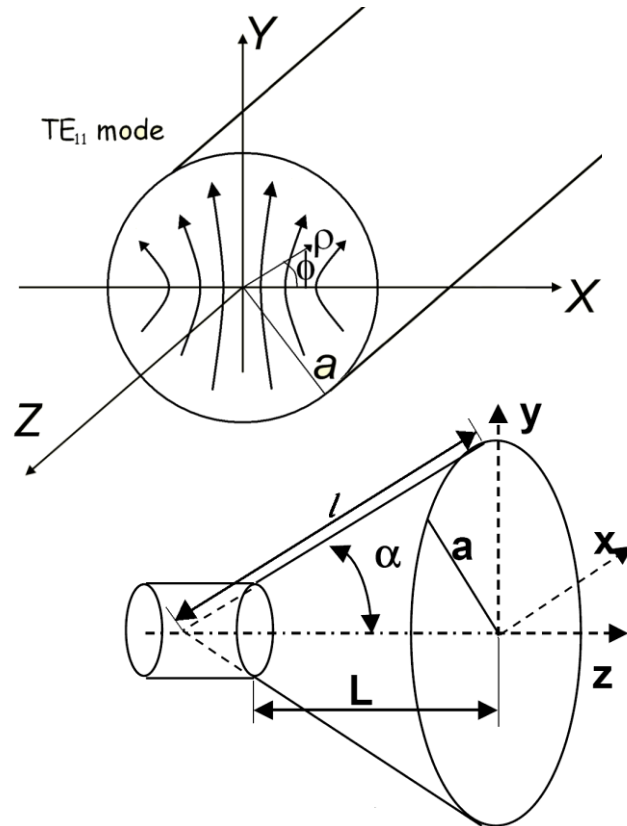




Conical feed-horns

Conical feed-horns

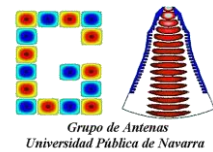
- Conical feed-horns are fed with a metallic circular waveguide in mono-mode operation with the fundamental TE_{11} mode:



- The main inconvenient of TE_{11} mode is that it doesn't present polarization purity

$$E_{\rho} = E_0 \cdot \frac{1}{\rho} \cdot J_1 \left(1.841 \cdot \frac{\rho}{a} \right) \cdot \sin \phi$$

$$E_{\phi} = E_0 \cdot \frac{1}{\rho} \cdot \frac{\partial}{\partial \rho} \left(J_1 \left(1.841 \cdot \frac{\rho}{a} \right) \right) \cdot \cos \phi$$

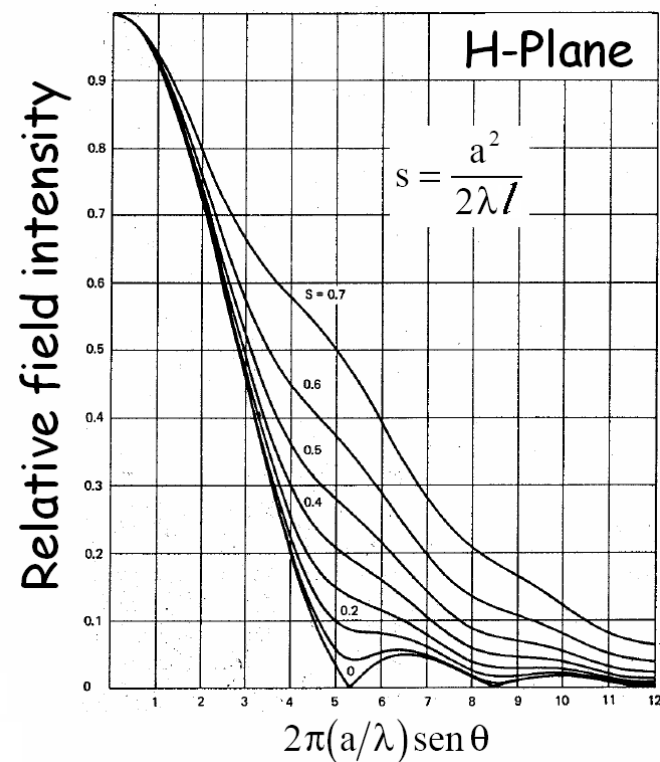
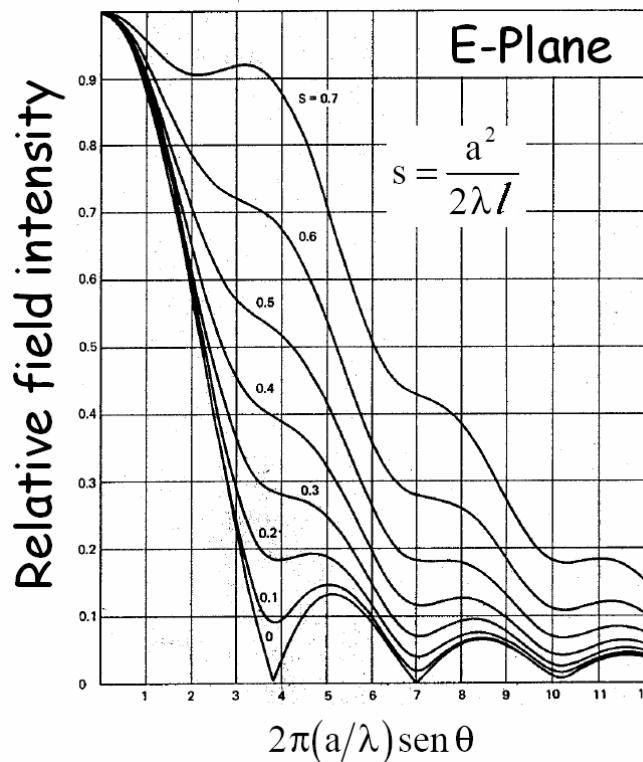


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Smooth-walled conical feed-horns

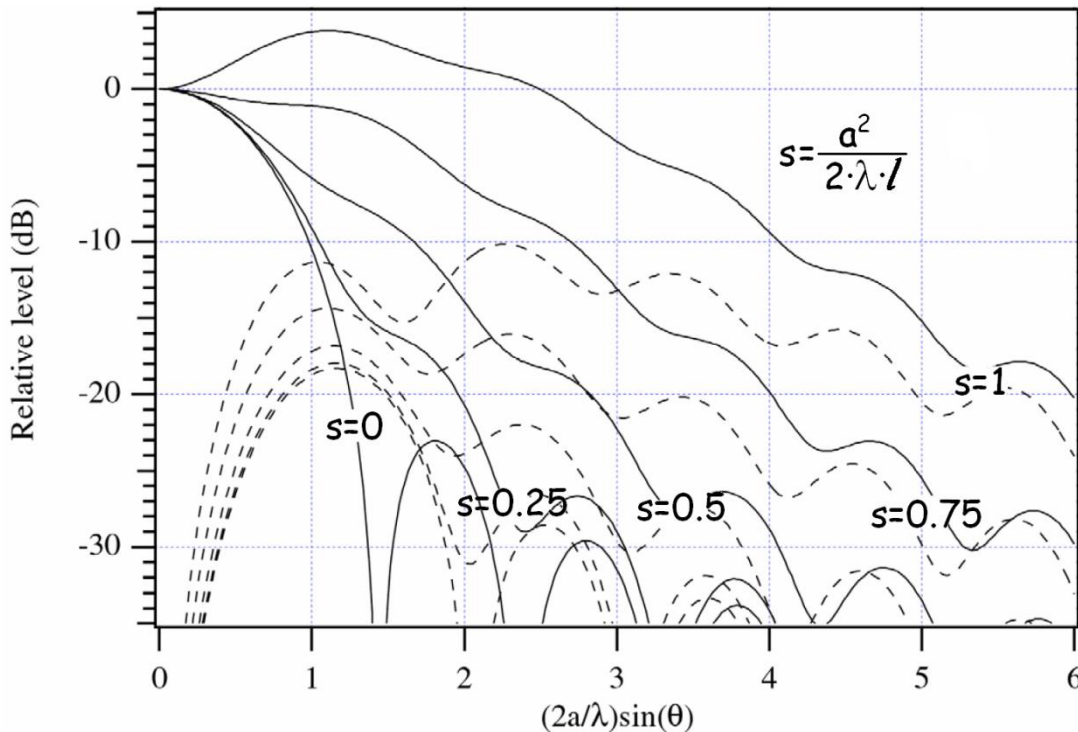
Conical feed-horns

- The universal radiation patterns for the conical feed-horn are represented with the phase error as parameter for both planes:



Conical feed-horns

- Also, the universal radiation pattern for the conical feed-horn for $\phi=45$ deg with the phase error as parameter is represented
- Cross-polar level for $\phi=45$ deg is also represented



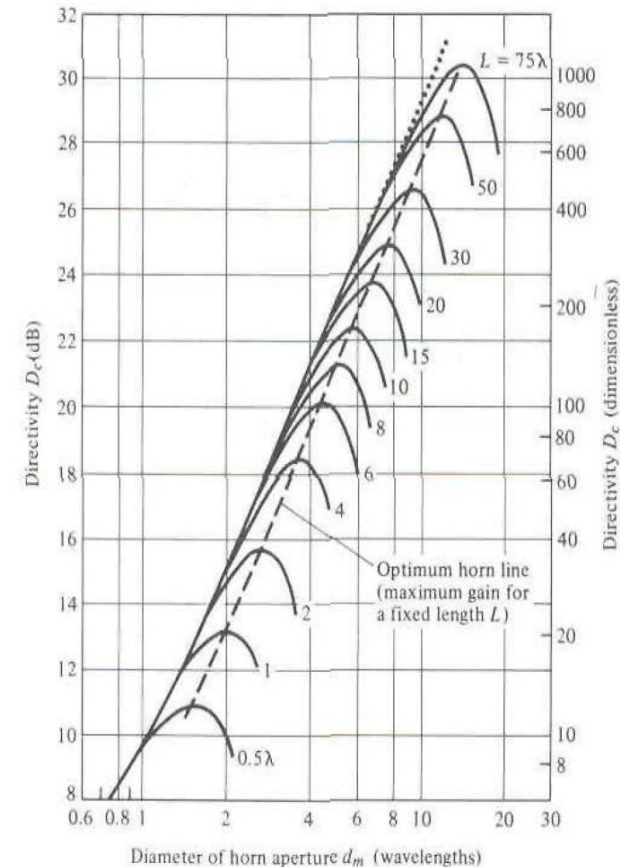
Conical feed-horns

- The directivity of the conical feed-horn is obtained integrating the power at the aperture
- For every length, (L), there is an optimum aperture diameter ($d_m = 2 \cdot a$) that gives maximum directivity
- Optimum aperture satisfy the equation:

$$d_m = 2 \cdot a \approx \sqrt{3 \cdot l \cdot \lambda}$$

- These optimum feed-horns present a phase error of $s=3/8$.
- The effective aperture of these pyramidal feed-horns is also around 50% of its physical size, so:

$$G = \varepsilon_{ap} \cdot \frac{4 \cdot \pi}{\lambda^2} \cdot A_p = \frac{1}{2} \cdot \frac{4 \cdot \pi}{\lambda^2} \cdot \pi \cdot a^2$$



Conical feed-horns

- Example: Design a 20 dB gain conical feed-horn at 10 GHz under optimum condition with equal E and H planes beam-width and with $a=11.7$ mm circular waveguide radius as feeding waveguide.
 - First we calculate the radius of the aperture by using approximate gain formula that relates gain and aperture surface:

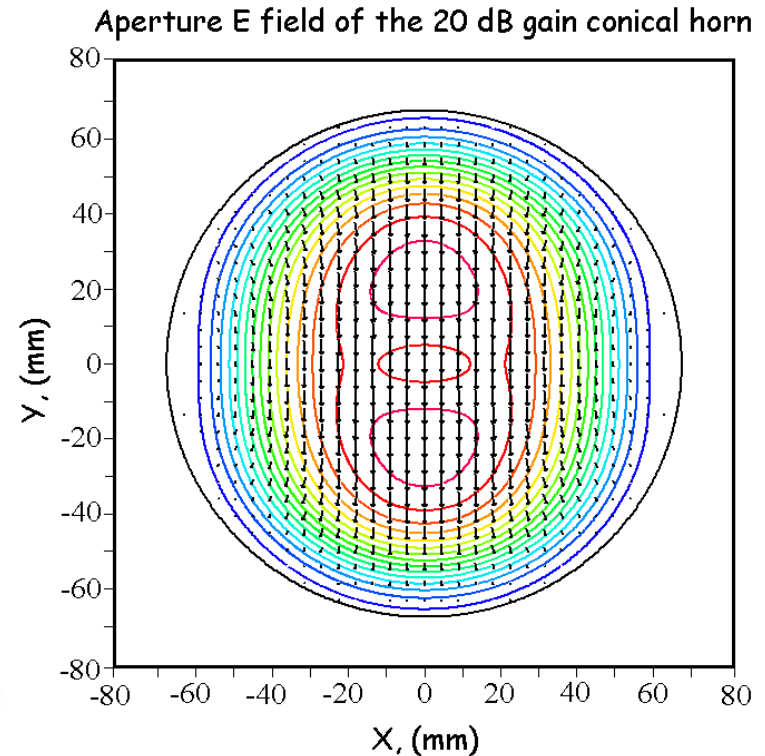
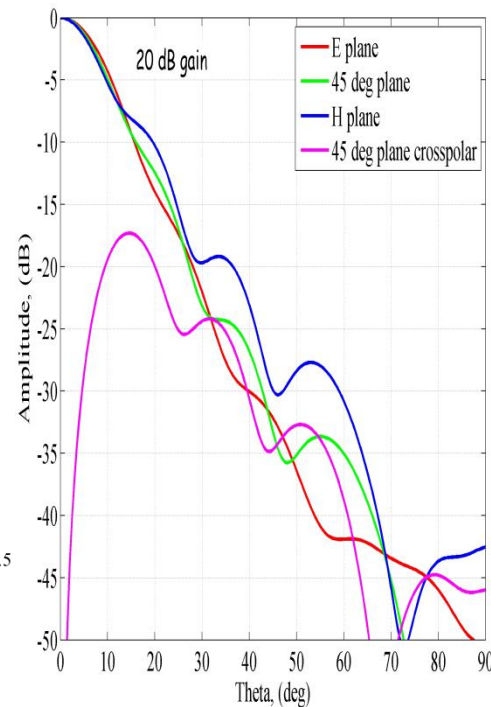
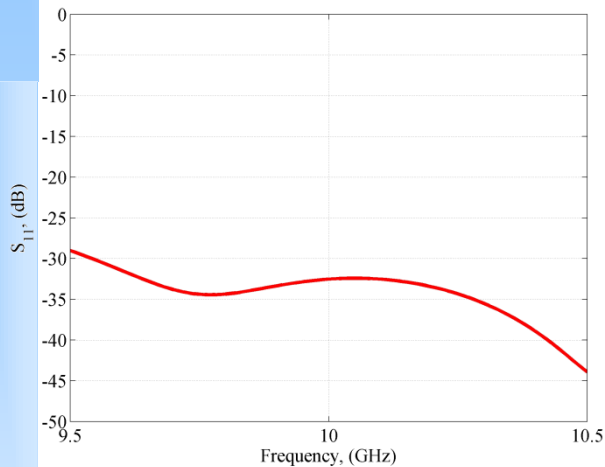
$$a = 67.5 \text{ mm}$$

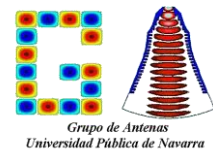
- For a metallic horn antenna with no dielectric material inside, gain and directivity can be considered to be the same.
- Now, we can calculate for optimum condition the error parameters, l , of the conical feed-horn:

$$l = 202.6 \text{ mm}$$

Conical feed-horns

- The rest of physical dimensions are easily calculated using trigonometric relations:
- $a = 67.5$ mm, $l = 202.6$ mm, $\alpha = 19.5^\circ$, $L = 158$ mm

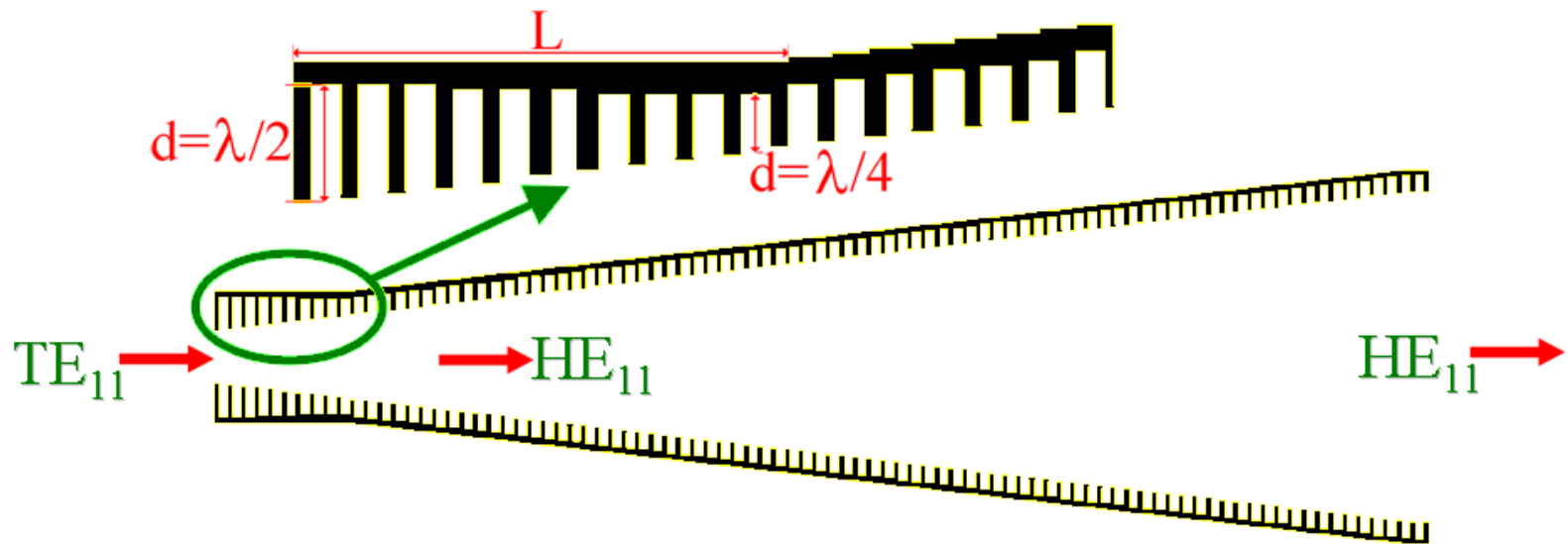




Conical corrugated feed-horns

Conical feed-horns

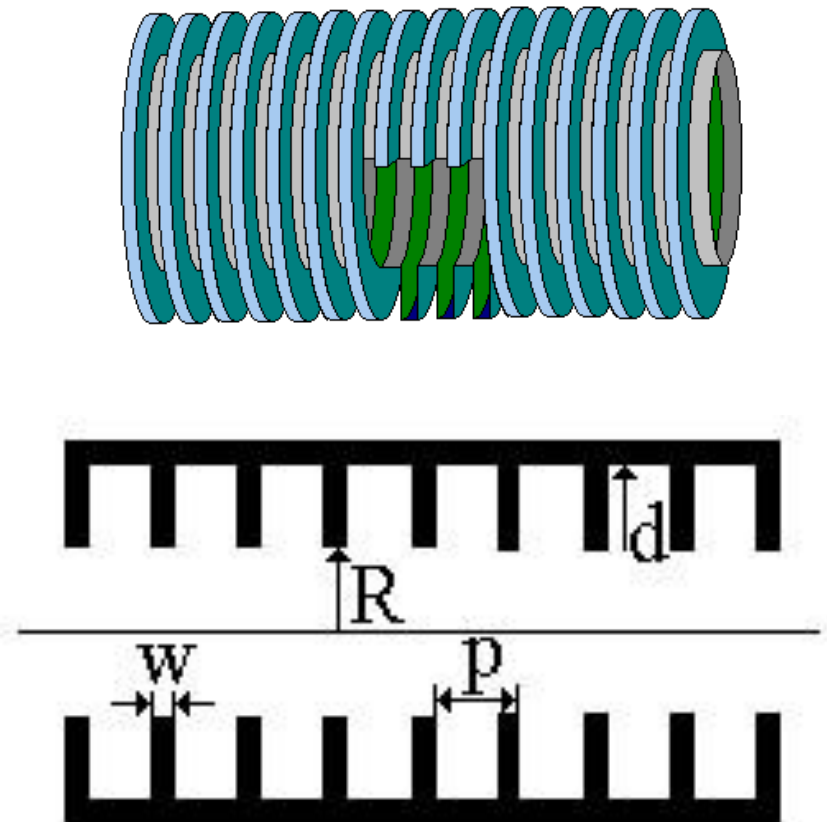
- Smooth-walled conical feed-horns present the main problem of the high cross-polar level
- Conical corrugated feed-horns overcome this problem:



Conical feed-horns

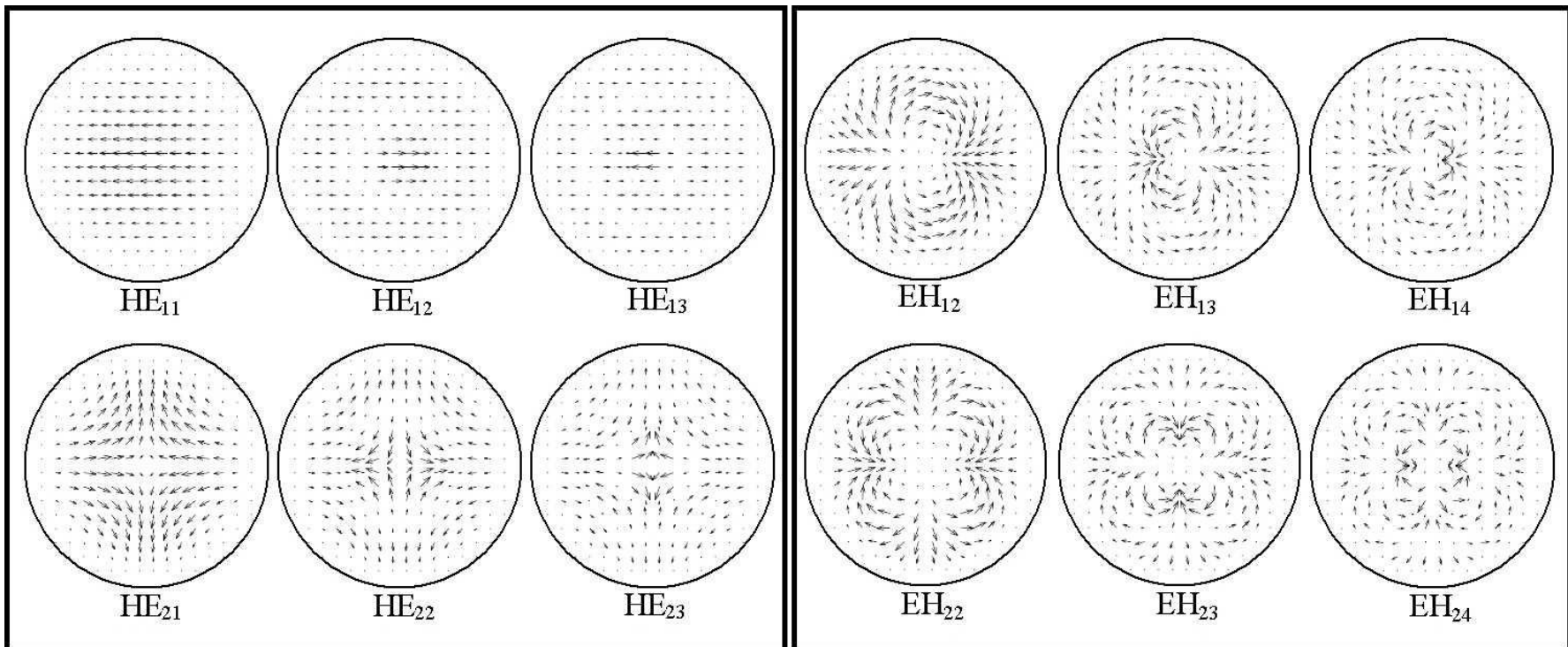
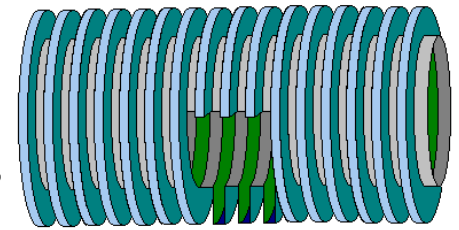
- A corrugated waveguide is defined by a series of smooth waveguides with different diameters (periodical structure):

- R defines the inner diameter
- d the corrugation depth
- Corrugation pitch is defined by p (period) and w (width of corrugation)



Conical feed-horns

- A corrugated waveguide can be defined by the TE_{mn} and TM_{mn} smooth circular mode family at each position, or in terms of HE_{mn} and EH_{mn} hybrid modes



Conical feed-horns

- The condition of zero cross-polarized field $E_y = 0$ can be obtained if:

$$p < \lambda/3$$

&

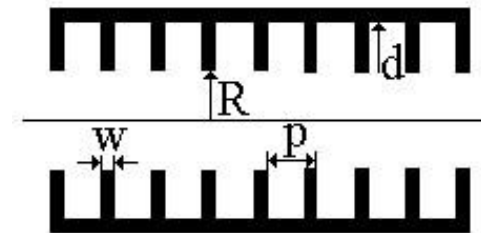
$$d \approx \lambda/4$$



Balanced

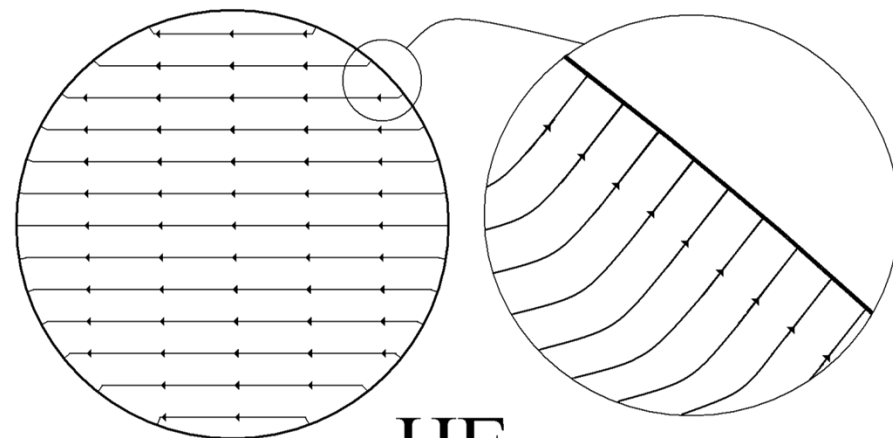
Hybrid

Condition



- The fundamental mode of a corrugated waveguide (HE_{11}) at an inner diameter can be expressed as:

$$\left. \begin{aligned} E_x(r, \phi) &= A_0 J_0 \left(\frac{2.4808 r}{R} \right) \\ E_y(r, \phi) &= 0 \end{aligned} \right\}$$

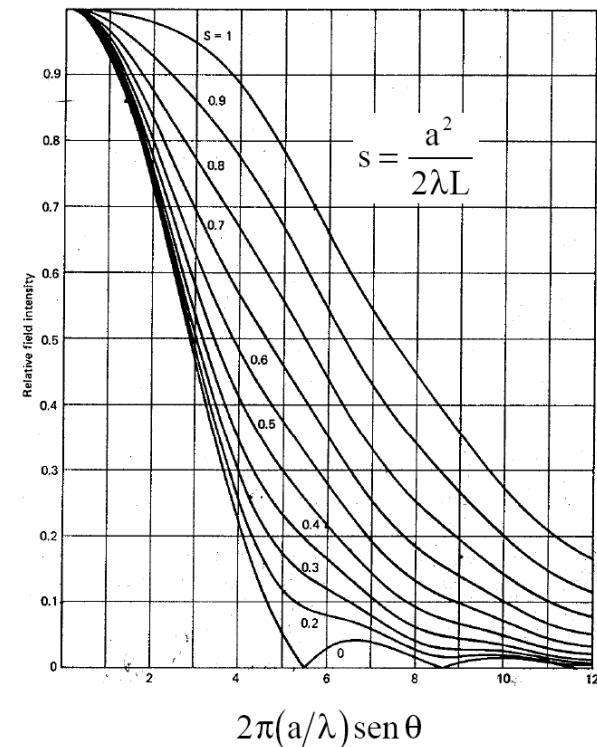
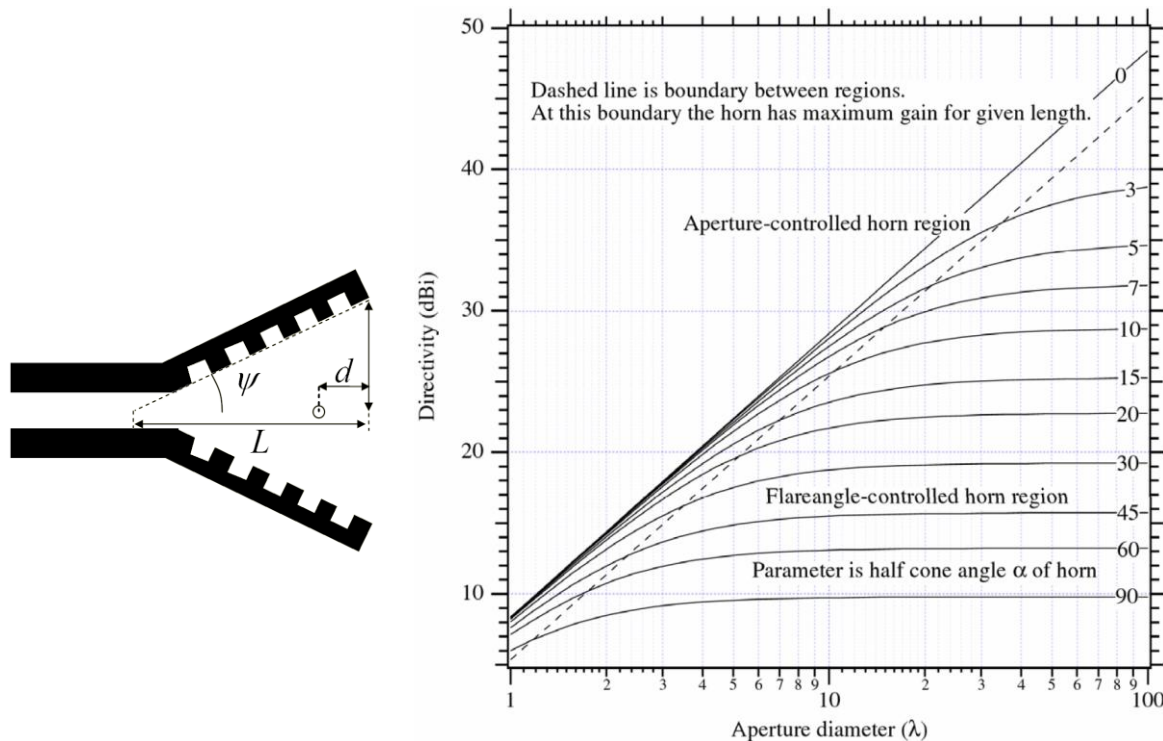


HE_{11}

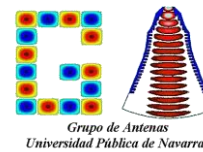


Conical feed-horn

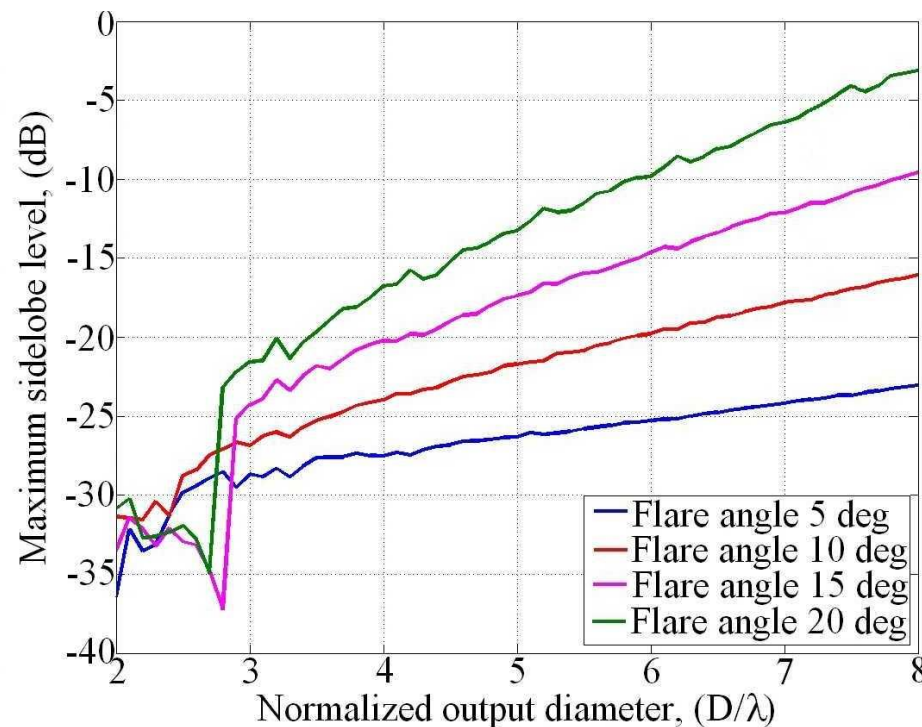
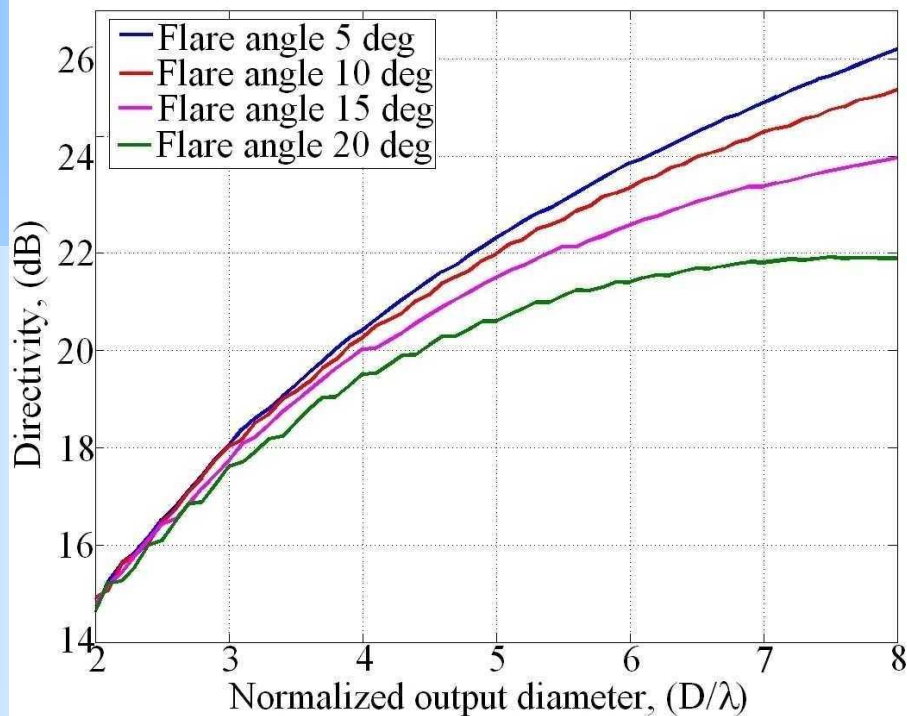
- The universal radiation patterns for the conical corrugated feed-horn are represented with the phase error as parameter (the radiation patterns are rotationally symmetric, independent of ϕ plane):



Conical feed-horns



- For small flare angles these figures can also be used
- In these figures also the side-lobe level can be selected



Conical feed-horns

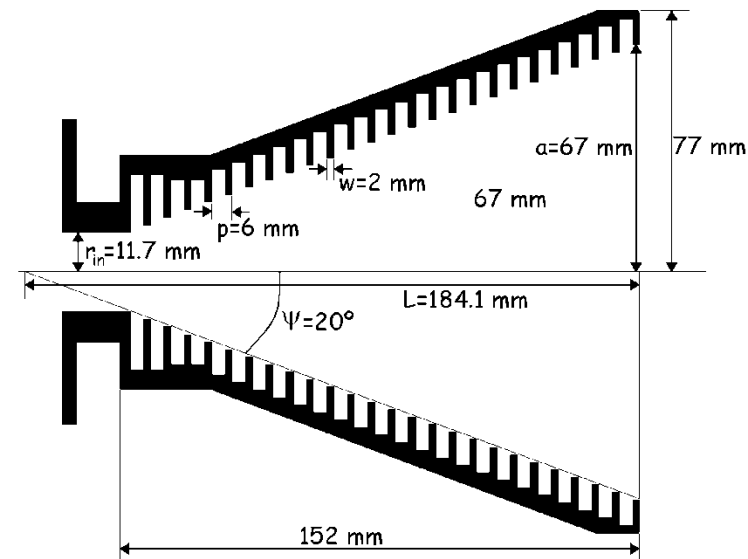
- Example: Design a 20 dB gain conical corrugated feed-horn at 10 GHz under optimum condition with equal E and H planes beam-width and with $a=11.7$ mm circular waveguide radius as feeding waveguide
 - First we calculate the radius of the aperture, from figures:

$$a = \frac{4.5 \cdot \lambda}{2} \approx 67 \text{ mm}$$

- Then we select the flare angle for optimum condition: $\psi = 20 \text{ deg}$
- After that we can select the corrugation parameters:

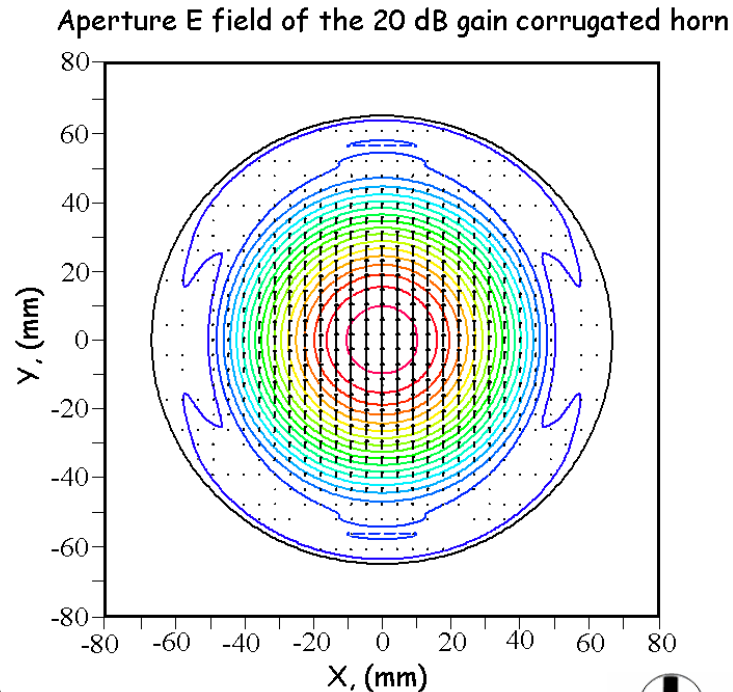
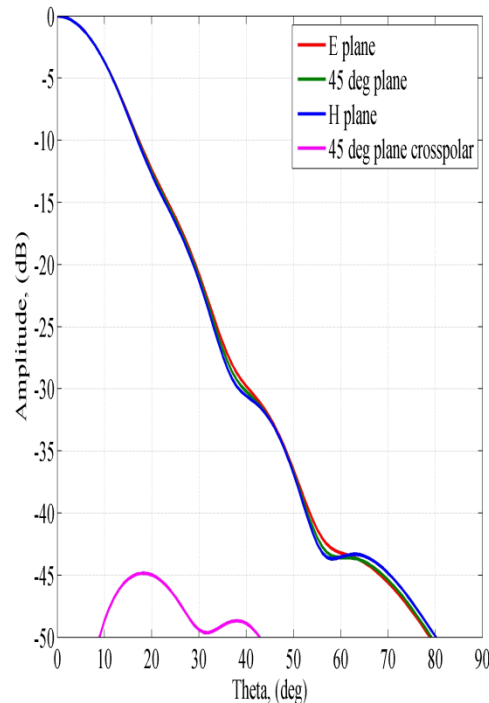
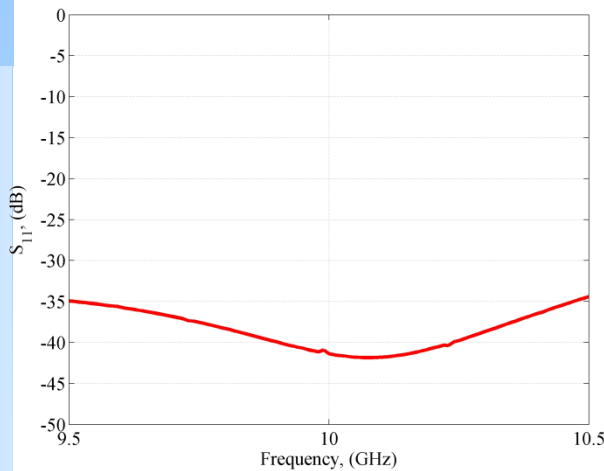
$$p = 6 \text{ mm}, w = 2 \text{ mm}, d = 7.5 \text{ mm}$$

- The feed-horn length of 152 mm

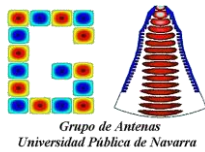


Conical feed-horns

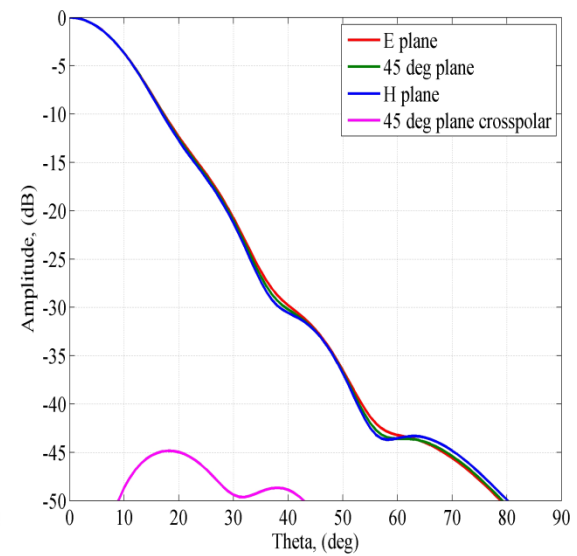
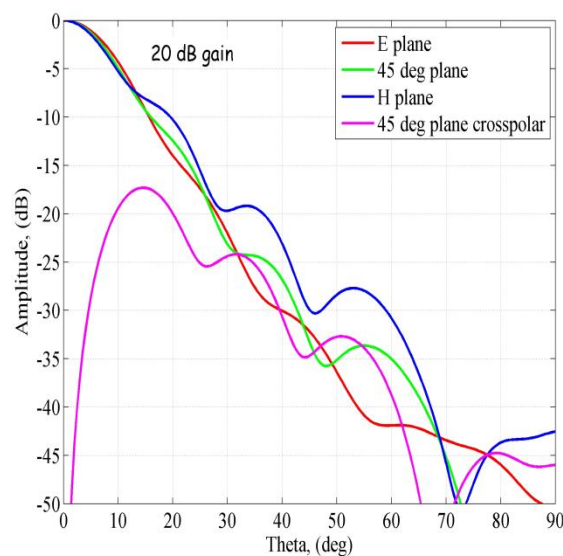
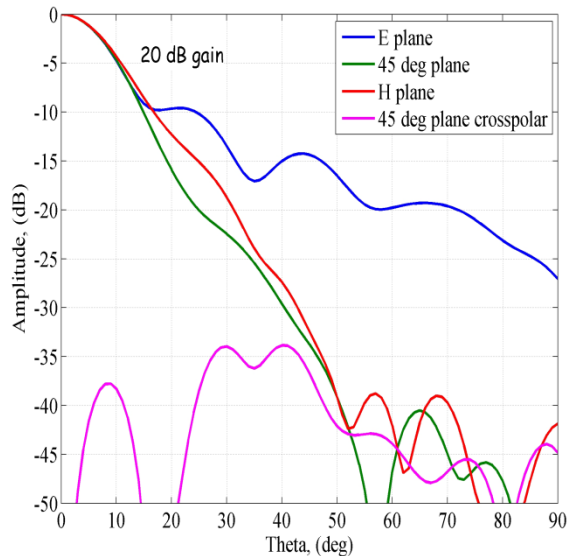
- Conical corrugated feed-horns present a very low cross-polar level, a rotationally symmetric radiation pattern and good matching
- They are also easy to design but more complicated to manufacture than a normal smooth-walled conical horn (a lathe is usually necessary)



Comp.: Pyramidal vs Conical



- Pyramidal feed-horns present much lower cross-polar levels than smooth-walled conical ones, but higher side-lobes (spillover)
- The size of Pyramidal feed-horns is slightly bigger than the size of conical feed-horns
- Conical corrugated horn is the best out of these three models but its weight will be higher

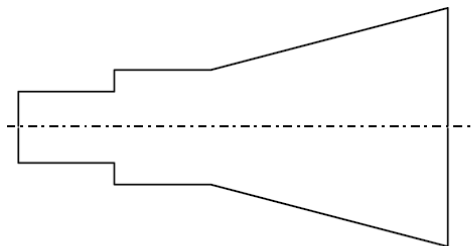




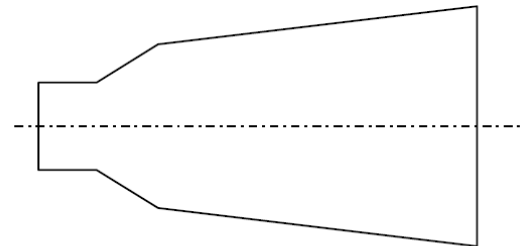
Profiled feed-horns

Profiled feed-horns

- There are other feed-horn types where the profile is not linear.
- The idea of these ones is to make compact designs with nice radiation characteristics.
- There are a lot of models available, in smooth waveguide and in corrugated waveguide, the different between them is that corrugated waveguide models are wide bandwidth and smooth-walled models are narrow band .



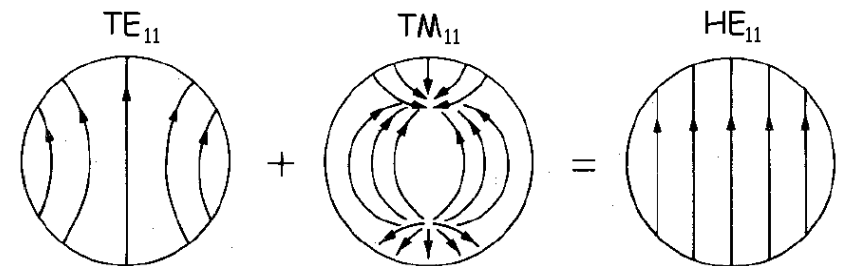
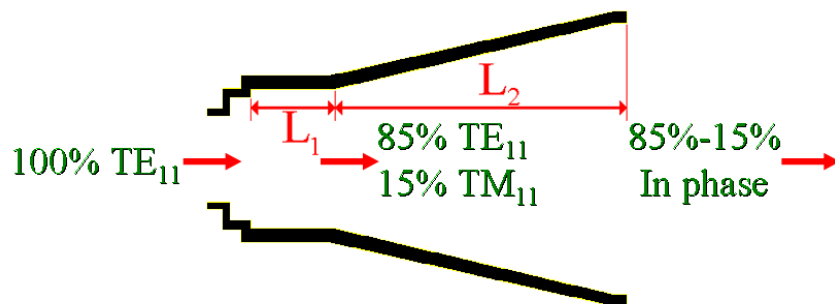
Potter



Turrin

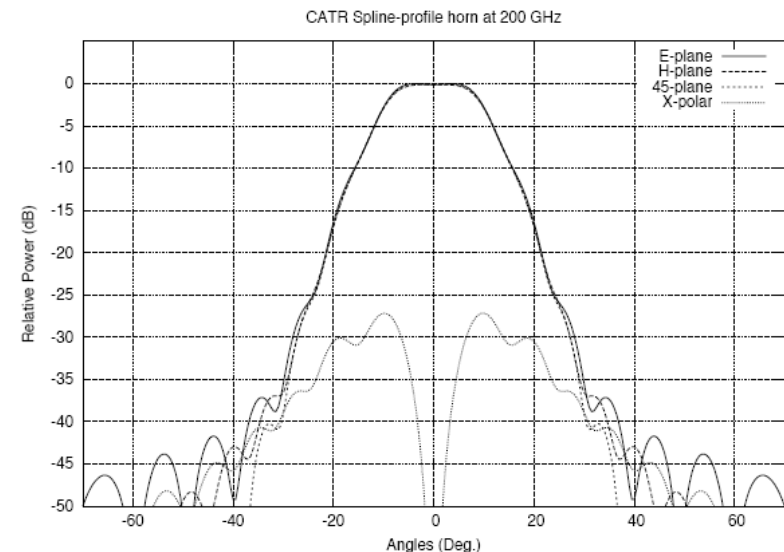
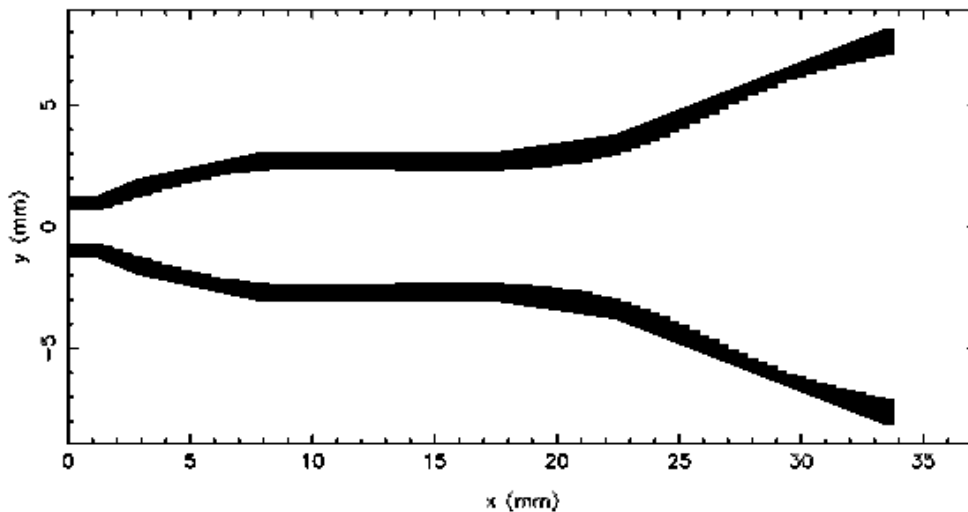
Profiled feed-horns

- The Potter type horn is the best known of multimode horns in smooth-walled waveguide.
- Its performance is very nice in a reduced bandwidth (5 to 8%) approximating to the performance of corrugated horn.
- By means of a proper step or taper in the horn radius, the right amount of TM_{11} (amplitude and phase) is excited.
- A smooth conical taper of a determined length after the step drives the mode mixture to the correct phasing at the aperture.



Profiled feed-horns

- They are other special types of smooth-walled multimode profiled horns.
- In fact, a horn antenna profile can be adjusted with very complicated geometries.
- These feed-horns usually pursue a particular type of radiation pattern.
- Bandwidth of these feed-horns is usually very poor.

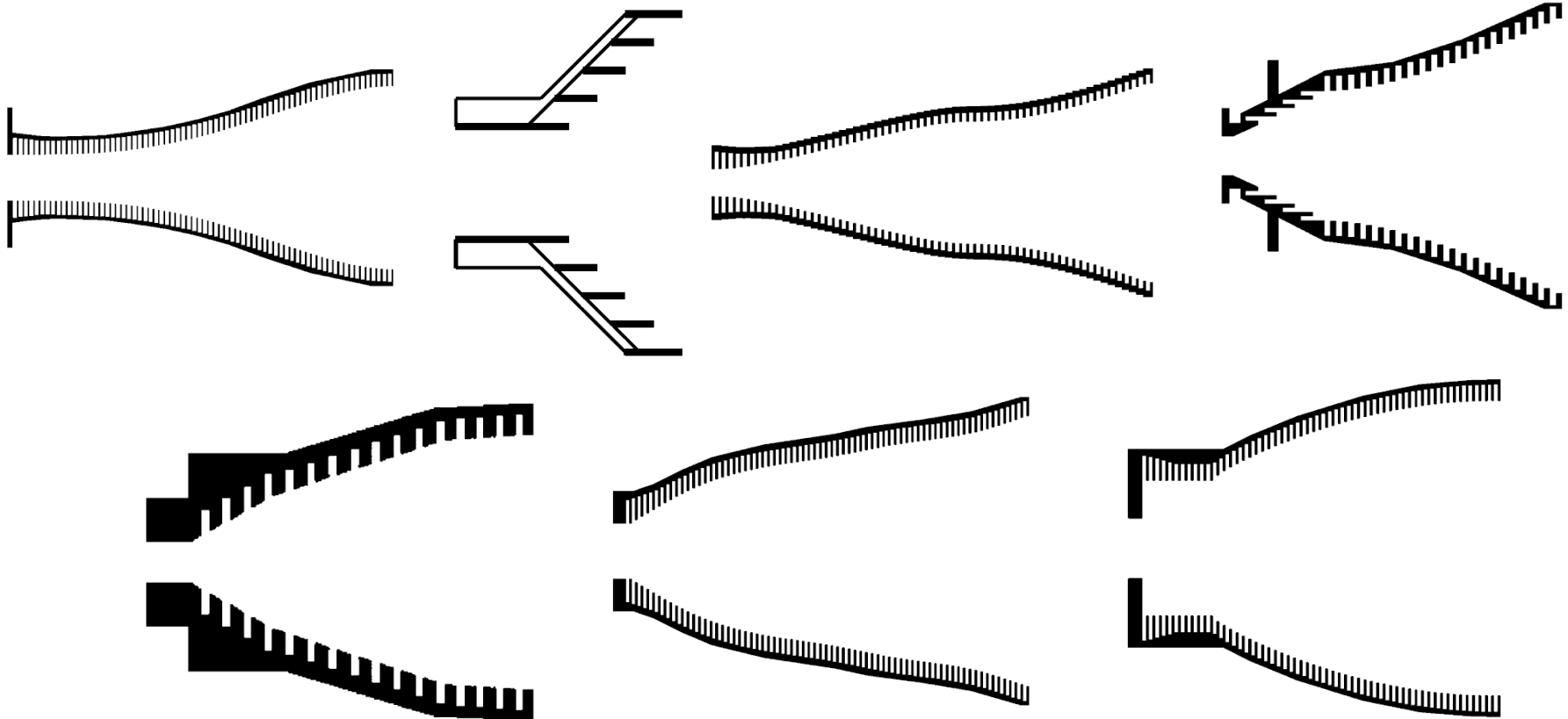




Corrugated profiled feedhorns

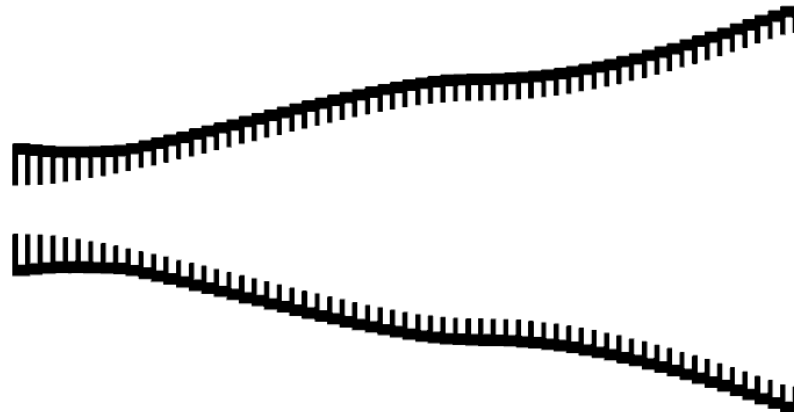
Profiled feed-horns

- There are a lot of models of corrugated profiled feed-horns, in fact, every research group in corrugated horn antennas has their own preferences.



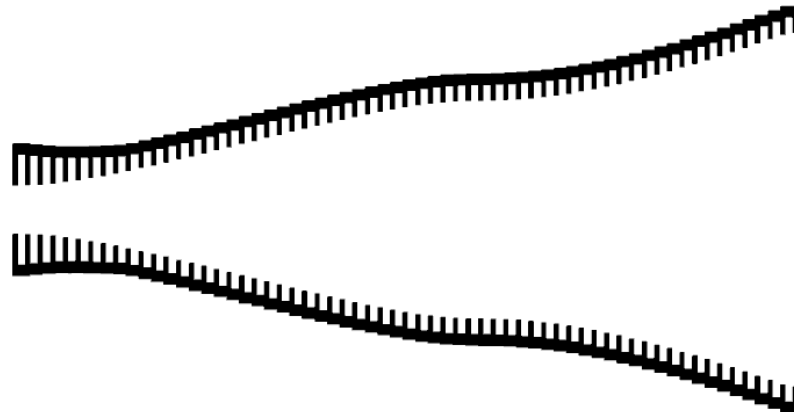
Profiled feed-horns

- One of the best performance profiled corrugated horns is the so called “Gaussian Profiled Horn Antenna” or GPHA
- To overcome the radiation limitations of the HE_{11} mode, a new profile to optimise the matching between the field travelling along the horn antenna and free space was developed.



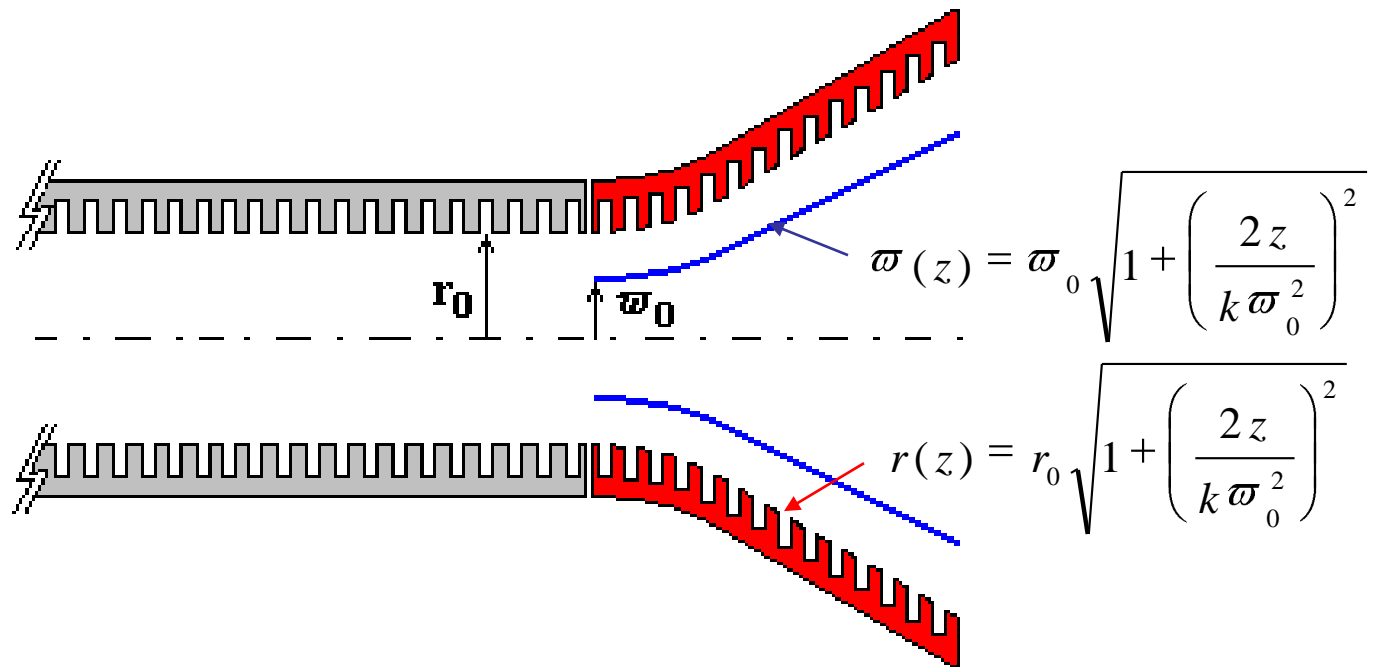
Profiled feed-horns

- GPHA's consist in the addition of a gaussian profiled corrugated part to any waveguide or horn antenna in which the HE_{11} mode is being propagated to improve its radiation characteristics.
- Corrugated GPHA are now-a-days one of the best choices to add at the end of any corrugated horn antenna for high requirement applications where low side-lobes and wide bandwidth are needed.



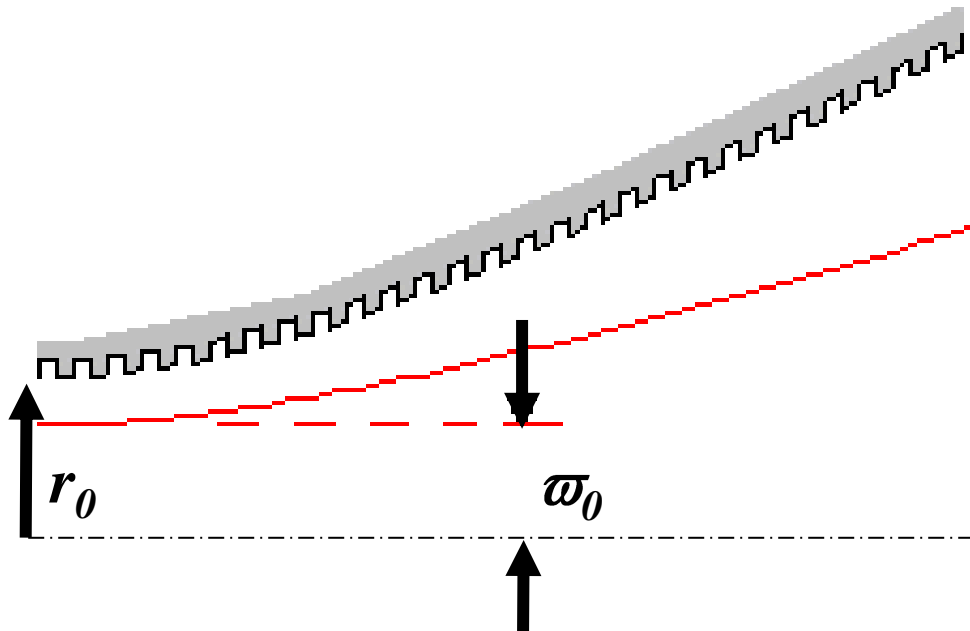
Profiled feed-horns

- The Gaussian Profiled Horn Antenna (GPHA) is defined following the propagation of a Gaussian beam:



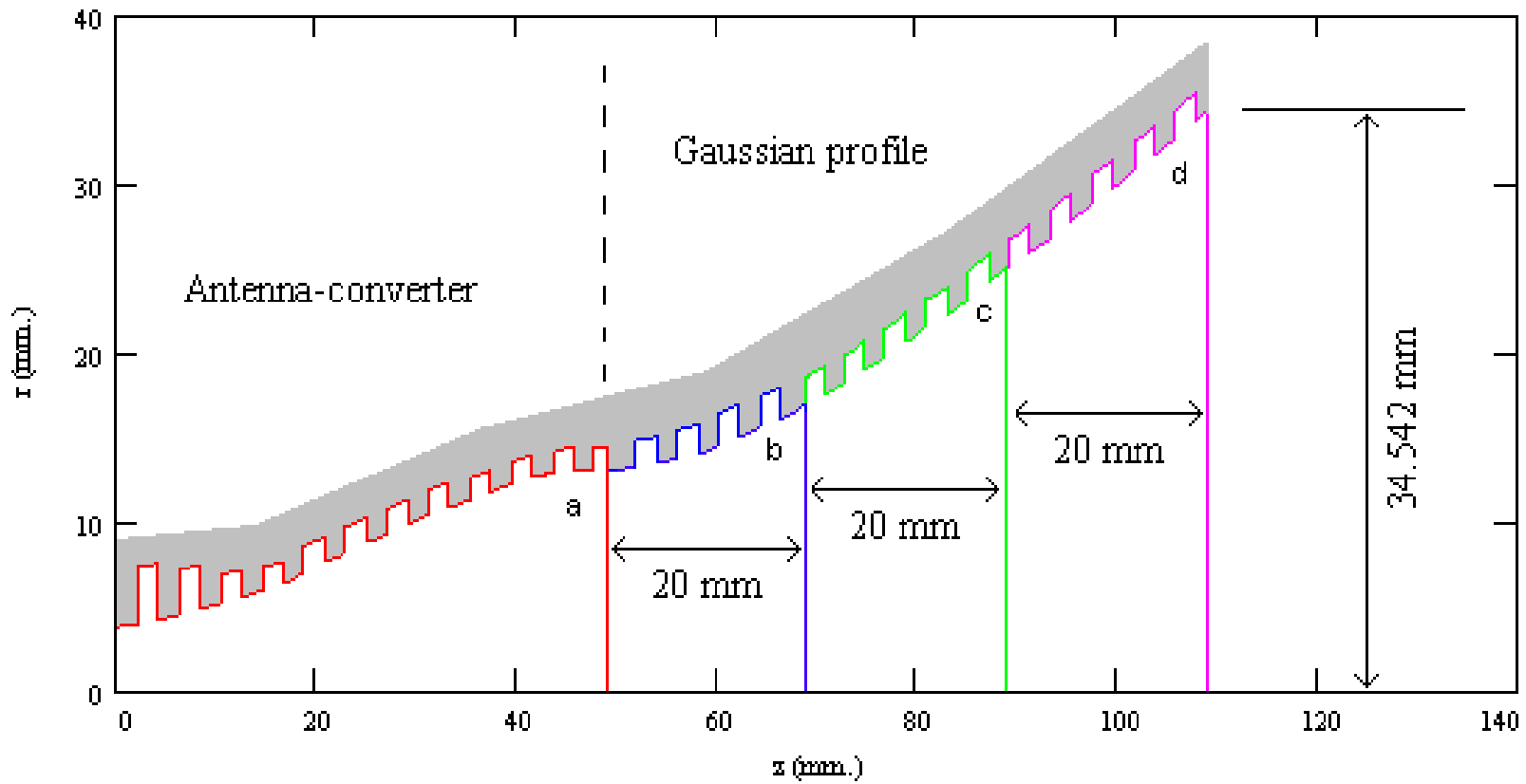
Profiled feed-horns

- To feed the corrugated GPHA, an HE_{11} mode or similar input field is necessary at its input aperture.

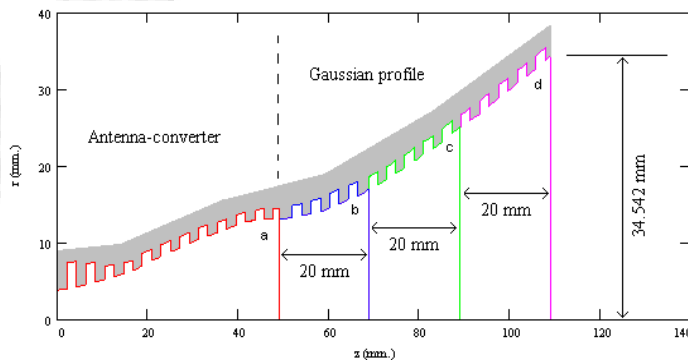


Profiled feed-horns

- Let's do an experiment:



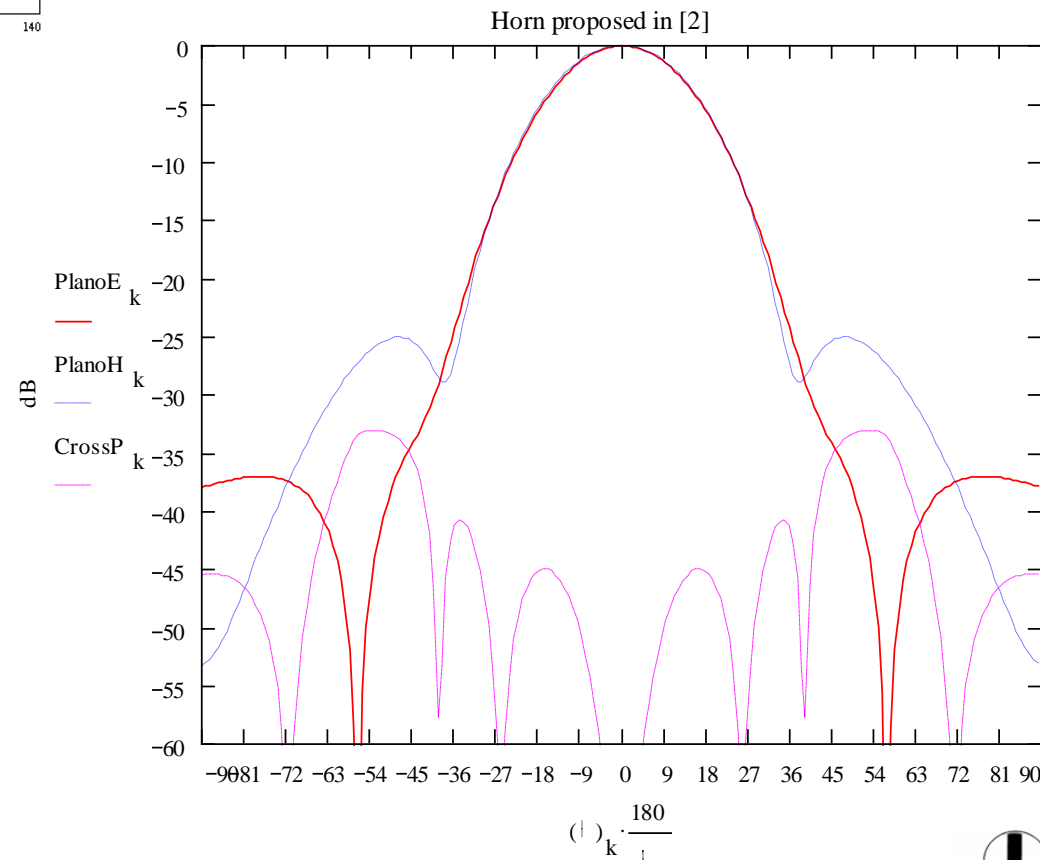
Profiled feed-horns



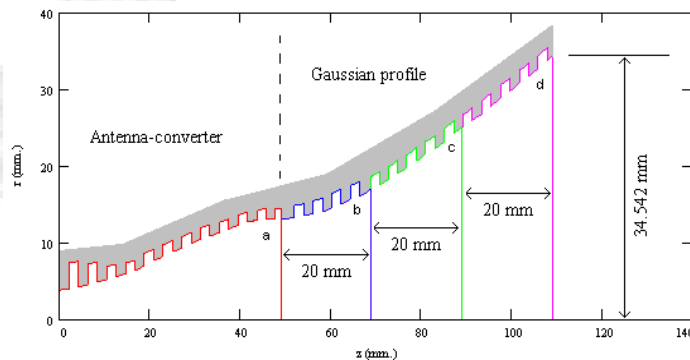
$R_{out}=13.356\text{mm}$
Gain=17.2 dBi
 $\eta_E=97.2\%$

$\omega_0=8.9485\text{mm}$.
 $z_o=L$
 $\eta_H=97.4\%$

Modes	Power(%)	Phase(°)
TE ₁₁	86.1188	-64.8973
TE ₁₂	0.5952	-128.8639
TM ₁₁	11.9079	-52.644



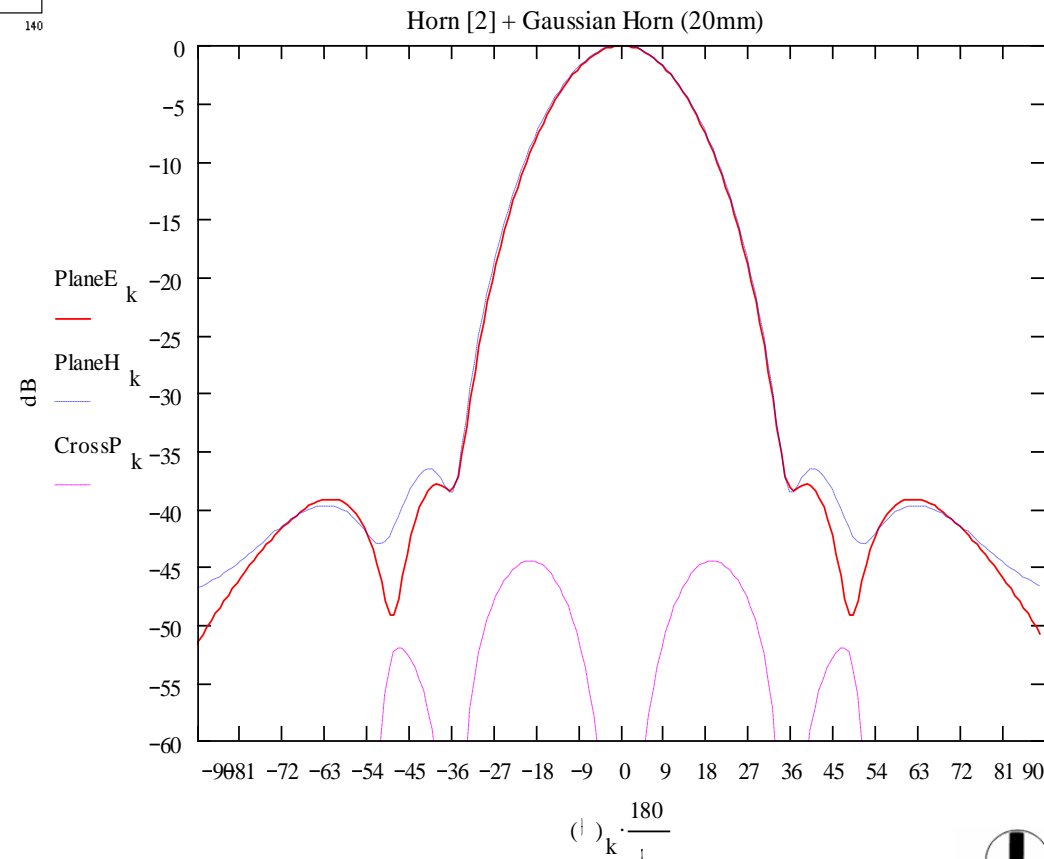
Profiled feed-horns



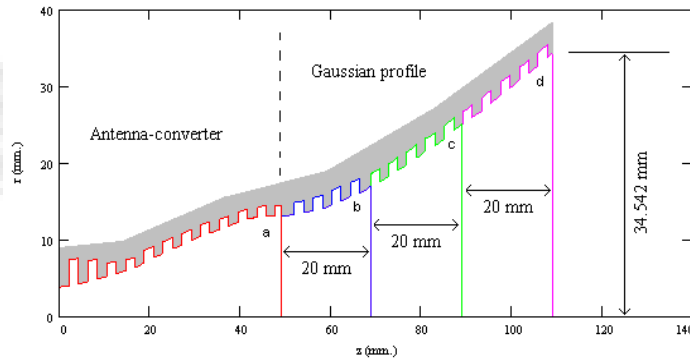
$R_{out}=17.063\text{mm}$
 Gain=18.6 dBi
 $\eta_E=98.9\%$

$\omega_0=10.15056\text{mm.}$
 $z_o=L+20$
 $\eta_H=98.9\%$

Modes	Power(%)	Phase(°)
TE_{11}	81.1053	-39.1864
TE_{12}	1.246	161.1546
TE_{13}	0.0373	-125.8252
TM_{11}	17.3833	-33.5072
TM_{12}	0.0697	-56.33
TM_{13}	0.1181	-44.3379



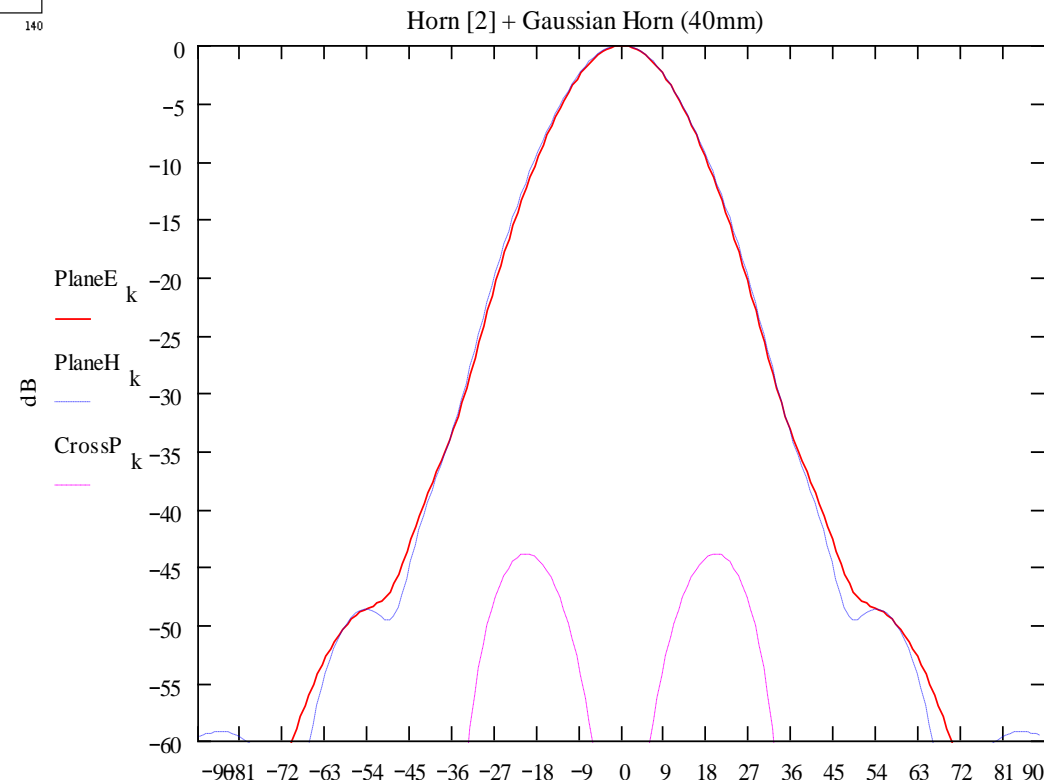
Profiled feed-horns



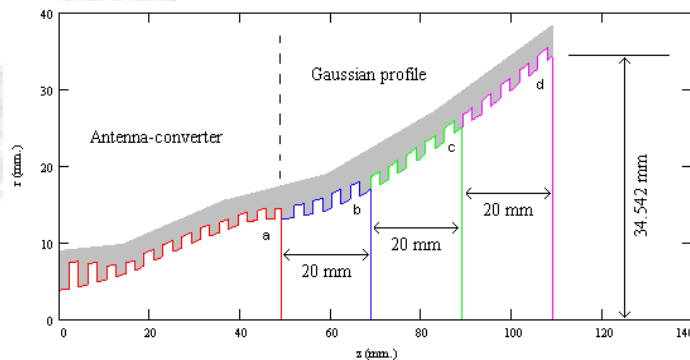
$R_{out}=25.087\text{mm}$
 $\text{Gain}=19.6\text{ dBi}$
 $\eta_E=99.9\%$

$\omega_0=10.41768\text{mm.}$
 $z_o=L+40$
 $\eta_H=99.8\%$

Modes	Power(%)	Phase(°)
TE ₁₁	56.0597	-33.3415
TE ₁₂	12.6842	-174.13
TE ₁₃	0.3934	87.1243
TE ₁₄	0.0064	79.0973
TM ₁₁	28.4174	-17.2304
TM ₁₂	2.3396	-139.6503
TM ₁₃	0.0677	168.4064
TM ₁₄	0.008	148.4184



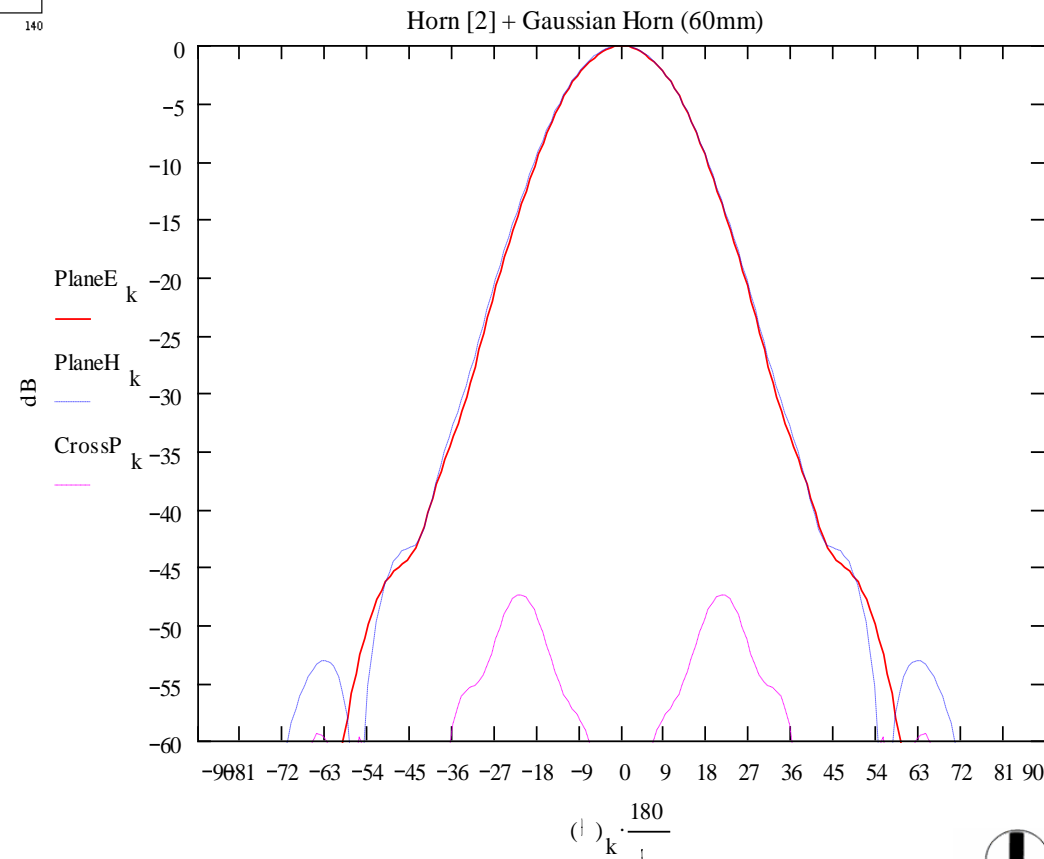
Profiled feed-horns



$R_{out}=34.541\text{mm}$
 Gain=17.2 dBi
 $\eta_E=99.9\%$

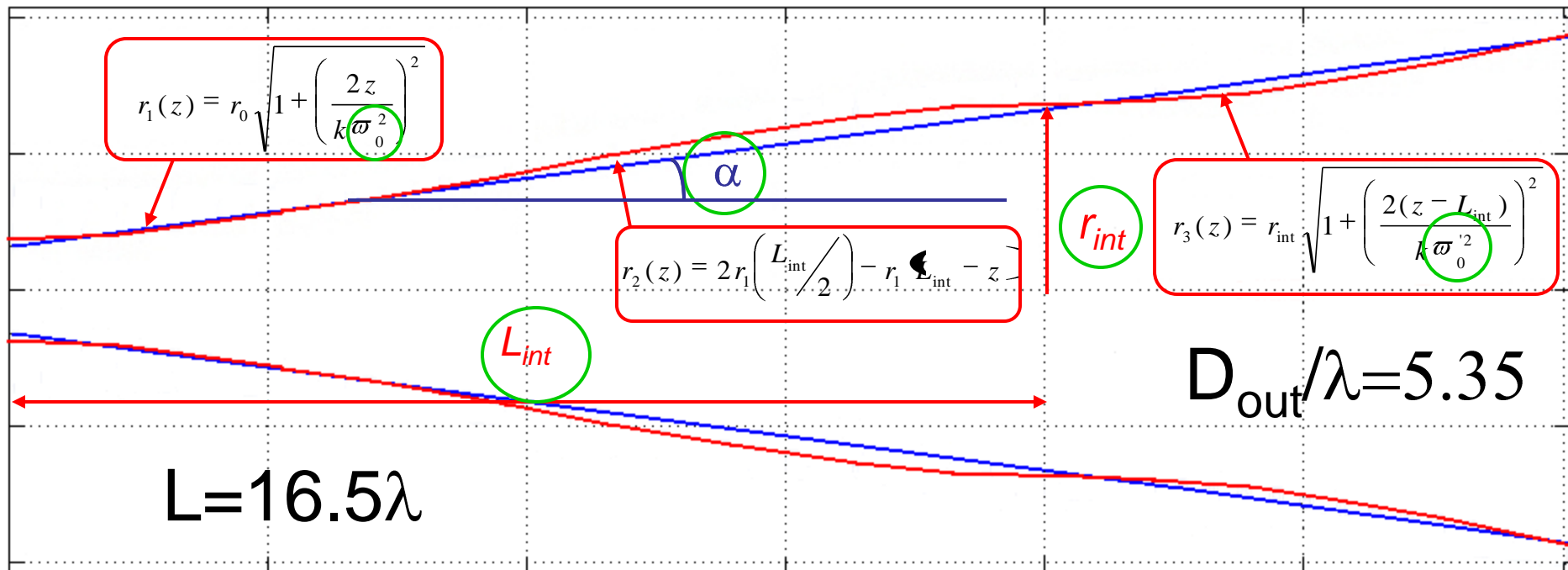
$\omega_0=10.41768\text{mm.}$
 $z_o=L+60$
 $\eta_H=99.8\%$

Modes	Power(%)	Phase(°)
TE ₁₁	33.9487	-34.5662
TE ₁₂	21.7407	-171.996
TE ₁₃	3.8836	80.7423
TE ₁₄	0.2659	10.2575
TE ₁₅	0.0159	-13.5281
TM ₁₁	29.7269	-14.157
TM ₁₂	9.3439	-138.6633
TM ₁₃	0.9579	130.5057
TM ₁₄	0.0629	85.7041
TM ₁₅	0.01	62.5564



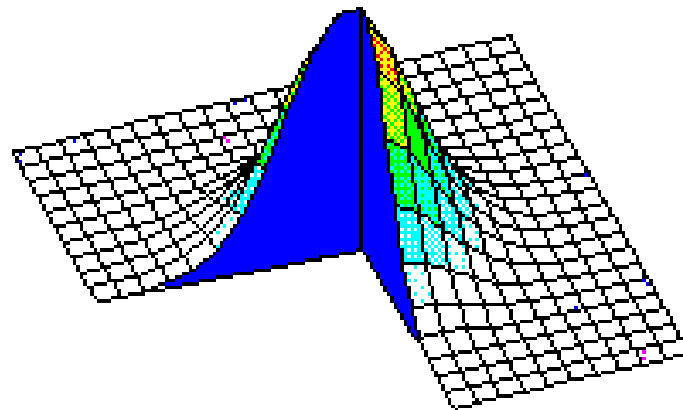
Theoretical Basis

- A comparative study of equal size antenna prototypes have been performed



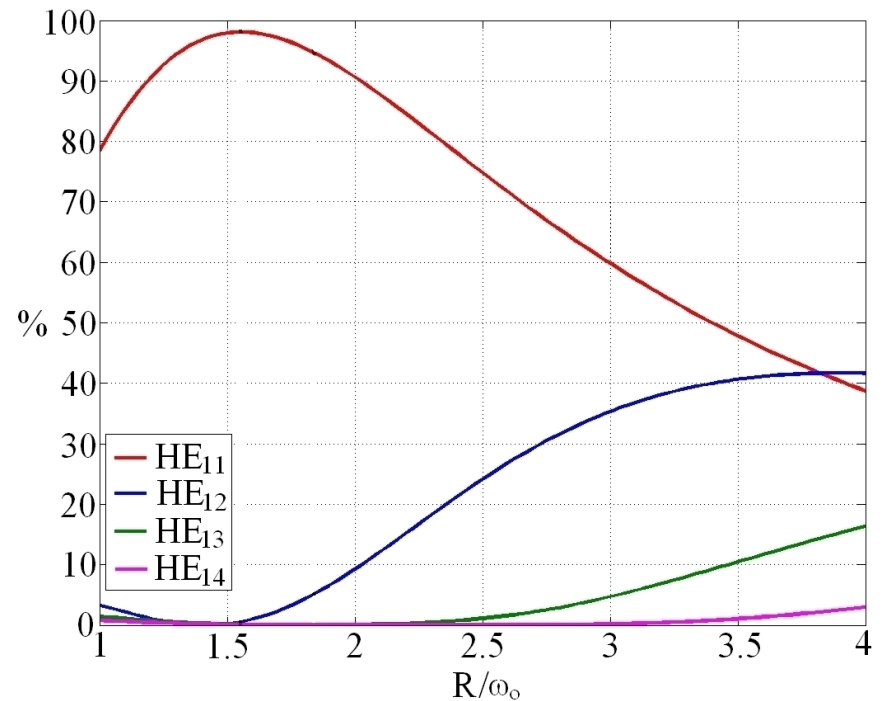
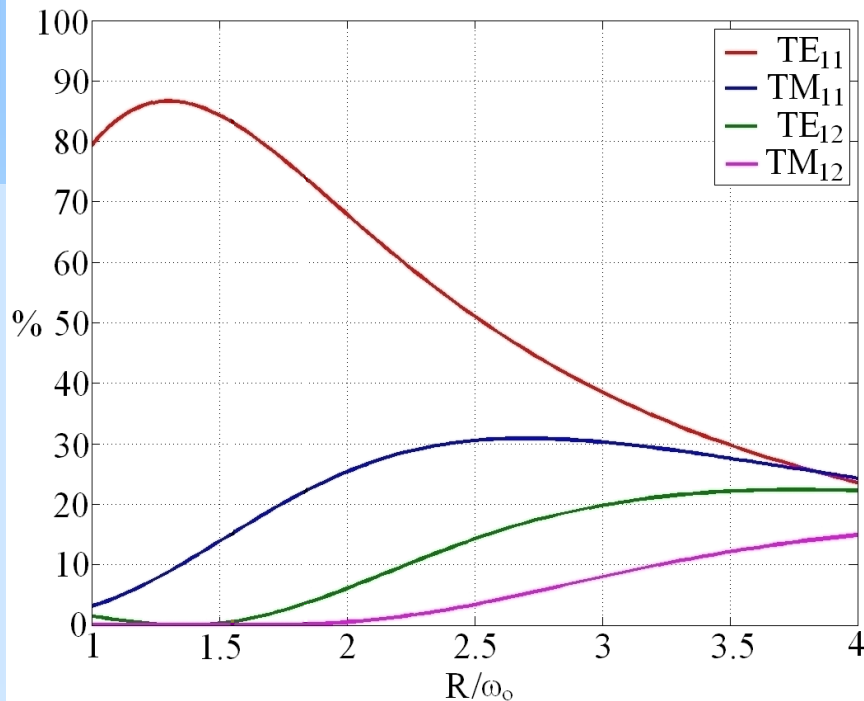
Theoretical Basis

- Relation between waveguide modes and space modes,
 - Waveguide modes
 - TE/TM
 - HE
 - Space modes
 - Gaussian beam modes
- Goal of GPHAs: to obtain the most pure Gaussian fundamental beam mode, Ψ_{00}



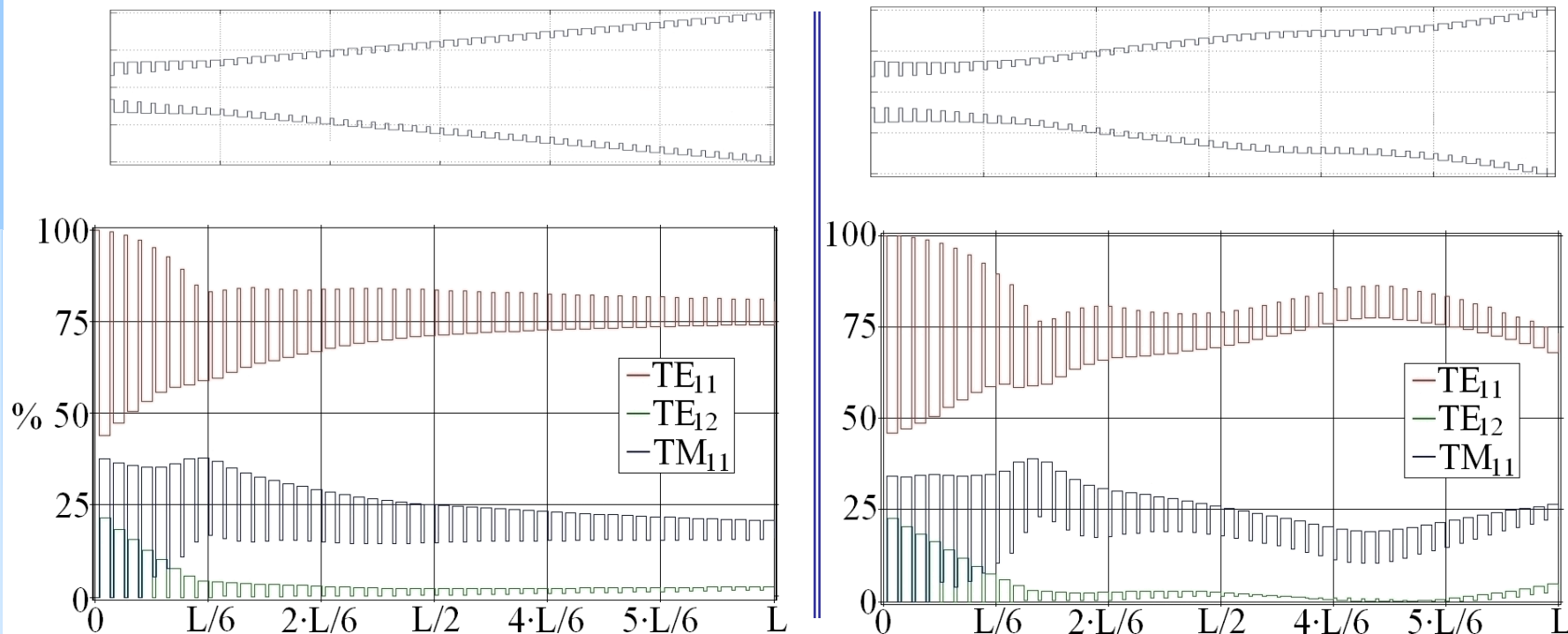
Theoretical Basis

- Expansion of Ψ_{00} in terms of TE/TM or HE modes, (correlation at aperture)



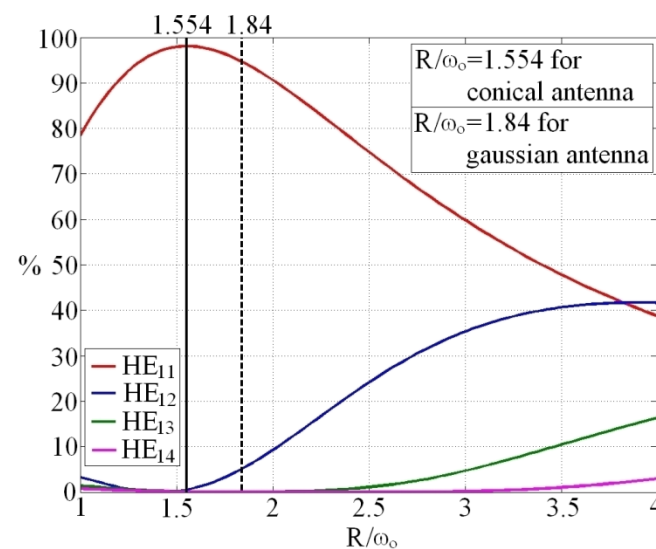
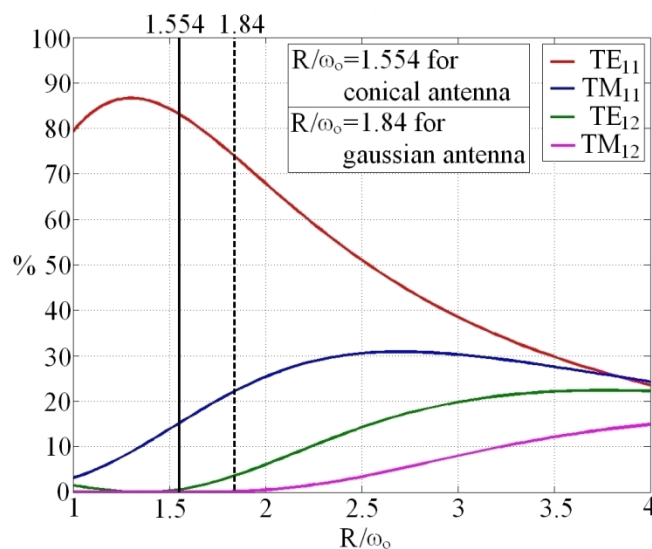
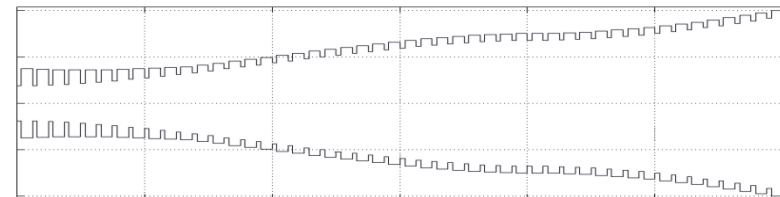
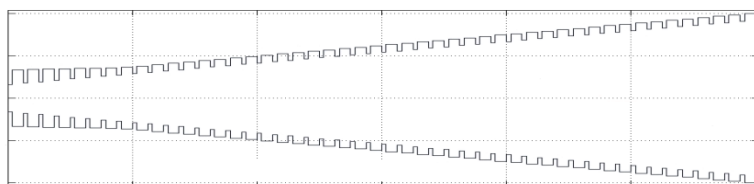
Comparison results

- Mode mixture along the antennas:



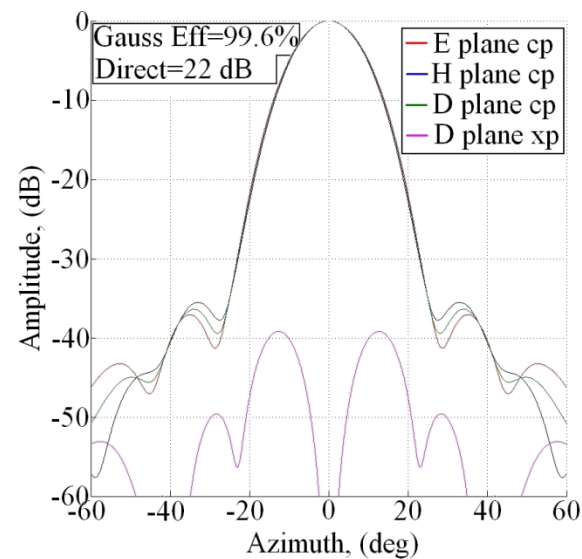
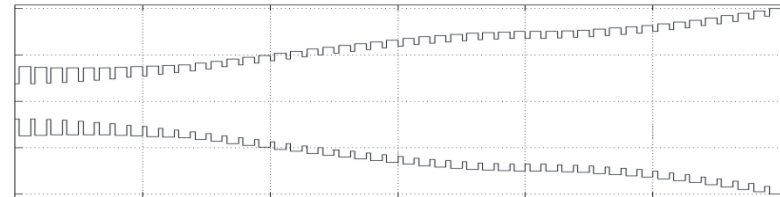
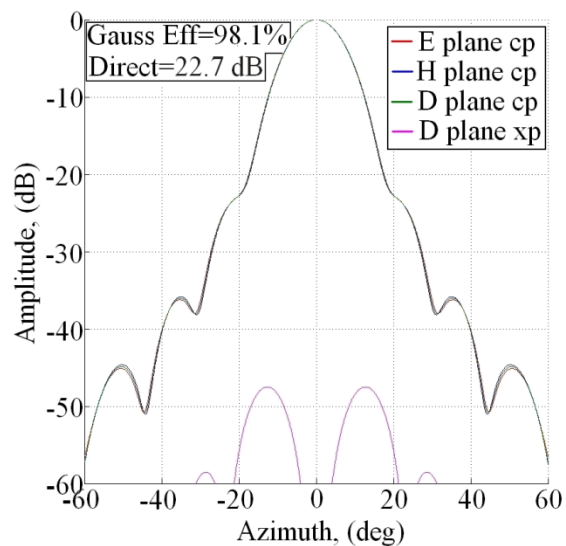
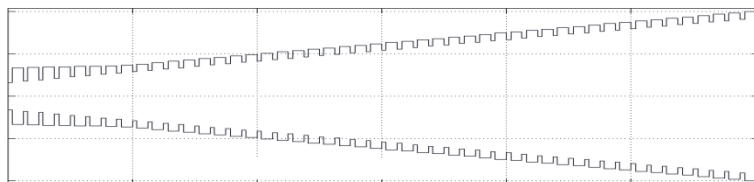
Comparison results

- Mode mixture at the aperture:



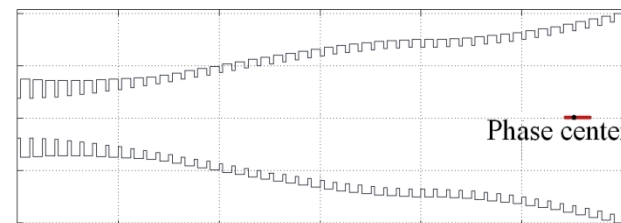
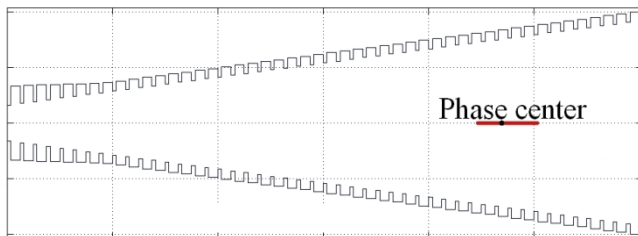
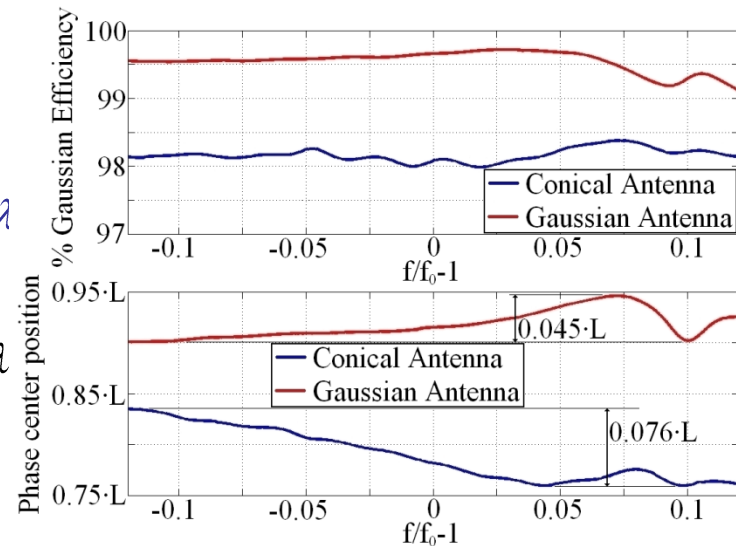
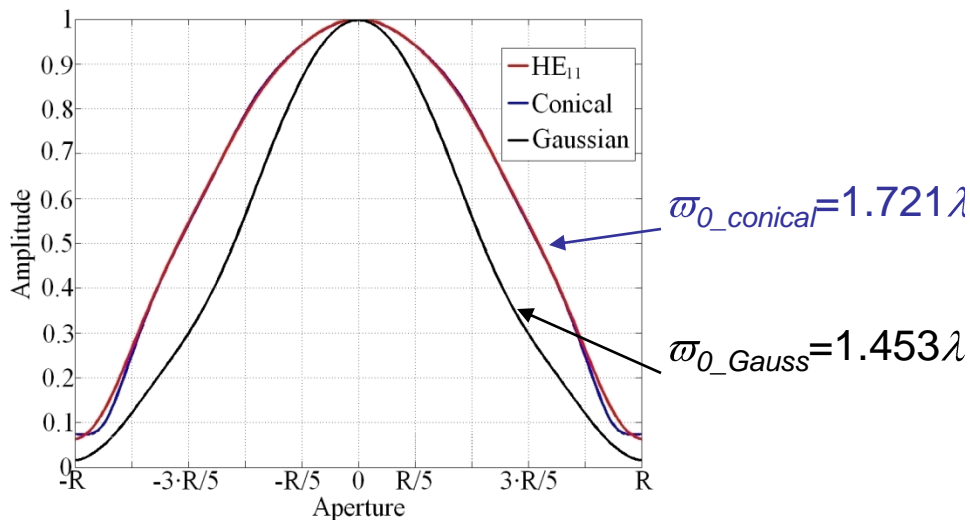
Comparison results

- The radiation pattern,



Comparison results

- The aperture field distribution,



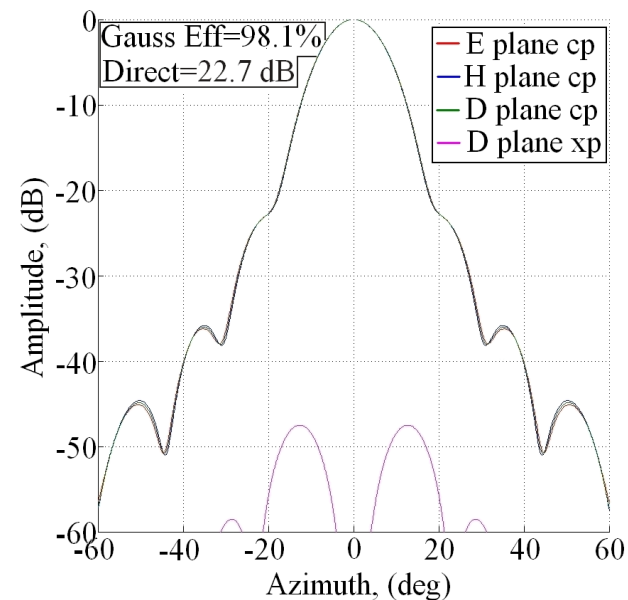
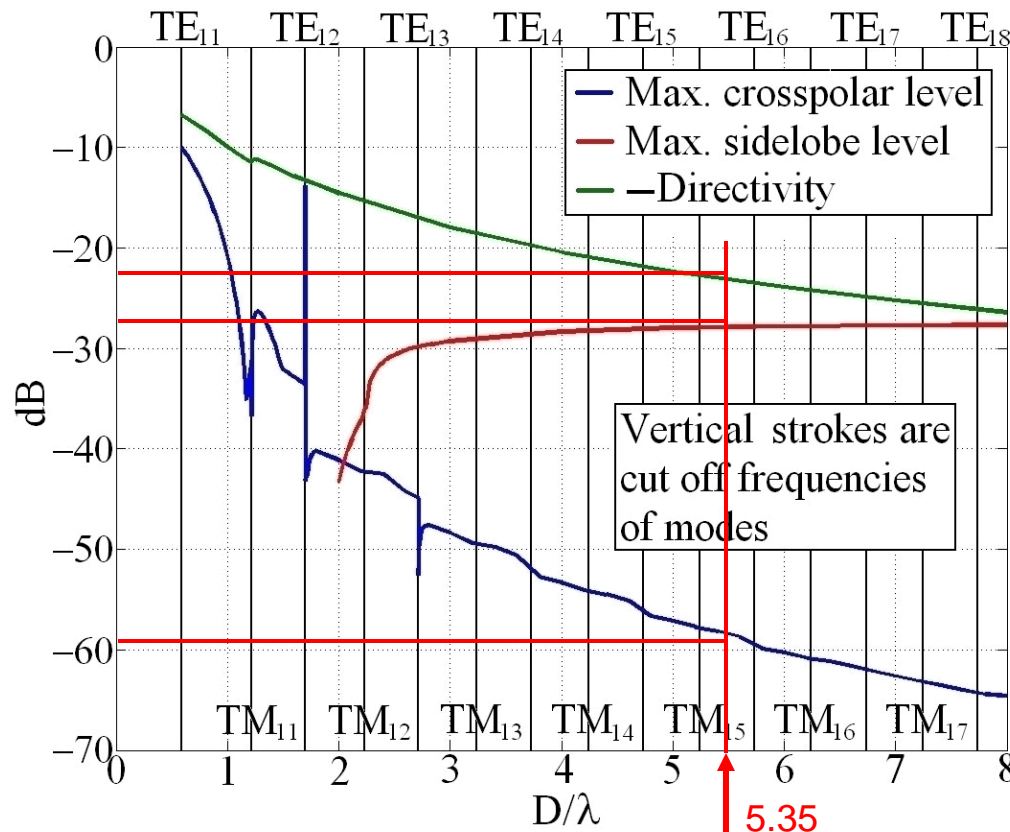
Comparison results

- Partial conclusions:

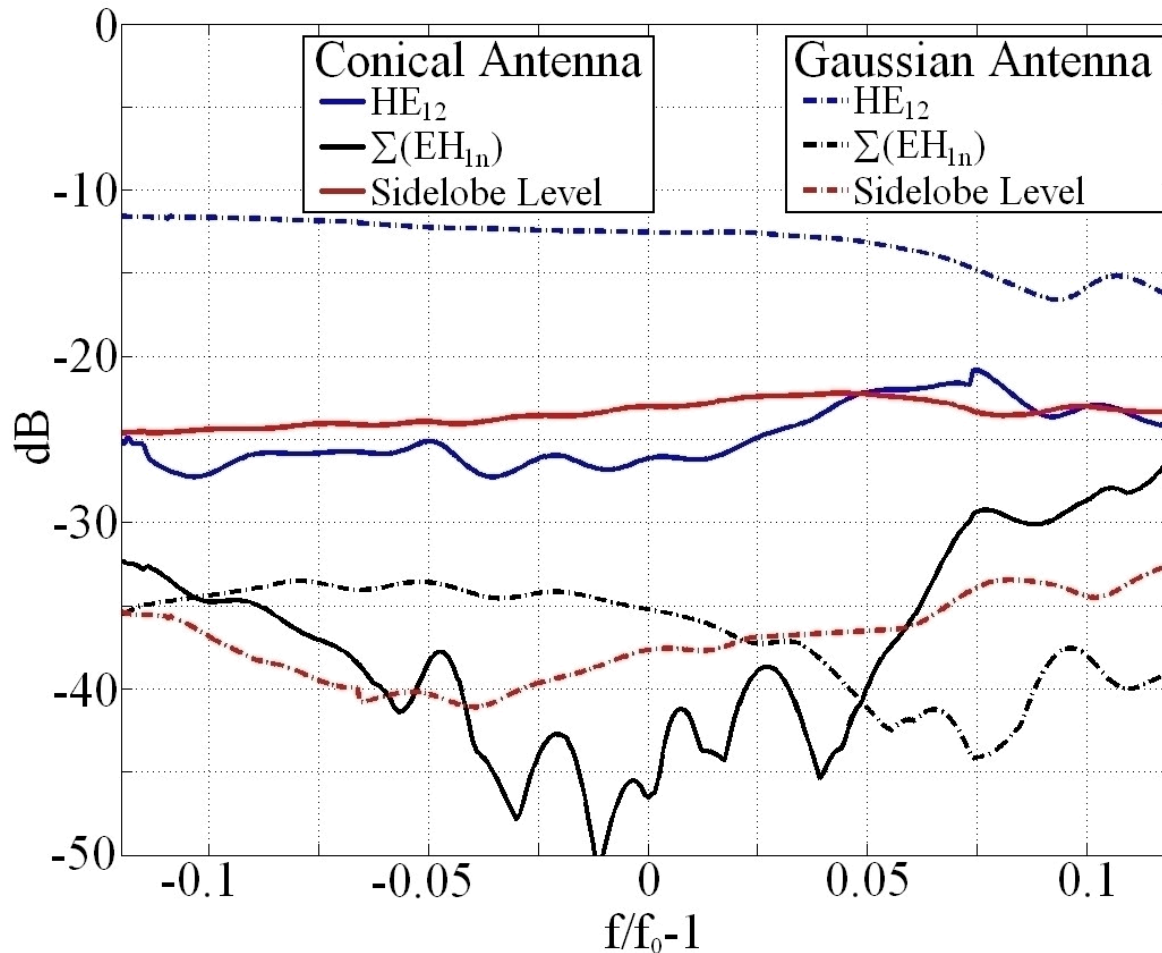
	<i>Conical</i>	<i>GPHA</i>
<i>Gauss Efficiency</i>	98.1%	99.6%
<i>Directivity</i>	22.7 dB	22.0 dB
<i>Phase center</i>	0.076L	0.045L
<i>Output Mode</i>	HE ₁₁	Ψ ₀₀

Comparison results

- Radiating features of HE_{11} mode



Comparison results

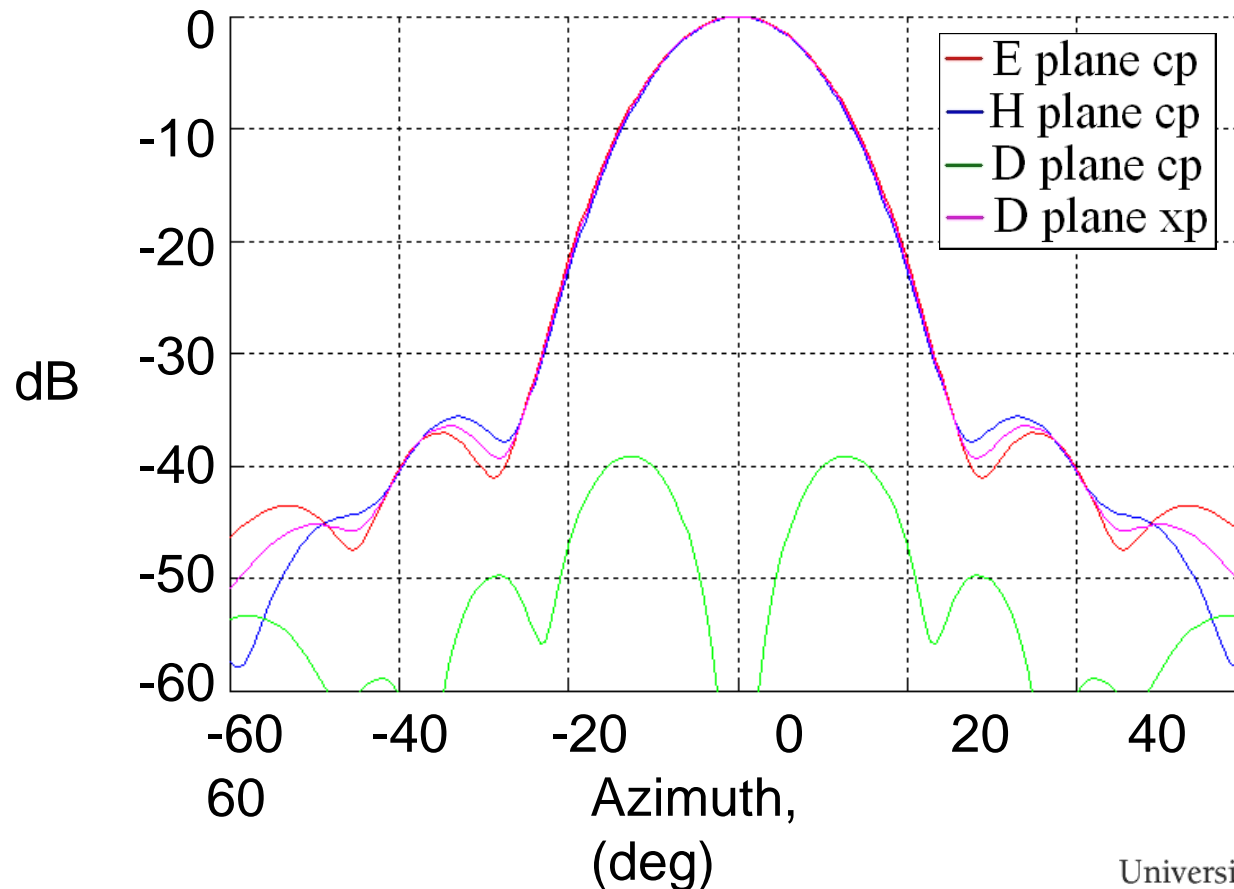


Comparison results

- Because of the four design parameters of the GPHAs, other possibilities:
 - Previous design parameters (GPHA)
 - $L_{\text{int}}=11.025\lambda$, $r_{\text{int}}=1.893\lambda$,
 $\varpi_0=0.928\lambda$, $\varpi_0'=1.323\lambda$
 - New design parameters (GPHA*)
 - $L_{\text{int}}=11.249\lambda$, $r_{\text{int}}=1.905\lambda$,
 $\varpi_0=0.904\lambda$, $\varpi_0'=1.288\lambda$
- Two equal-size GPHAs

Comparison results

Total length = 16.5λ $D/\lambda = 5.35$



Conical

D = 22.7 dB

Eff. = 98.1%

GPHA*

D = 22.6 dB

Eff. = 98.6%

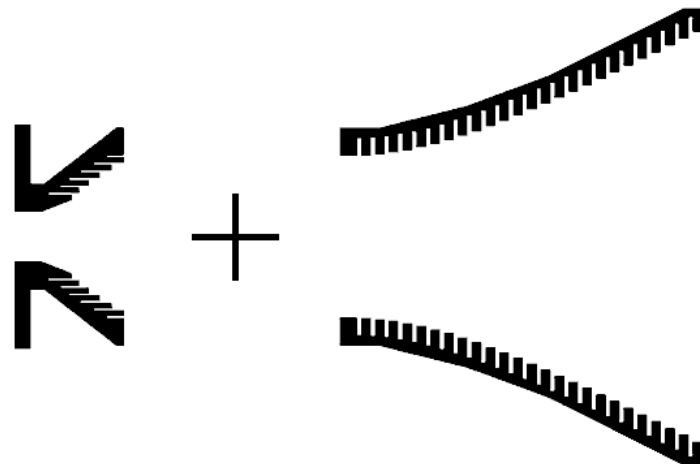
GPHA

D = 22 dB

Eff. = 99.6%

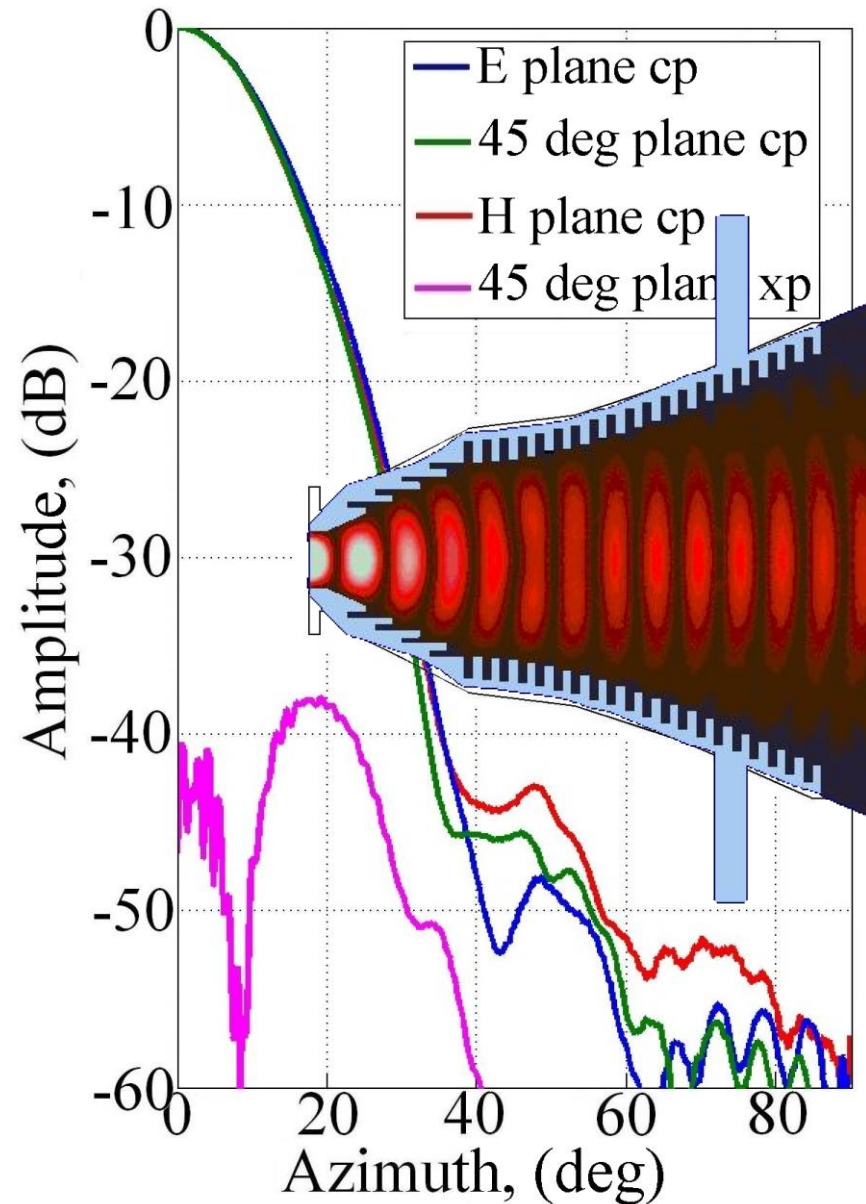
Profiled feed-horns

- In the last few years, a new type of profiled corrugated horn antenna has appeared.
- They consist in the combination of two technologies:
 - Horizontal corrugations for throat region
 - Vertical corrugations for flare region
- These profiles are very short for a reasonable nice bandwidth (25 to 30 %) and high performance.



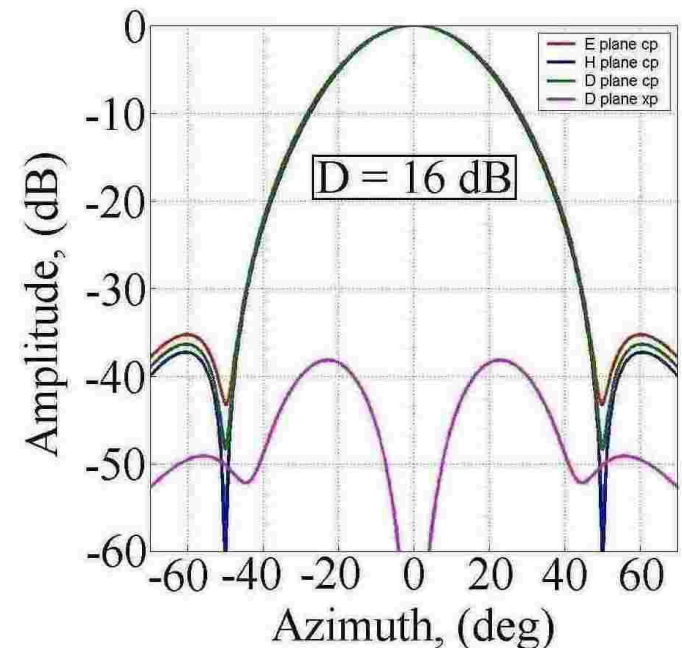
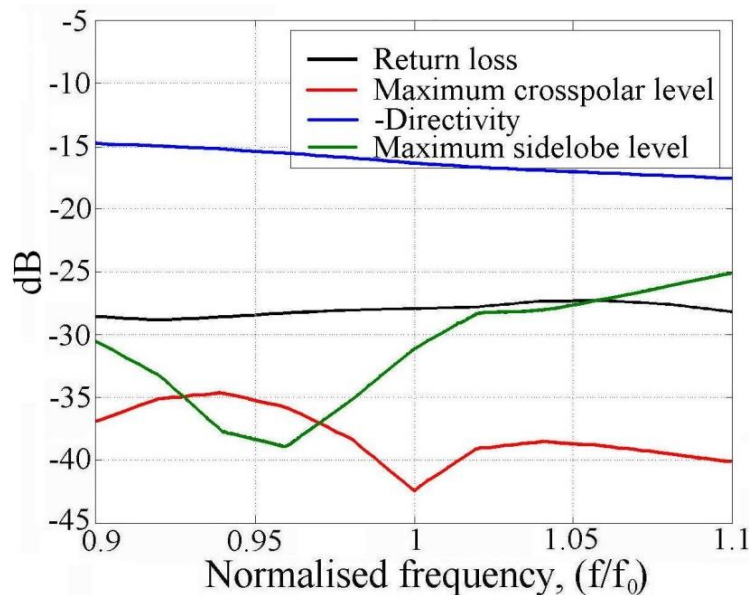
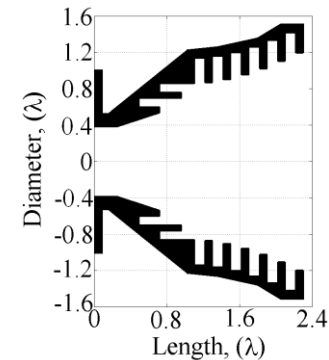
Profiled feed-horns

- First design of the combination of vertical and horizontal corrugations:



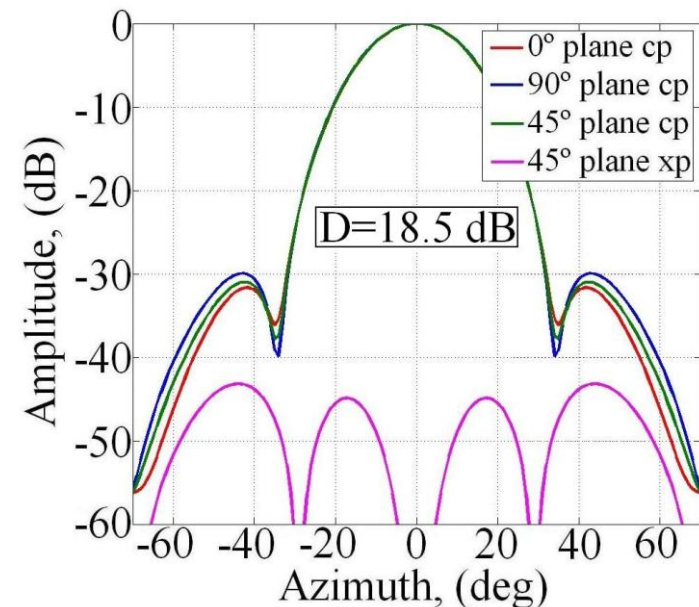
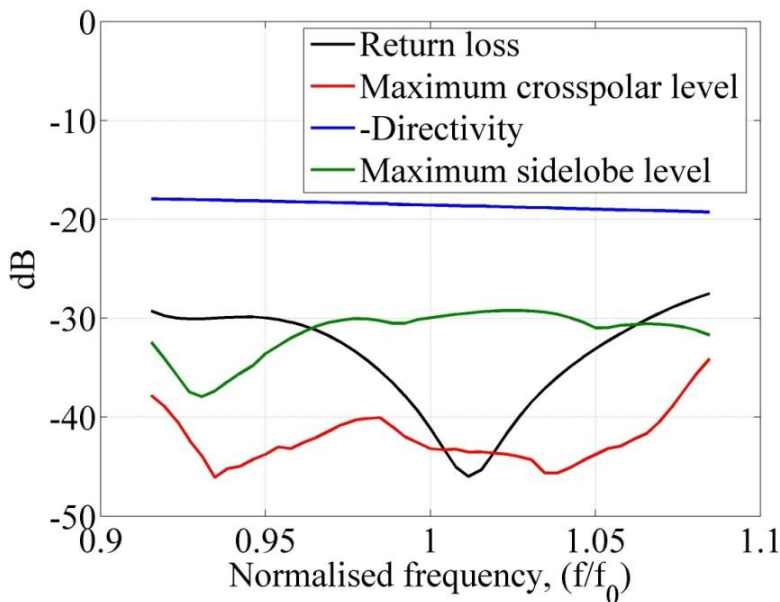
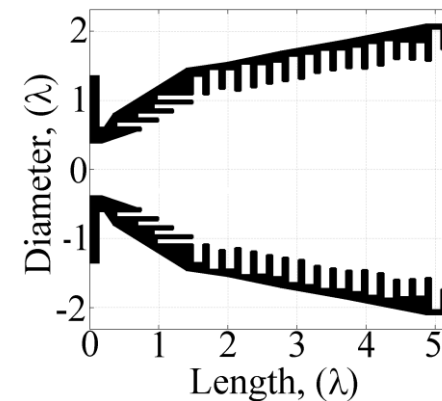
Profiled feed-horns

- 16 dB directivity horn antenna, 20% bandwidth in 2.3λ length, ensuring:
 - Return loss < -27 dB
 - X-pol < -35 dB
 - Side-lobes < -25 dB



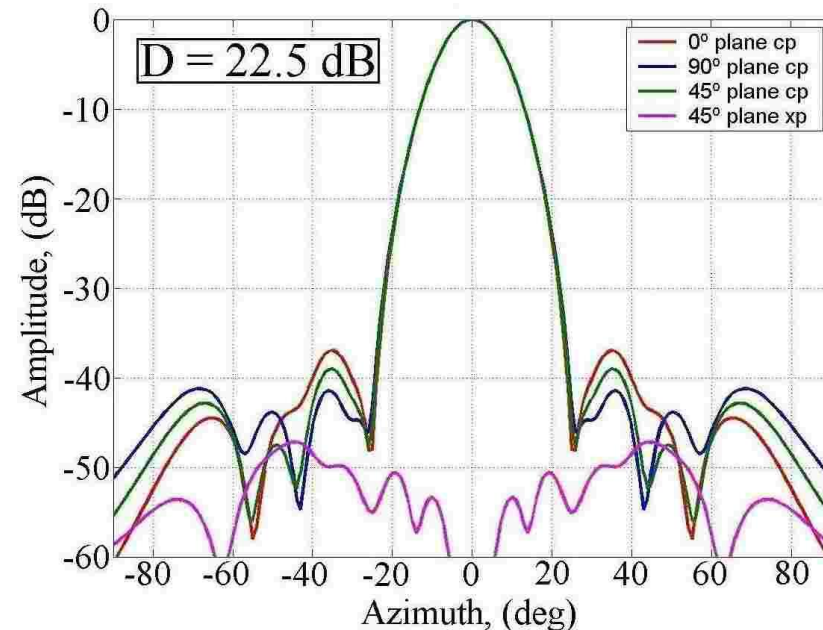
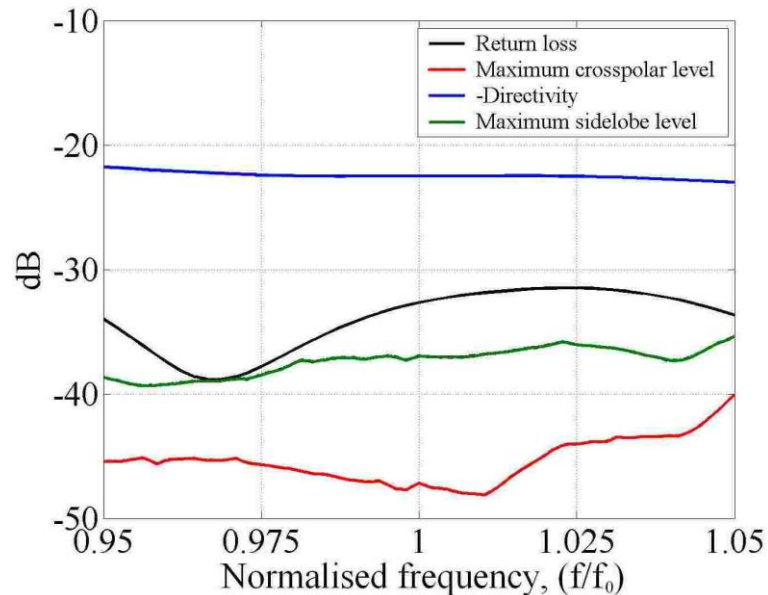
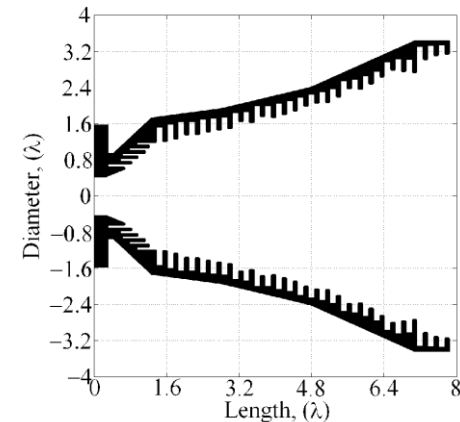
Profiled feed-horns

- 18.5 dB directivity horn antenna, 15% bandwidth in 5.2λ length, ensuring:
 - Return loss < -30 dB
 - X-pol < -40 dB
 - Side-lobes < -30 dB



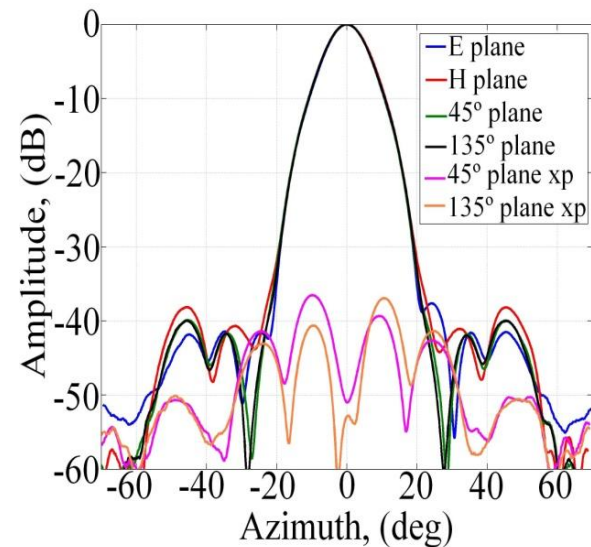
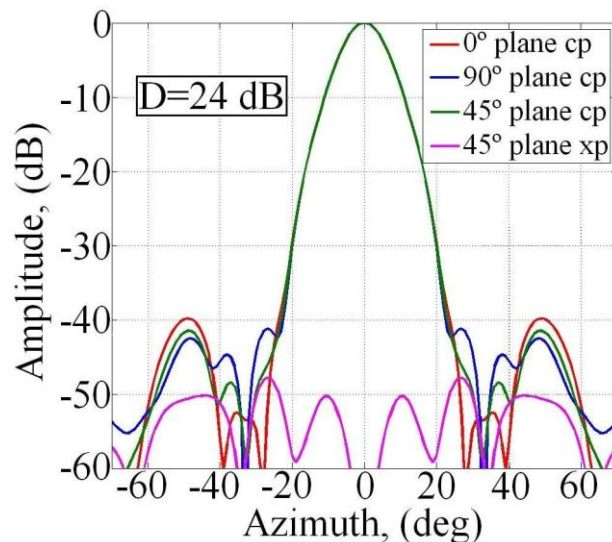
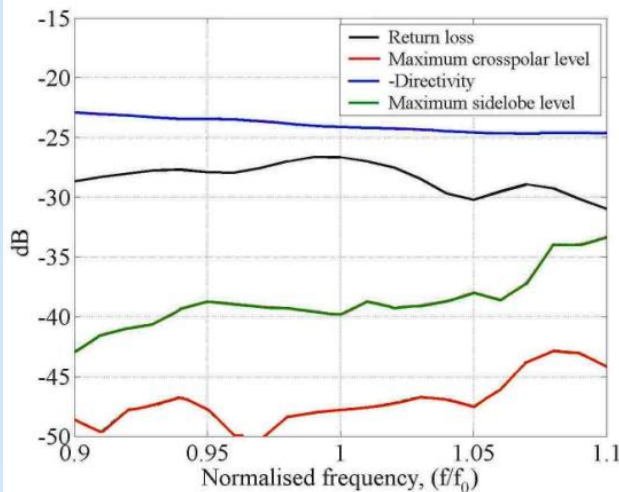
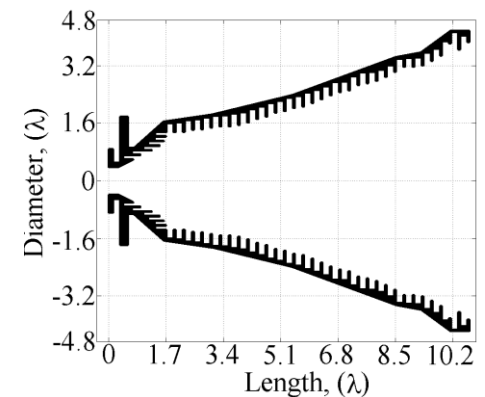
Profiled feed-horns

- 22.5 dB directivity horn antenna 10% bandwidth in 7.9λ length ensuring:
 - Return Loss < -31dB
 - X-pol < -40dB
 - Side-lobes < -35dB



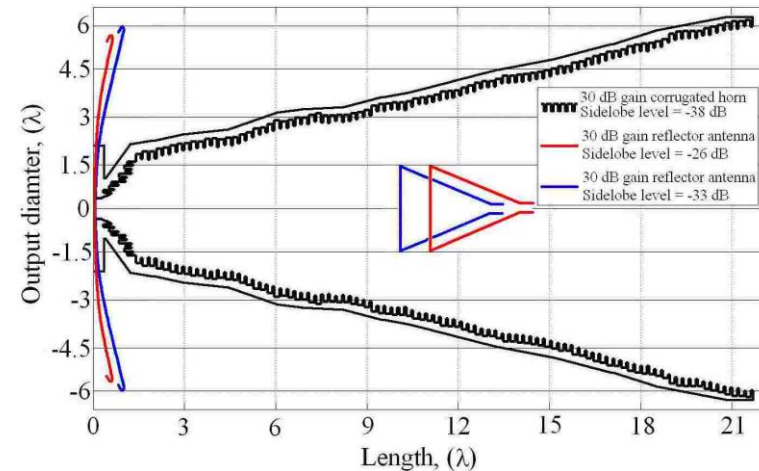
Profiled feed-horns

- 24 dB directivity horn antenna, 20% bandwidth in 10.5λ length ensuring (manufactured and Measured):
 - Return Loss < -27dB
 - X-pol < -43dB
 - Side-lobes < -34dB

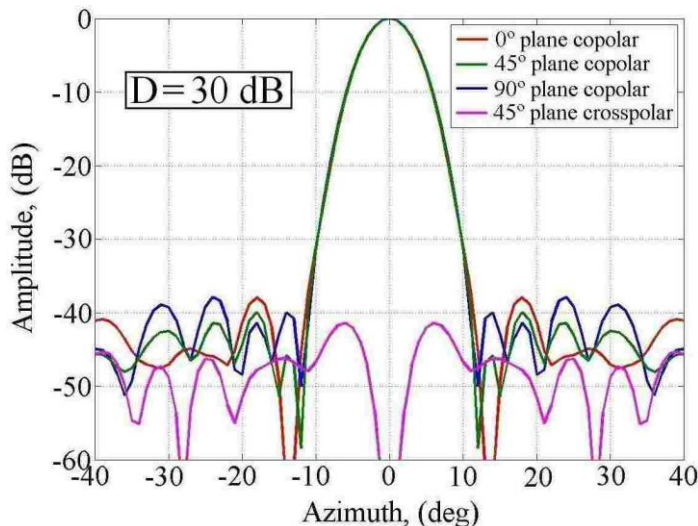


Profiled feed-horns

- The size reduction for new corrugated horns may drive to be considered as substitutes of reflector antennas in the range between 24 to 30 dB Gain



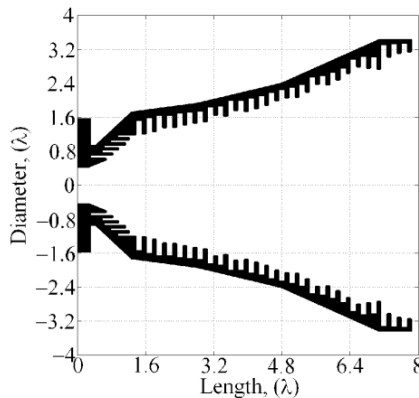
$$G = 10 \cdot \log_{10} \left(\eta \cdot \left(\frac{\pi \cdot D}{\lambda} \right)^2 \right)$$



	Reflector antenna 1	Reflector antenna 2	Corrug. horn
Directivity	30 dB	30 dB	30 dB
Diameter	$11.2 \cdot \lambda$	$12 \cdot \lambda$	$13 \cdot \lambda$
Focal length	$11.1 \cdot \lambda$	$10.2 \cdot \lambda$	-----
Total length	$14.1 \cdot \lambda$	$13.2 \cdot \lambda$	$22 \cdot \lambda$
Illum. taper	-11 dB	-16 dB	-----
η	0.808	0.704	-----
Sidelobe level	-26 dB	-33 dB	-38 dB

Profiled feed-horns

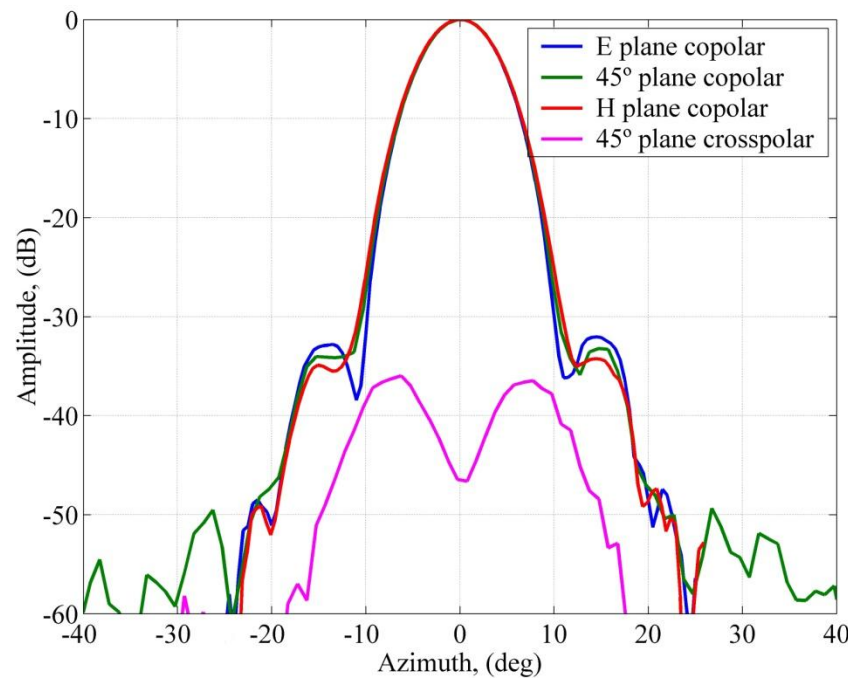
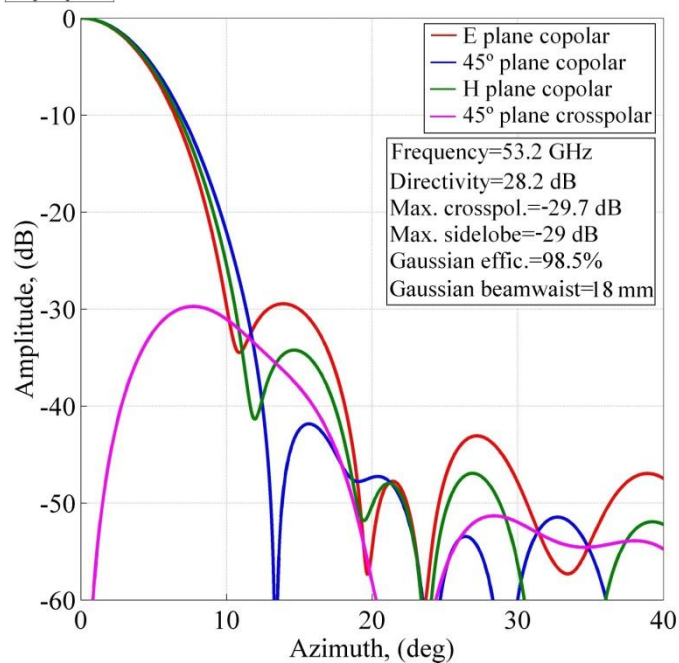
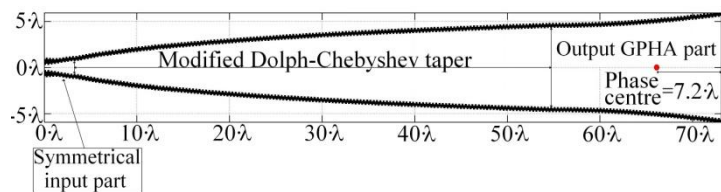
- In the table an study of the spillover significance has been made as an example for the 22.5 dB gain antenna increasing artificially its side-lobe level from -40 dB to -20 dB
- In the table can be seen the dramatic reduction in total efficiency that happens when side-lobe level is below -30 dB



	η_i	η_s	$\eta_i \cdot \eta_s$
-20 dB	0.885	0.720	0.637
-25 dB	0.885	0.846	0.749
-30 dB	0.885	0.896	0.793
-40 dB	0.885	0.919	0.813

Examples

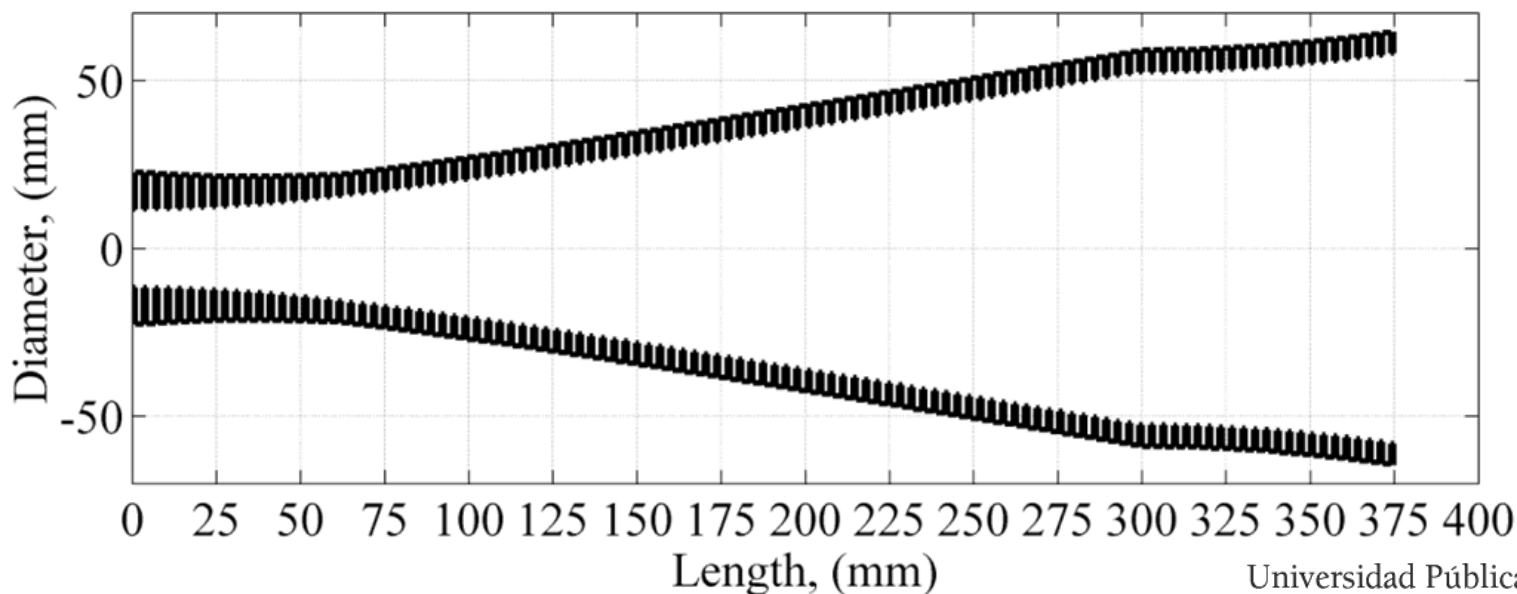
- Corrugated GPHA design for low power testing of the quasioptical transmission lines at TJ-II stellerator:



Examples

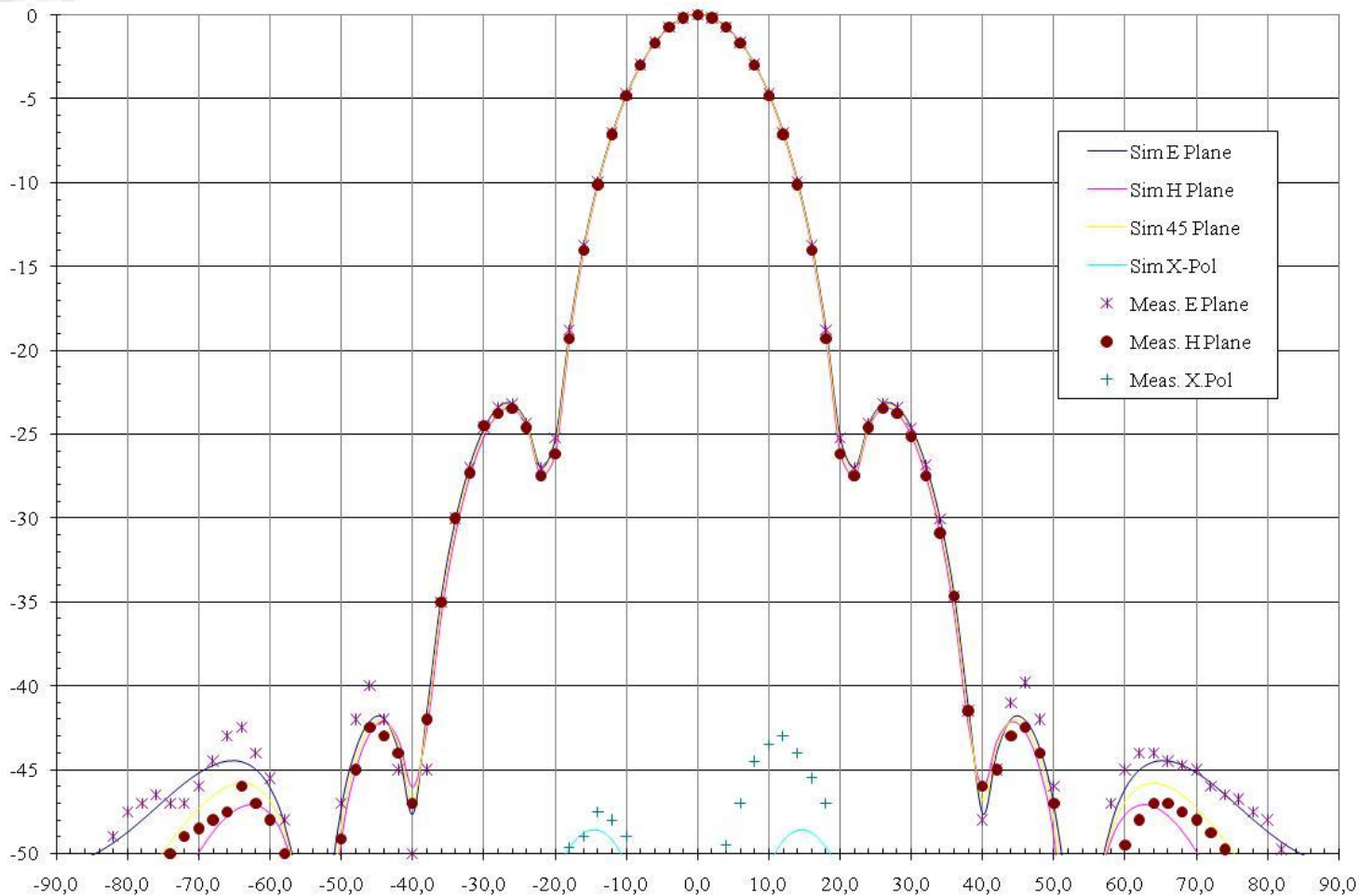
- Corrugated GPHA design for Hispasat 1C satellite:

- Frequency range: 11.7 to 12.2 GHz
- Illumination: 19 deg below -22 dB
- Maximum cross-polar: below -45 dB
- RL below -28 dB
- Input dia.: 23.4 mm

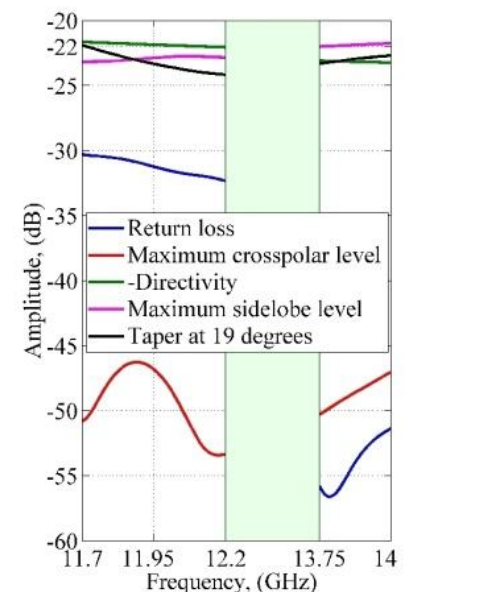
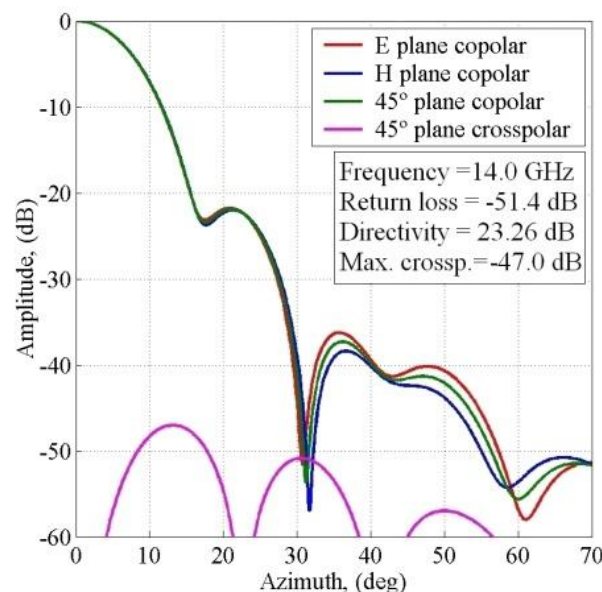
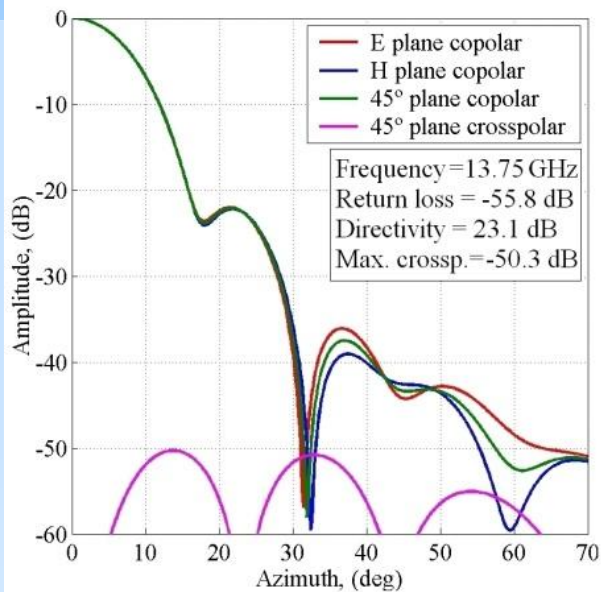
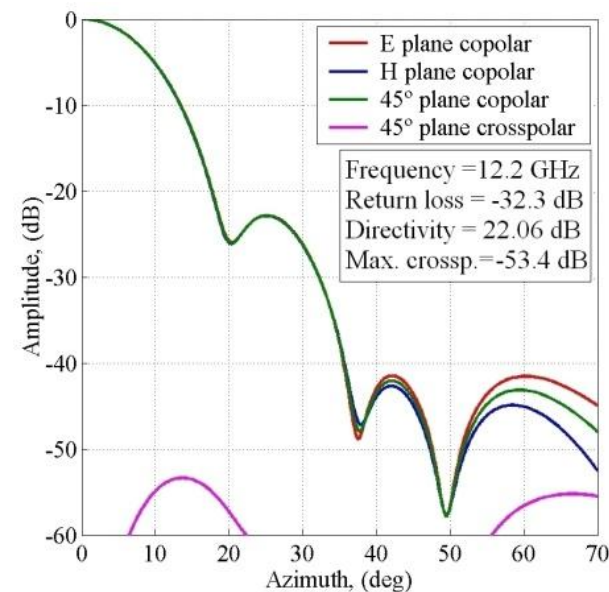
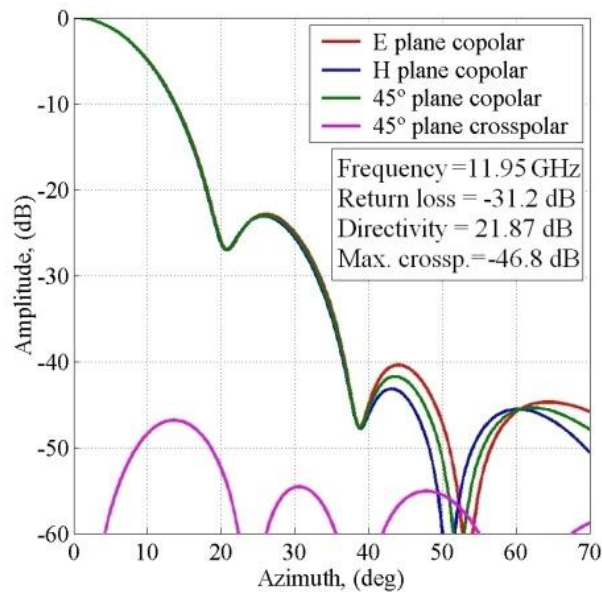
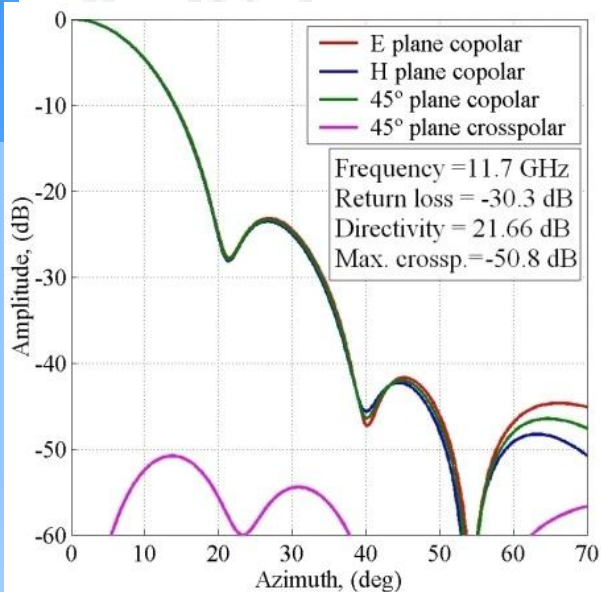


Examples

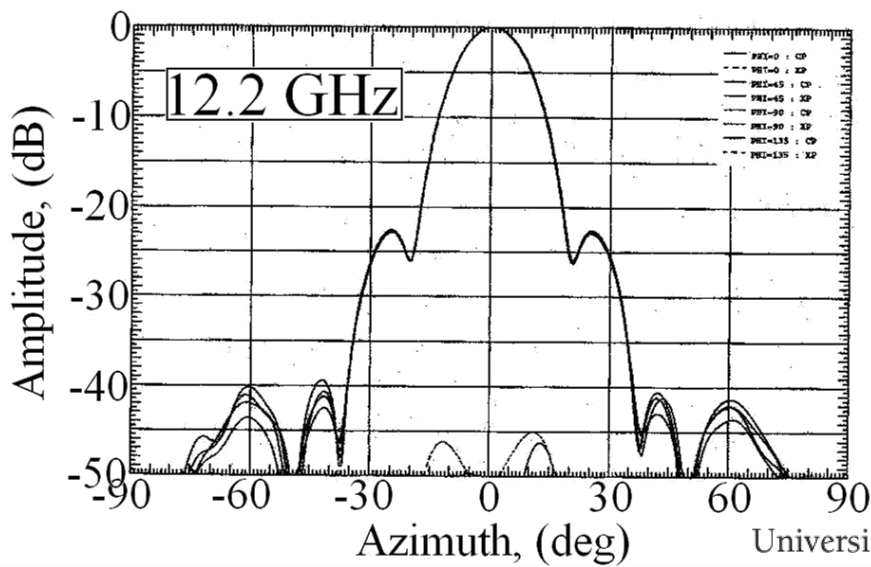
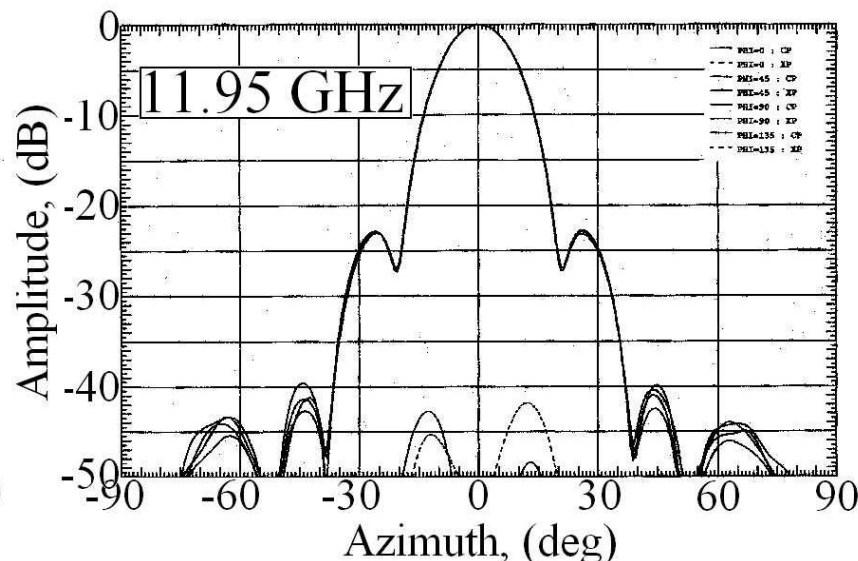
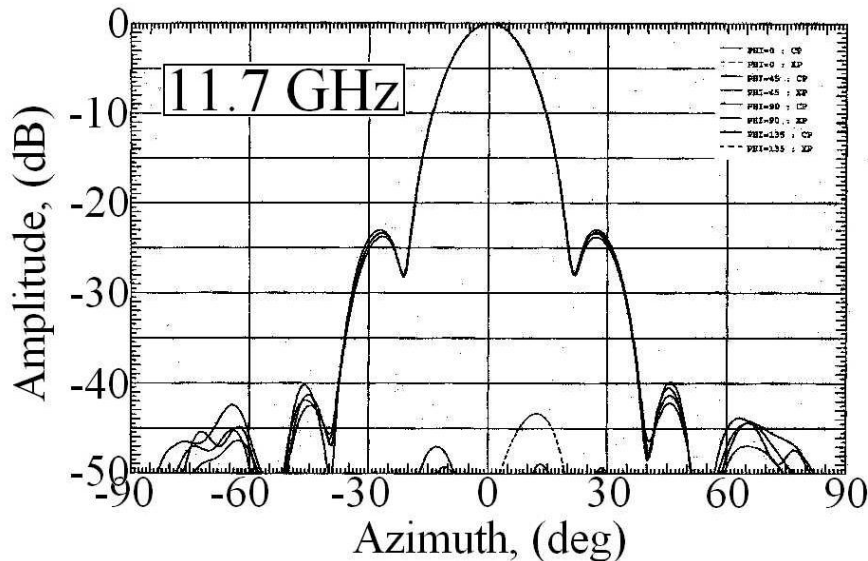
- Comparison between simulation and measurements...



Examples

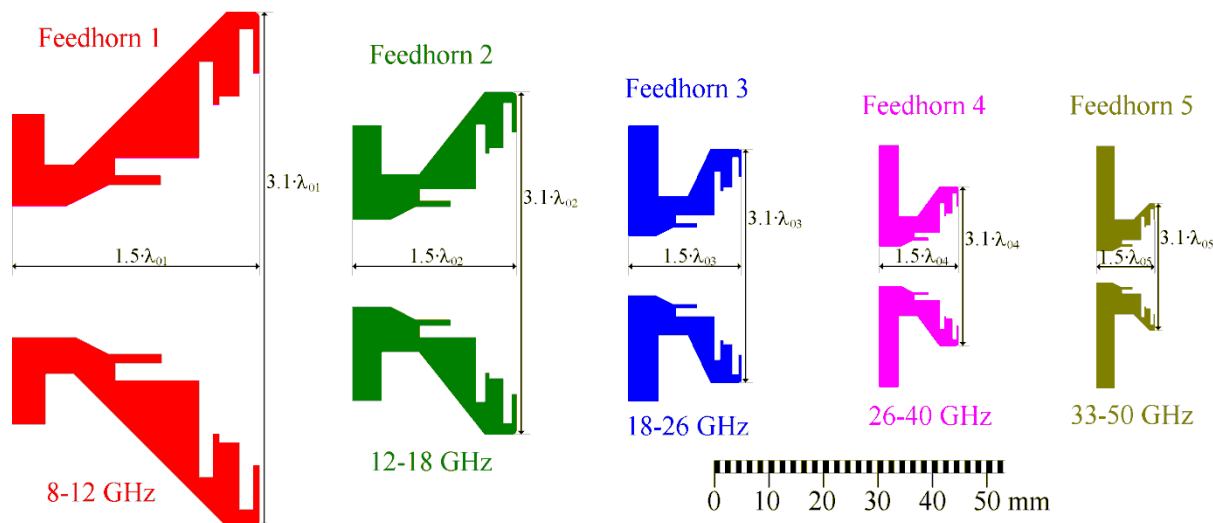
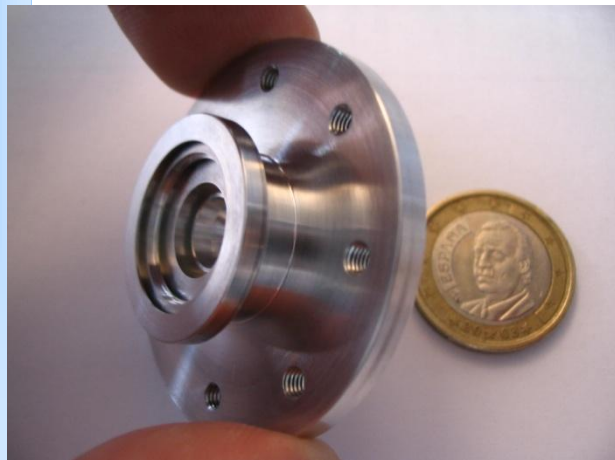
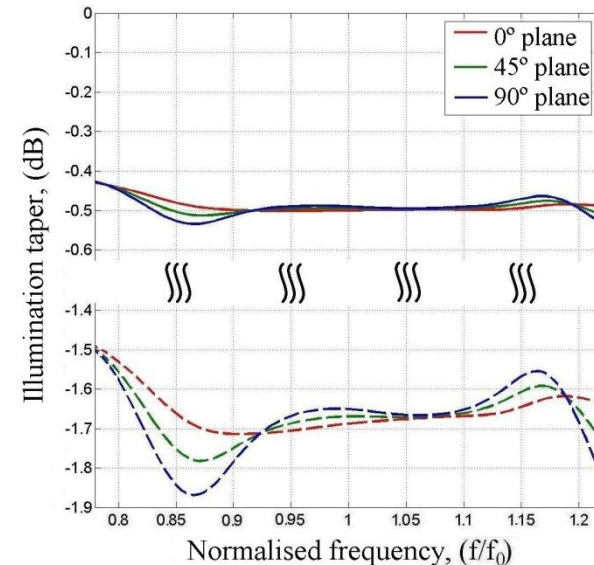
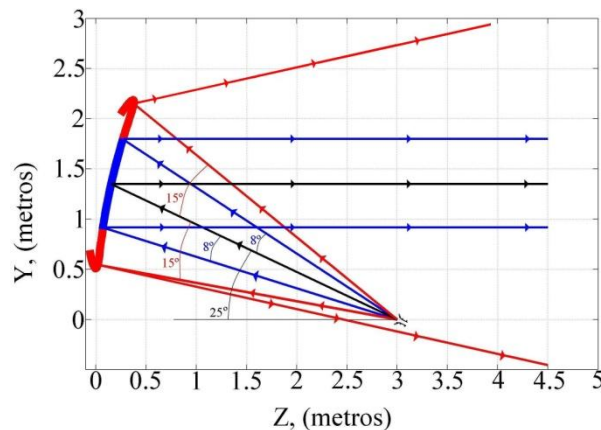


Examples



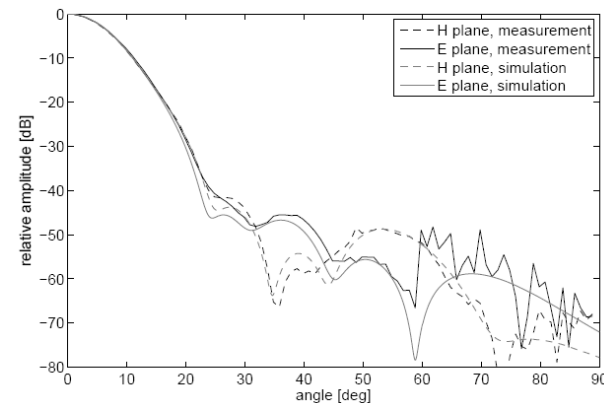
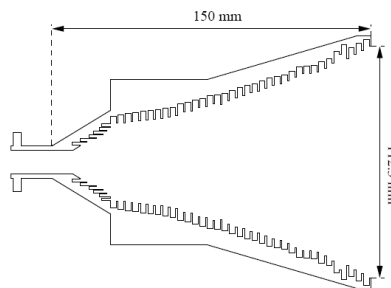
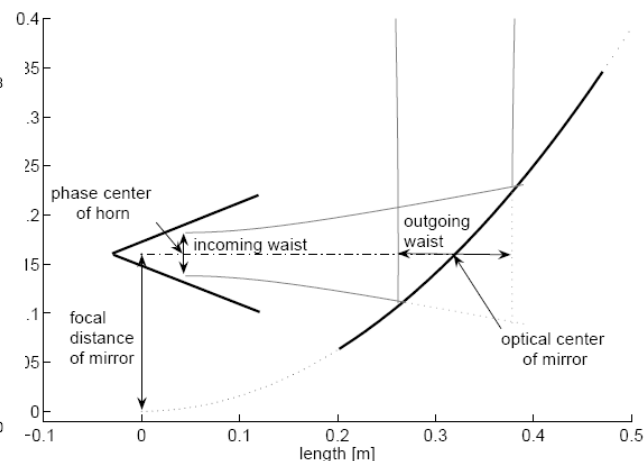
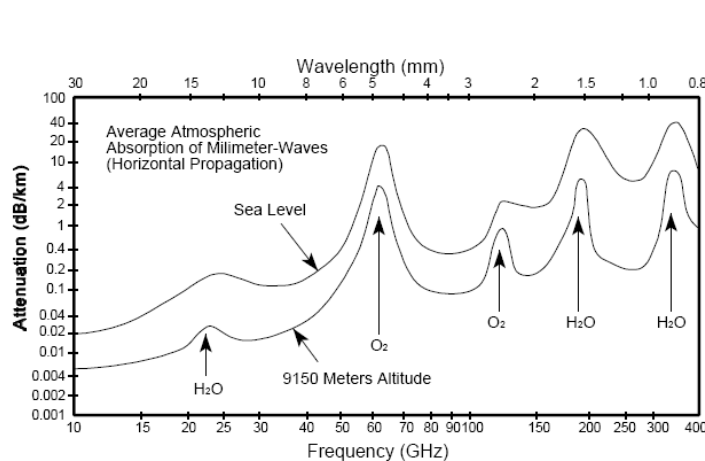
Examples

- Corrugated profiled feed-horns design for Limoges' University CATR



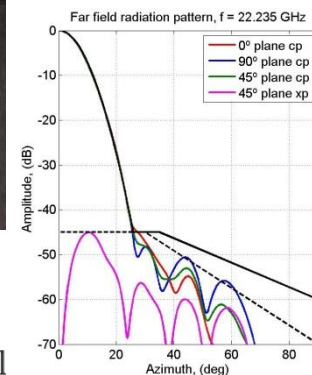
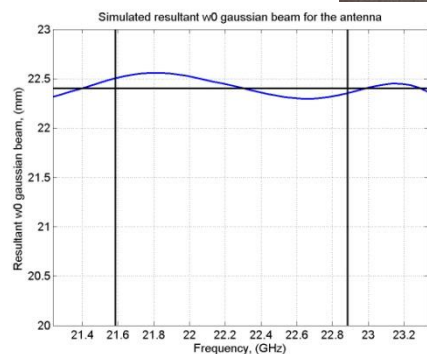
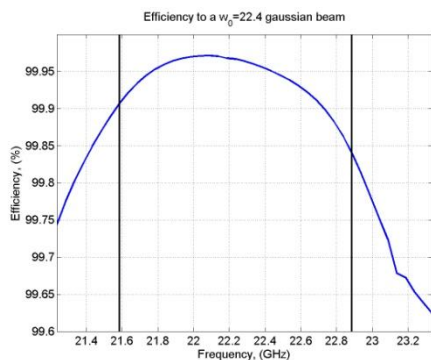
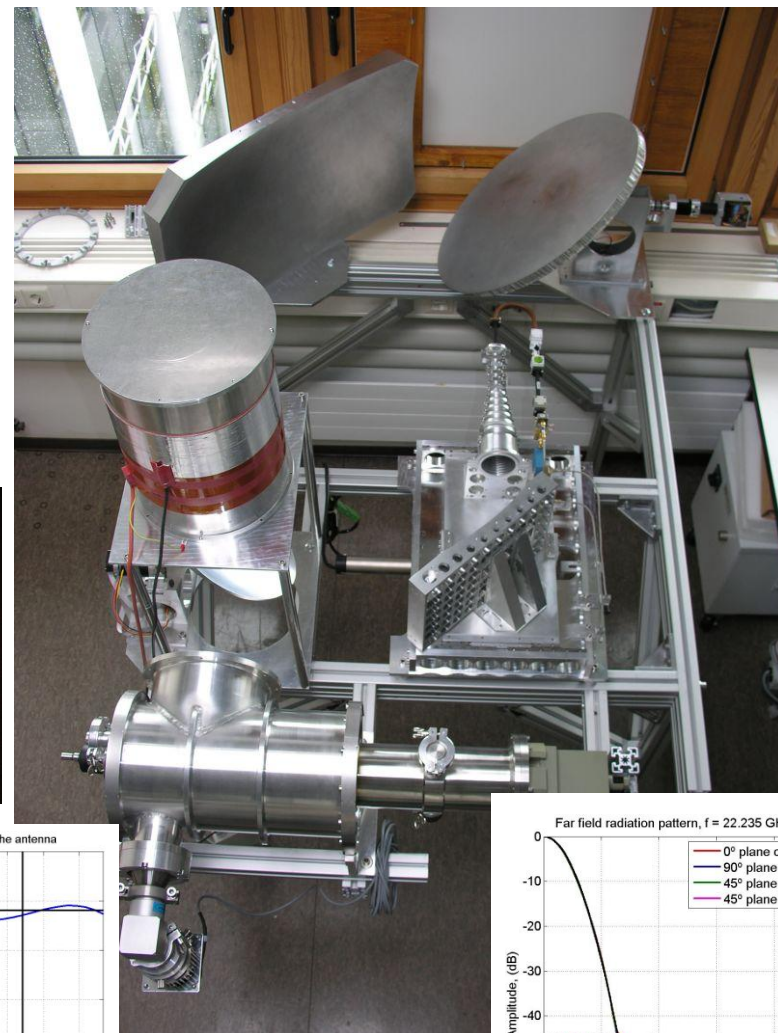
Examples

- Corrugated profiled feed-horn design for MIAWARA-C radiometer at 22 GHz in Bern University



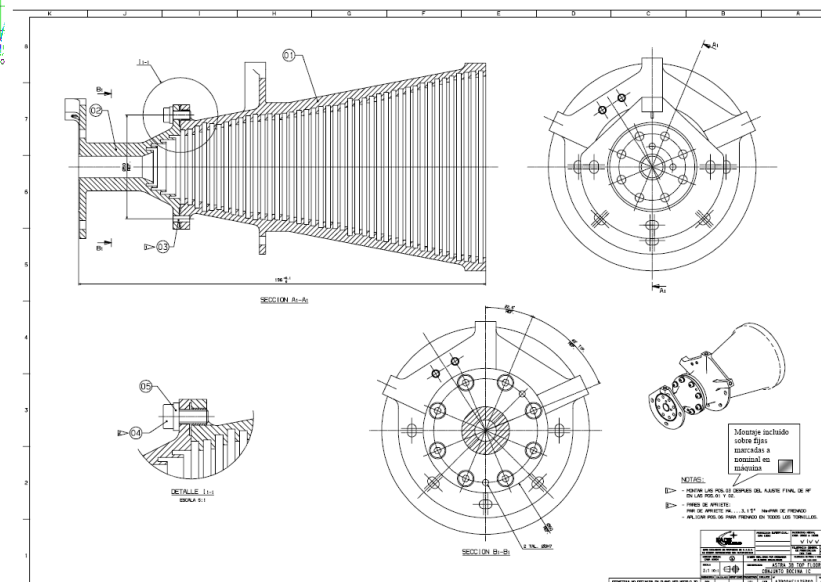
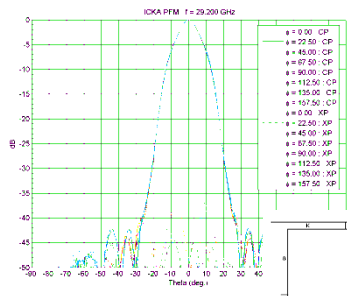
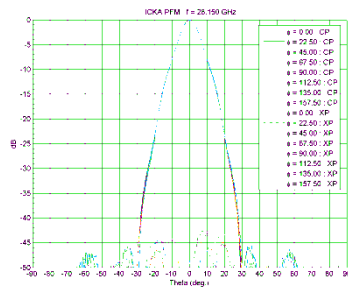
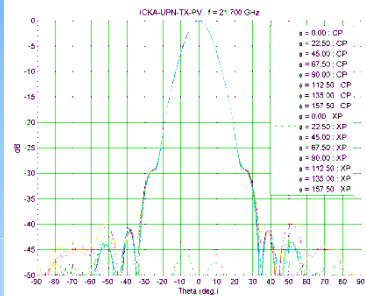
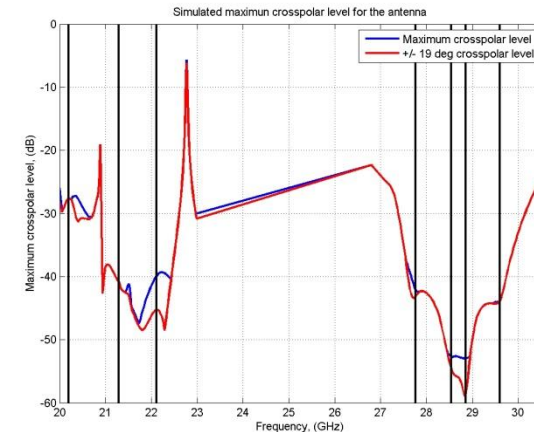
Examples

- Corrugated profiled feed-horn design for MIRA-5 radiometer at 22 GHz in Forschungszentrum Karlsruhe (Institute for Meteorology and Climate Research)



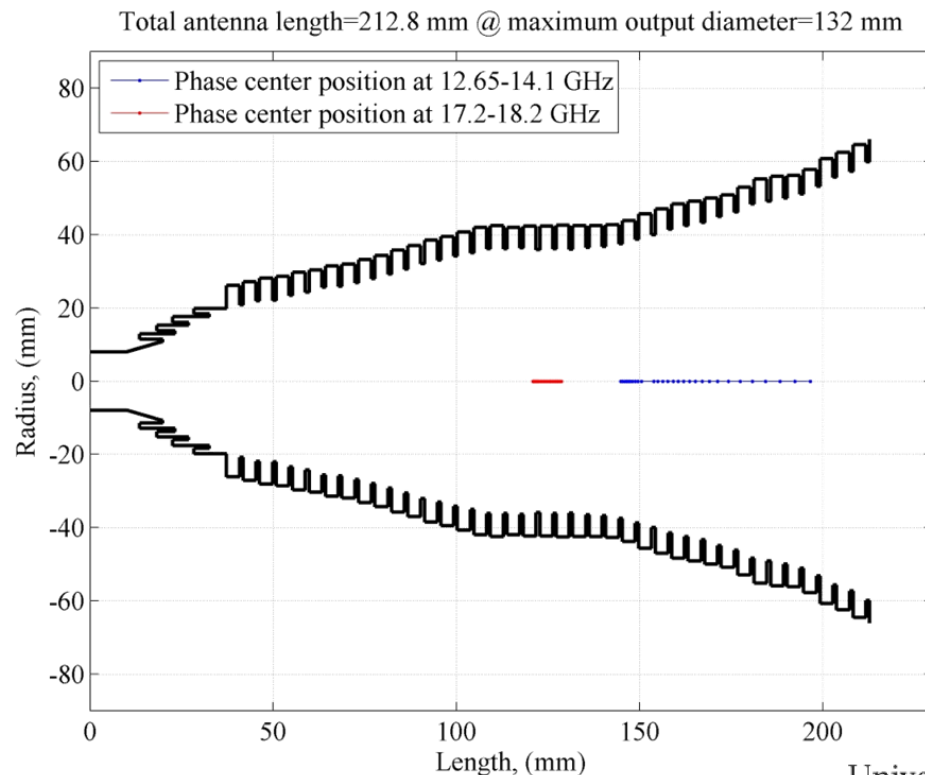
Examples

- Corrugated profiled feed-horn design for EADS/CASA to be onboard ASTRA 3B satellite IC horn:
 - Tx: 21.29–22.11 GHz,
 - Rx: 27.76–28.54 GHz + 28.81–29.60 GHz



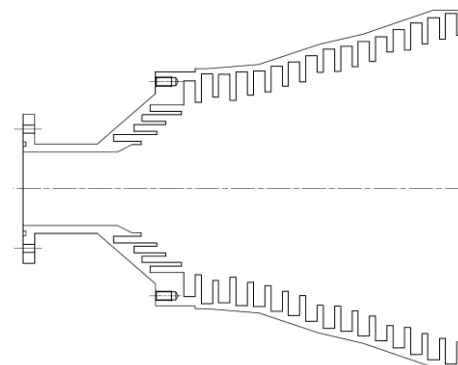
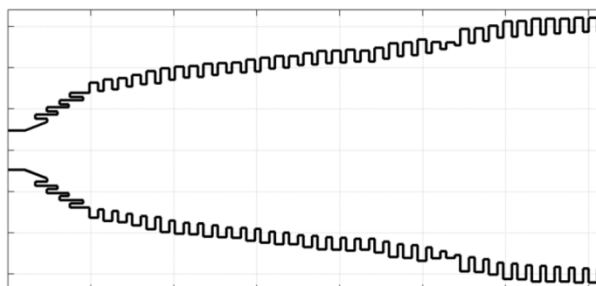
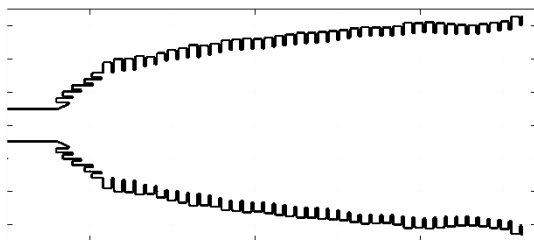
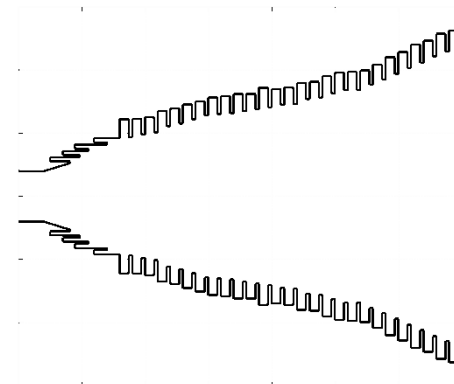
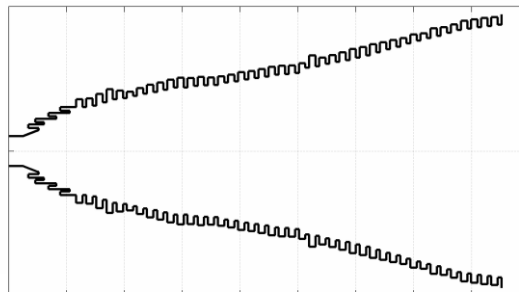
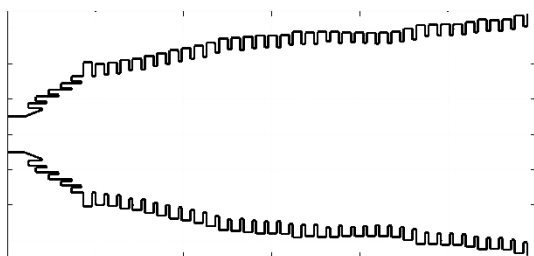
Examples

- Corrugated profiled feed-horn design for EADS/CASA to be onboard Hispasat 1E satellite
 - Europe Rx Horn: 12.65-14.10 GHz + 17.20-18.20 GHz
 - Principal requirement: phase center stability ± 15 mm



Examples

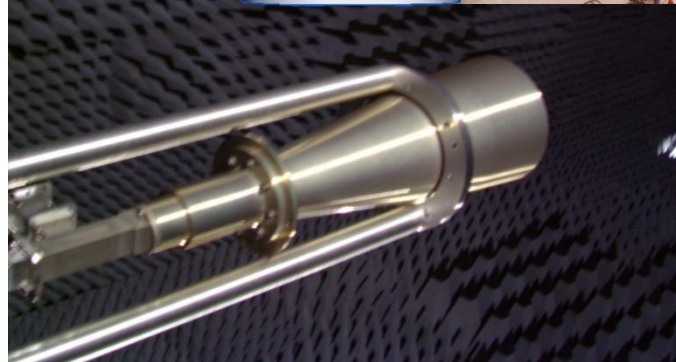
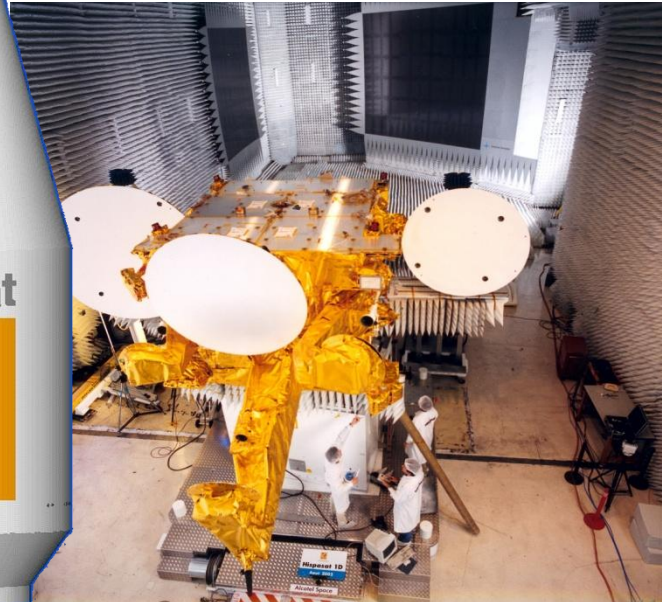
- Corrugated feed-horns profiles for different satellite and terrestrial applications



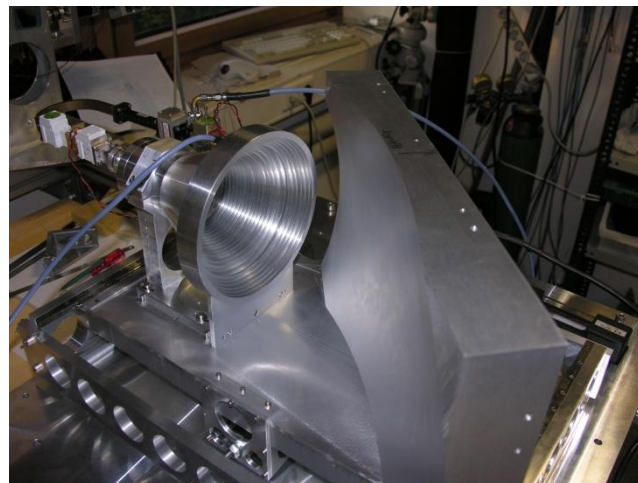
Some pictures... the Group



Some Pictures...



Some pictures...



Some pictures...

